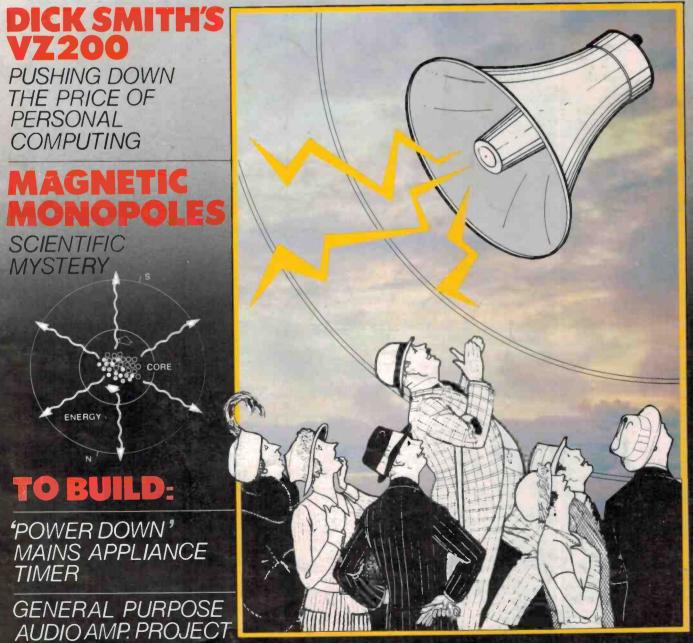
SCANNERS' WORLD-NEW BEARCAT REVIEWED



JULY 1983 \$2.35* NZ \$2.75

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INTRODUCING THE REVOLUTIONARY NEW MARANTZ COMPACT AUDIO DISC PLAYER.



IGITAL AUDIO, the greatest improvement in music reproduction since the birth of stereo is now available to give you sound more pure than any you have previously heard.

THE MARANTZ CD EXPERIENCE. It's dramatic. And instant. Plug the Marantz Compact Disc Player directly into your exist

Plug the Marantz Compact Disc Player directly into your existing system and it immediately upgrades the sound-limited only by the performance of your current equipment. You can expect astonishing channel separation. Very precise spatial imaging. Sensational dynamic range. Rich bass notes. Pure true treble. And, because the encoded music is read by non-contact laser-absolutely no background noise and no disc wear.

MARANTZ FEATURES. The Marantz CD73 is gold toned. Elegant. Simple. The control panel is clean and neat, with LED signals to indicate function and track selection. The highly sophisticated technology is push-button operated. The disc drawer glides with the smooth precision of electronic control.

Marantz is control convenience.

And technologically, Marantz uses a special integrated circuit with three functions (oversampling, a transversal filter and noise-shaping) which processes the original signal through various stages to give a dynamic range of 97dB. This amounts to a 1dB improvement over most other systems. You may never hear the difference. But Marantz cared enough to make their Compact Disc Player demonstrably closer to perfection.

IS THE MARANTZ CD73 REALLY ANY DIFFERENT? David Prakel for Hi Fi Answers magazine (UK) who did hear the difference said: "I have been surprised by the quite audible difference between different CD players and have already stated a preference for the sound of the Marantz machine in terms of its handling of 'ambience' and its sheer unfatiguing listenability. Other players I've heard in direct comparison have shown a bright veiling effect with more up-front presentation and a fatiguing quality."

Hear the CD73 for yourself. Call our local office for your nearest Marantz Dealer or write for further information.





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-comment ------

next month

OVER THE PAST few months, following a change of ownership and a traumatic move of the complete office and laboratory to a new location, publication of the magazine has got a little 'off the rails'. If you like your regular 'hit' of electronics/computing/ communications/audio every month and haven't been able to enjoy it when you've come to expect it, then we apologise. However, from next month we'll be back on the rails again with maybe just the odd 'loose wheel' to fix.

A serious hiccup in the distribution in New Zealand earlier this year has alarmed a great many loyal New Zealand readers. That problem was outside our control but has now been fixed. For good, we hope.

In this column last January, I said we'd be introducing new features during the coming year. The VIC-20 computer column, introduced in April, is well under way, the new Scanners' World column, introduced last month, is off to a good start and this month, we have a 32-page Op-amp Cookbook for you. We intend to include more cookbooks in future issues and, if the response is what we expect it to be, they'll become a regular feature.

Stick with us, we've got yet more goodies in store!



Roger Harrison Editor

services

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THE KUIPER AIRBORNE OBSERVATORY IN AUSTRALIA

A Lockheed L-300 jet transport aircraft, fitted with a 36" aperture telescope is operated by NASA to provide scientists with a very special infrared observatory. The telescope is carried to heights of 12 km (40 000 ft) to get it above 90% of the earth's water vapour, which absorbs infrared rays. This craft carried out a series of observing flights for Australian scientists in May and June. Jennifet Whyte tells the story.

SEVEN PERSONAL STEREO CASSETTE PLAYERS REVIEWED

A sort of 'across the market' review of seven 'famous name' personal stereo cassette players — we give the facts on how they measure up. Are they hi-fi or low-fi? Find out next month!

DEVELOPING THE MPF-1B MICROPROFESSOR

Here's one for the hackers! Ever wanted an alphanumeric character set for your MPF-1B? No problem — in fact, you have a choice of a limited 32-character set or a 64-character set. Sorry, no preprints, you'll just have to wait till August!

SCREEN SPOTTER' LIGHT PEN FOR THE MICROBEE

Just what all you 'bee bugs have been waiting for! This simple, low-cost device plugs into the 8-bit port and operates in the lo-res mode (that's because a hi-res light pen is 100 times as difficult, and a lot'more expensive). Software listing is Included.

ETI-166 FUNCTION AND PULSE GENERATOR

Now you can get into the guls of the project. Any electronics workshop worthy of the name needs a good function and pulse generator — this means you. All the functions and features you could wish to want. Another project in the David Tilbrook 'lab-standard' series.

Although these articles are in an advanced state of preparation, circumstances may affect the final content. However, we will make every attempt to include all features mentioned here.

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6 - July 1983 ETI



Channel 9 busted over unlicensed microwave equipment

Two television stations, Consolidated Press Holdings Ltd's Channel 9 in Sydney and Brisbane station QTQ-9, owned by AWA, have had summons issued against them for the alleged use of unlicensed microwave links between Sydney and Brisbane.

The Federal Department of Communications, which issued the summons, seized microwave equipment on June 3, following investigations by officers from the radio frequency management division.

The seized equipment consisted of two microwave stations; one had been set up at Somersby on the NSW central coast and the other was at Coolangatta, just over the Queensland border.

The Government acted on this matter because it doesn't want telecommunications links to sidestep Telecom, unless the Minister for Communications authorises otherwise. The department claims the two stations were links in a chain, comprising both authorised and unauthorised stations, which provides an alternative unofficial route for some television transmissions between Sydney and Brisbane. It was alleged that these stations had no purpose other than transmitting television programmes between north and south.

The Government's action is taken under Section 6 of the Wireless Telegraphy Act which bans the unauthorised establishment, erection, maintenance or use of "any station or appliance for the purpose of transmitting or receiving messages by means of wireless telegrapy".

The Act prescribes a penalty of five years' imprisonment or \$1000 fine.

Section 7 says such unauthorised equipment shall be forfeited to the Commonwealth.

Neither the Federal Department of Communications nor the Station Manager of Channel 9 in Sydney would comment on the matter.

The department launched its seizures after talks with the Deputy Crown Solicitor's offices in Sydney and Brisbane. The Queensland action was taken with the support of State police, but the NSW police refused to take part in the NSW action on the basis that it was a Federal matter.

Channel 9 has been summonsed to appear in the Federal Court on July 11 for a hearing.

In a separate inquiry, department officers are understood to be investigating unauthorised microwave links between Sydney and Melbourne.

Investigations claim to show evidence of a temporary system which could be set up at short notice with towers and microwave dishes in place but with no transmitters or receivers installed.

So far there is no evidence that these links have been in unauthorised operation.

Twelve years, 147 ETI issues, six overseas editions, six new magazines, and it's time to move on . . .

Founder-editor of Electronics Today International (ETI), Collyn Rivers, and his technical-editor wife Jan Vernon have left the publication and its now world-wide associated editions.



Jan and Collyn have started a technical writing and publishing company, Vernon, Rivers & Associates Pty Ltd, 18 Clifton Lane, East Balmain NSW 2041. other activities Among Vernon, Rivers & Associates will be preparing business computing and general technology features for the Bulletin, Australian Business, and the monthly general, software and equipment news sections of the recently redesigned Australian Computer Society Bulletin, and other publications worldwide.

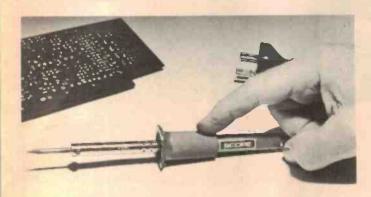
The organisation will also be undertaking specialised writing assignments for a number of trade and professional organisations in the computing and general technical industries.

Planning is also apparently well in hand for a new technical publication due to be launched within the next three months.

Electronics Today International, Collyn's original publication, was owned by Murray Publishers before its acquisition by the Federal Publishing Company division of Eastern Suburbs Newspapers earlier this year. Further info: Collyn Rivers (02)818-3559.



NEWS DIGEST



For hot fingered solderers

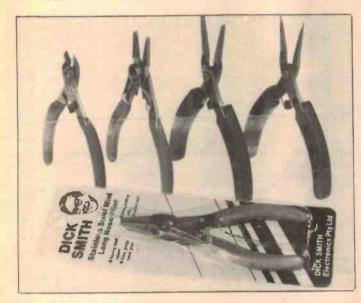
Scope Laboratories now offers a mains voltage 20 W non temperature controlled iron.

It features an air insulated grip, which helps keep the fingers cool during continuous soldering, and long life anti-corrosive iron plated tips.

The interchangeable tips come in four sizes and the tip and barrel are fully earthed for component protection.

Trade price is expected to be around \$15 in most states (tax extra).

For further information contact Scope Laboratories, 3 Walton St, Airport West Vic. 3042. (03)338-1566.



The pick of pliers

Dick Smith has a brand new range of four directly imported stainless steel high guality pliers.

The s/s transistor nipper, super sharp with a long life cutting edge, is ideal for pc board work. Cat T-3205.

The s/s mini flat nose plier is also for working on pe boards and can be used as a heatsink while

soldering. Cat T-3325.

The s/s mini needle nose plier is for getting into small spaces and working with small components like DS549s. Cat T-3570.

The s/s mini long nose plier can be used in 'rat's nests' of wires, great for pulling components out of tight spots. Cat T-3565. The pliers are priced at \$7.75

each and are available from all Dick Smith Electronics stores Australia-wide.

Electronic night fishing

The Matsushita Battery Company (National Panasonic) of Japan has been producing very small, light-weight batteries since 1978 with thicknesses down to 0.7 mm.

Recently they have employed lithium to obtain a higher energy density and this has resulted in a new application.

A fisherman's float using one of these batteries connected to a light emitting diode can be used as an imitation glow-worm. It is claimed that fish will bite at the glowing diode which (when fitted with a suitable hook) results in a much greater catch than would otherwise be obtained at night!

A reasonably high capacity is

required to maintain the glow for an adequate time, yet the batteries must not be large enough to alarm the fish. Lithium paper cells, nominally 3 V, with a diameter of 4.2 mm and a length of 26 mm weighing only 0.5 g offer a 20 mA-hr capacity. A cell of the same diameter and 10 mm in length has a 40 mA-hr capacity and is said to be able to attract fish for a period of at least 4 hours.

B. Dance

Innovative lead-acid battery design

Under the Federal Government's Assistance to Inventors Scheme, James Mackaness, an engineer of Cheltenham (New South Wales) has received \$10 000 to manufacture prototypes of a new lightweight lead-acid battery.

Mr Mackaness designed the battery after discovering that almost half the lead used in conventional batteries was wasted. By using much finer lead for the conducting function, and substituting polypropylene for the structure, he was able to reduce the wasted lead by 70%.

Known as a lead-acid automotive (LAAUTO) battery, Mr Mackaness' design also eliminates the need for the heavy — and expensive — lead grids used in conventional lead-acid batteries. This not only reduces the weight, but improves the charging and discharging rates.

While the battery's main market is presently in the vehicle industry, it offers a number of new applications, particularly for portable tools, machines, lighting and electronics.

Further details are available from Eric Hunter, Director of Public Relations, Department of Science and Technology, Canberra ACT 2600. (062) 64-4145.

Self-help for handicapped people

Prototypes of FRED (Friendly Education Device), a new training aid for handicapped people, are currently being manufactured by Thorn-EMI Electronics Australia.

The tested product is expected to be commercially available within the next 18 months.

A portable, microprocessorbased training unit, FRED is designed to be connected to a television set in the same way as computer-based television games. It uses computer-generated programs to provide displays on the television screen, which a handicapped person can control with a lever or joystick.

FRED allows handicapped people to move through training

or rehabilitation programs at their own speed and can be used by people with mental or physical handicaps, as no technical knowledge of the equipment is required.

The unit is expected to cost between \$1 200 and \$1 500 and further information is available from Eric Hunter, Director of Public Relations, Department of Science and Technology, Canberra ACT 2600. (062) 64-4145.

Magnetic monopoles a scientific mystery

Single north and south magnetic poles are predicted to exist but haven't yet been found. Could this be because of their unexpected properties: massive, slow-moving and rare?

Jennie Whyte

FOR MORE THAN 50 years physicists have been looking for magnetic monopoles, elementary particles which carry a single magnetic pole.

They are predicted to exist but so far no one has produced enough conclusive evidence of an observation. A new theory explains that this may be because the monopoles are too massive, slow-moving and rare.

Blas Cabrera, a researcher at Stanford University, thought that he might have found one early in 1982. But he has yet to verify an effect he has observed only once in six months. Theory suggests that there are not enough monopoles around today to explain why one should have been found in such a relatively short space of time.

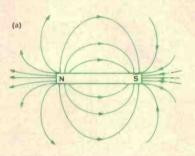
As early as 1269 Petrus Peregrinus, a French investigator of the magnetic properties of materials, noted that magnetic objects have paired regions of opposite polarity; that is, all magnets are dipoles. It seems that magnetic 'charges' or poles always occur in pairs, unlike electric charges which can occur as isolated positive or negative charges.

If you cut a bar magnet in half you end up with two smaller magnets, each with a north and south pole, rather than two pieces with opposite poles. This is because every atom in a magnetic material behaves as a tiny magnet, each atomic field being generated by electrons orbiting the atomic nucleus (just as an electromagnet is created by an electric current looping round a coil).

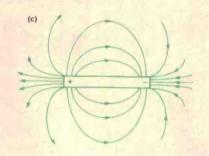
Speculation about the possible existence of magnetic monopoles has been going on for a long time. However, in the theory of electromagnetic phenomena, formulated by James Clerk Maxwell in 1864, the possibility of isolated magnetic charges was ignored since none had ever been observed. Over the past century Maxwell's theory has been put to many experimental tests and has never been found wanting. That fact alone severely limits the contexts in which magnetic monopoles might be found.

Interest in the idea intensified in 1931 when the British physicist Paul A.M. Dirac showed that an important observed property of electrically charged particles could be explained by assuming the existence of single magnetic poles.

Dirac was trying to explain the quantisation of electric charge; the fact that electric charge appears only in multiples of the charge of the electron and the proton. Dirac showed that if an isolated magnetic pole exists anywhere in the universe, electric



Dipole fields. Set up by a bar magnet.



An electric structure analogous to the above. Opposite electric charges are deposited at the ends.

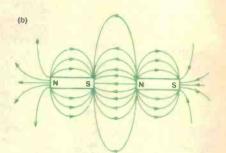
charge must be quantised everywhere. Until recently, Dirac's magnetic monopole hypothesis was the only explanation of the observed quantisation of the electric charge.

Dirac's monopole has a minimum unit of magnetic charge about 70 times as large as the corresponding unit of electric charge. He also predicted that the magnetic monopole would be matched by a magnetic antimonopole. However, his theory made no prediction about the mass or size of the magnetic monopoles, or about their abundance in the universe.

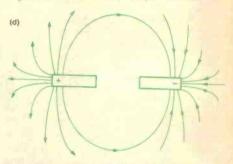
Dirac's predictions stimulated a rush of theoretical papers on the expected properties of the hypothetical monopoles, and several experiments were undertaken to detect them. Physicists searched for them in particle accelerators, cosmic rays and even moon rocks, but with no luck. However, they may not have been looking for the right effects.

Some interesting properties of magnetic monopoles arise when Maxwell's equations of electromagnetism are augmented to include magnetic charges and magnetic currents.

For example, as the velocity of a moving

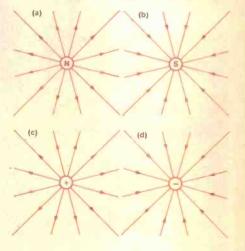


Magnet cut in half. Two smaller dipoles are created.



Electric analogue cut in half. The field remains dipolar because the electric charges that generate the field remain in place.

electric charge approaches the speed of light, its properties should increasingly resemble those of a magnetic charge. Similarly, a mov-



Symmetry exists. A north monopole (a) would have as its antiparticle a south monopole (b), just as the proton (c) has as its antiparticle the antiproton (d). ing magnetic monopole would begin to take on the properties of an electric charge at a speed approaching the speed of light.

These transformations, which follow from Einstein's special theory of relativity, have been confirmed experimentally for moving electric charges but not of course for moving magnetic charges.

A moving electric charge can lose energy by ionising matter; that is, it detaches electrons from their atoms. Because of the much stronger charge of the magnetic monopole, it would ionise atoms some 10 000 times more effectively. Thus a magnetic monopole passing through a photographic emulsion of the type employed by physicists to detect electrically charged particles would leave a track thousands of times darker than the track left by an electric charge moving at the same speed.

Because the monopole would lose energy to the ionisation process so quickly, it would slow down much sooner on entering a substance than does an electrically charged particle with the same kinetic energy.

A magnetic monopole traversing a superconducting coil one metre long would gain more energy than a proton acquires in the largest particle accelerator yet built.

The physics of magnetic monopoles has another curious feature which can only be made apparent by imagining that the flow of time can be reversed. This is a thought experiment suggested by Robert K. Adair of Yale University.

A proton is in a magnetic field which arises not from an electric current but from the presence of a magnetic monopole. Reversing time does not alter the polarity of the monopole and therefore leaves the direction of the magnetic field unchanged. The proton's path in the field of a monopole depends on the direction of time, an effect that violates the principle of time-reversal invariance.

The predicted effects of a magnetic monopole when time is reversed were for many years viewed as a serious argument against its existence.

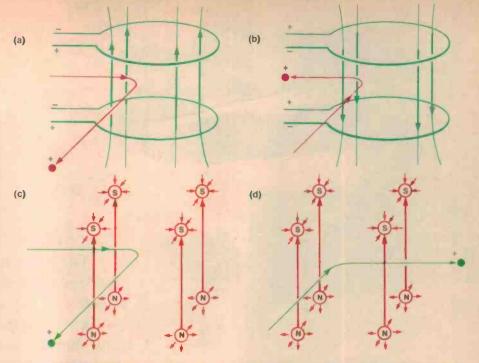
In 1964, however, an experiment was done at Princeton University which discovered an effect much like a violation of time-reversal invariance in the decay of the particles called neutral kaons. As this finding has become better understood some of the opposition to the idea of magnetic monopoles has abated.

Searching north and south

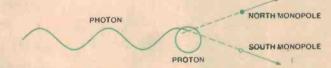
Knowing something of the properties of magnetic monopoles means that the experimental search for their existence can take a more positive direction. Soon after every new particle accelerator is commissioned magnetic monopoles are looked for in the debris of the initial high-energy particle collisions.

Monopoles have also been sought among the by-products of collisions between cosmic rays and atoms in the atmosphere. Other experiments have tried to detect them among the atoms of terrestrial and extraterrestrial substances. Samples of iron ore collected from the rocky outcroppings of old mountains are another potential source of material.

One detection method, first discussed in the 1960s, was implemented in the 1970s by Luis W. Alvarez and his colleagues at the Lawrence Berkeley Laboratory of the University of California. In their device a sample



Time reversal. In (a) a proton is moving through a perpendicular magnetic field generated by electric currents. If the direction of time is reversed (b) everything is reversed, the particle retracing its path in the opposite direction. In (c) the magnetic field is produced by an idealised array of north and south monopoles. Reversing time would leave the magnetic field unchanged (d). Although the proton would reverse direction, it would not retrace its path; a violation of the principle of time-reversal invariance.



Particle-antiparticle pair. A north monopole and a south monopole could be created when a high-energy photon interacts with an electrically charged particle such as a proton. The mutual attraction between the monopoles, however, would cause them to collide, converting their mass back into photons.

of material suspected of harbouring magnetic monopoles is passed repeatedly through a superconducting coil. On each pass of a magnetic monopole the electric current in the coil would presumably increase by a small amount. Because the coil is superconducting the incremental induced current would persist indefinitely. It's then a matter of measuring the extremely small signal induced by the multiple passes of a single monopole.

By means of this technique Alvarez and his colleagues were able to show that the density of magnetic monopoles in rock samples recovered from the surface of the moon is less than one for every 10^{28} protons. Even at this limiting abundance, however, there could still be an average of one monopole in every 20 kilograms or so of matter.

A less direct way of hunting for magnetic monopoles is to look for signs of the creation and destruction of a monopole-antimonopole pair. In theory a pair of this type could be created when a high-energy photon passes near a proton, just as an electron-positron pair is known to be produced. The oppositely charged monopoles would exist for only a moment, however, as they would soon come together and annihilate each other, converting their mass into additional photons.

In 1975 investigators at the University of California at Berkely and the University of Houston announced that they had discovered a magnetic monopole. Their evidence was an anomalously thick, dark track, presumably of cosmic ray origin, recorded on a stack of photographic emulsions and plastic sheets.

The detector had been exposed to cosmic rays while it was suspended from a balloon flown at high altitude for two and a halfdays. The area-time factor of the detector was roughly a million times smaller than that attained in previous searches in which no monopole had been seen.

Other problems with the monopole interpretation of the event subsequently led the experimenters to suggest instead that the track might have been caused by the passage of a superheavy atomic nucleus or a massive antiparticle.

One benefit of this episode is that it inspired a careful evaluation of how a magnetic monopole would lose energy through ionisation. Even so, the question still remains unsettled.

Superheavy particles

The prospects for magnetic monopole hunters suddenly brightened in the mid-1970s as a result of the independent work of Gerard 't Hooft in Utrecht in the Netherlands and Alexander M. Polyakov in Moscow. They both found that a certain class of theories of elementary particle interactions not only allows magnetic monopoles but also demands them.

These grand unification theories attempt to unify the four basic forces in nature gravity, electromagnetism, and the strong

Why Direct

Don't tangle with Technics. The majority of audio systems – even the most beautifully designed – have something ugly to hide.

It's that mass of jumbled-up connecting leads that you find, all too easily, at the rear of the equipment. Not only are they ugly, they're inconvenient, too.

And as audio components become smaller, the problem becomes bigger and more unsightly.

To solve this problem, Technics developed their Direct Connector systems, which eliminate all audio connecting leads between the tuner, amplifier, graphic equalizer and cassette deck.

Each of these components features a special flip-up connector to allow them to be literally plugged in to each other!

It's an elegant piece of Technics technology that results in a stylish, neat installation that can be put together or taken down for re-location in a matter of seconds.

The 315 Series.

But Direct Connector capability is not the only innovative feature in this new and compact series from Technics.

The SL-5 direct-drive, linear-tracking turntable employs its own plug-in connector system for the pickup cartridge.

This unique Technics development has been adopted as a World Standard.

It means you can compare and evaluate cartridges from leading manufacturers like Audio Technica, Ortofon, Shure, Stanton, Empire, Pickering, ADC and, of course, Technics without conventional setting up procedures.

Technics developed Connector systems.

No adjustment of tracking weight or bias correction is needed.

The innovations continue in the rest of the components: the SU-5 amplifier includes a Super Bass switch to enhance the bass response of a speaker system without inducing bass boom; the ST-5 quartz synthesizer digital tuner provides random access memory for 16 pre-set stations; the SH-E5 graphic equalizer – offers adjustment of 12 audio bands from 16Hz to 32Hz on each channel; whilst the RS-5 cassette deck – has soft touch controls, auto selection of metal, CrO₂ and normal tape settings plus convenient Cue and Review functions.

Finally, a pair of SB-F5 speakers with horntype tweeters and bass reflex porting turn the high quality electrical signals of the rest of the system into the high quality sound you expect.

Compact components, full-size warranty.

All components in this series are perfectly matched in styling and performance. **Technics**

And all are covered by a full 2-year warranty backed by Technics' reputation. Visit your Technics stockist soon and experience the superb



styling and brilliant sound of Technics' compact Series 315 for yourself.



and weak nuclear forces — into one graspable mathematical structure. According to this theory monopoles are 'superheavy', 10¹⁶ times the mass of the proton or ten nanograms, which is about as heavy as an amoeba.

Such a particle is so much heavier than any other elementary particle yet discovered that it could well explain why previous searches for monopoles have been unsuccessful.

Such heavy particles cannot be created at even the highest energies particle accelerators can reach, but they could have been produced copiously in the aftermath of the big bang with which, cosmologists generally believe, the universe began.

Up to times as little as 10⁻³⁵ seconds after the big bang, the universe would have been hot enough (almost 10³⁰ degrees Kelvin) to generate such particles. Both north and south magnetic monopoles would have been formed, and a small fraction of them would have recombined, annihilating each other. Most of the superheavy monopoles would have escaped an early death, however, and there is no reason to think they would not have survived to the present.

Researchers at the European Organisation for Nuclear Research (CERN) in Geneva decided that the interaction of monopoles with the galactic magnetic field sets a limit on the ratio of magnetic monopoles to protons of about one to 10²⁰. Given that abundance, some 200 monopoles per year would be expected to pass through an area of one square kilometre. A more conservative estimate, based on a more uniform distribution of monopoles in the universe, would result in a flux of a few monopoles per year per square kilometre.

So for the first time the theory of magnetic monopoles provides estimates of the expected mass and flux of magnetic monopoles. At least these estimates, even if they are rough, provide a fresh field for experimenters to explore.

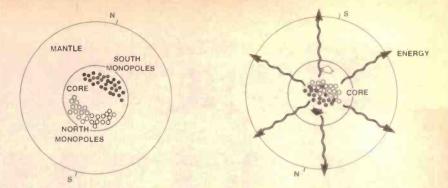
One place to look for superheavy monopoles is in large-scale natural effects. So what was the fate of monopoles in the material which collected together to form the solar system? One could speculate that as the earth condensed the monopoles would have sunk toward the centre under the influence of the planet's gravitational and magnetic fields. North monopoles would have collected near the south geomagnetic pole and vice versa.

From the geologic record it is known that the earth's magnetic field has reversed many times. Such a field reversal would cause the two separated populations of monopoles to migrate toward and then through each other. During their journey some monopoles and antimonopoles would be annihilated, liberating the enormous energy embodied in their mass.

From the measured heat flow at the surface of the earth one can set a rough limit on the number of monopoles trapped in the core; the number calculated in this way is consistent with other experimental limits on the abundance of superheavy monopoles.

But how do you find them?

The design of a detector to search for these heavy, rare particles is, however, not obvi-



Monopoles trapped in the earth. They tend to collect at two places in the earth's core near their opposite geomagnetic poles. Following a reversal of the earth's magnetic field (an event observed repeatedly in the geological record) the two segregated populations of monopoles would migrate through each other. Some would meet and annihilate each other, converting their mass into energy, which could be observed in the outflow of heat at the earth's surface.

ous. But there are a lot of ideas around, many of them quite bizarre.

The massive monopoles are expected to travel slowly, at speeds far below the velocity of light. The collision of a superheavy monopole and a stationary atomic nucleus would be like a steamroller hitting an ant. A cosmic ray monopole could lose a huge amount of energy to such encounters as it ploughed its way ponderously through the earth, and it might still emerge virtually unscathed from the other side.

So under these circumstances it is difficult to predict what degree of ionisation would be observed in a detector. Whatever happens, it's obvious that an extremely large detector is required if the experimenter is to observe a monopole event in his lifetime.

One detector which records the light generated by ionisation and covers many square kilometres has been developed at the University of Utah. The device, called the fly's-eye detector, is an array of photomultiplier tubes directed at the night sky; it registers the light given off by secondary particles produced by rare ultra-high-energy cosmic ray interactions in the upper atmosphere. As the secondary particles shower down toward the earth they collide with nitrogen atoms in the atmosphere, causing them to scintillate. However, the passage of a magnetic monopole, even with the most optimistic estimate of its ionisation rate, would give rise to less than a tenthousandth of the light needed to set off the detector

The ability of such a detector to respond to particle-induced scintillations is limited by background illumination from stars, overflying aircraft and other sources such as beacon lights on distant radio towers. Perhaps a fly's-eye detector could be installed in a large cave or salt mine such as those now being used to look for proton decay.

Another large-volume detector is the Deep Underwater Muon and Neutrino Detector which will be sensitive to events within a cube of ocean about a kilometre on a side. This detector will respond to the Cerenkov radiation emitted when a particle moves through the seawater faster than the speed of light in water. Unfortunately superheavy magnetic monopoles would probably move too slowly to give off Cerenkov radiation.

Some of the largest existing scintillation detectors are too small by a factor of about 100 to have a good chance of observing magnetic monopoles if the flux is limited by the galactic magnetic field. The contrary view holds that all searches with ionisation detectors are doomed to failure because the slow-moving, superheavy monopoles will cause no ionisation.

Another possible means of detection is based on the fact that the passage of any charged particle through metal is accompanied by eddy currents. These eddy currents are independent of the particle's speed and whether its charge is electrical or magnetic. A spherical metal detector has been designed but the signal can only be detected above the background noise if the detector is cooled to a few millidegrees above absolute zero; a difficult technical requirement with a large detector.

One comparatively simple strategy for detecting superheavy monopoles, which does not rely on assumptions about mass, calls for a superconducting coil similar to the one used by Alvarez and his colleagues.

Blas Cabrera claimed to have detected a monopole with a superconducting niobium coil five centimetres in diameter, kept in liquid helium at a temperature only 4.2 degrees above absolute zero.

Another plan is to mount a superconducting detector under an iron-ore processing plant which heats more than a million tons of ore per year to a temperature of 1700° C. At this temperature any magnetic monopoles trapped in the iron would be released, allowing them to fall through the detector.

The discovery of a magnetic monopole would rank as one of the finds of the century, comparable to the discovery of the positron, Dirac's other great prediction. If the monopole was found to be very massive, the case for some form of grand unified theory of elementary particle interactions would be strengthened.

Even if no magnetic monopoles are found, physicists, being what they are, will not view the negative evidence as conclusive.

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- C. Sutton, Magnetic monopoles fail to oblige the physicists. New Scientist, vol 97; no 1349, p 721, March 1983.

Yamaha's commercial power amps and recording mixer

Betamovie

The four heavy duty two-channel power amplifiers have a nominal power output (per channel, both channels driven into eight ohms) as follows: PC1002-120 W; PC2002 and PC2002M-240 W; PC5002M-500 W.

The 'M' suffix denotes a pair of large peak-reading meters that display power output in watts or dB.

All models have a typical separation of better than 80 dB wideband (95 dB at 1 kHz) so the two channels may be used for different programs (or two bands of bi-, tri- or quad-amped system). By sliding a rear panel mono switch, the amplifier outputs can be bridged to deliver significantly more power into eight or 16 ohm loads.

Harmonic distortion is typically 0.005% (20 Hz to 20 kHz, both channels driven) and intermodulation distortion is below 0.01%.

Their noise floor is from 110 dB to 120 dB below maximum rated output. Massive side-mounted heatsinks cool by convection. The RM804 is a new 8x4 bus mixer designed for four and eight track recording. Each channel has an electronically balanced mic/line input and an unbalanced tape input.

The XLR mic/line inputs have continuously variable gain trim controls which enable the sensitivity to be optimised for levels from -60 to -20 dB, covering most microphones and electric instruments as well as many line level sources.

Following the mic/line-tape switch, the signal goes through a 3-band equaliser.

For more information on Yamaha products contact Rose Music Pty Ltd, 17-33 Market St, Sth Melbourne Vic. 3205. (03)699-2388.

Betamovie here in November

Sony's revolutionary compact colour video camera — it packs the standard Beta-format cassette into its own body, resulting in easy handling and increased portability — will go on sale in Australia in November.

Called the Betamovie, the Australian package will include the camera, an ac adaptor and a battery pack, and is expected to sell for under \$1500.

The currently available Sony SL-F1E portable video recorder, which comprises separate camera and cassette pack, will be promoted as the 'pro' model of the Sony portable range, offering editing facilities and instant playback. The Betamovie will be aimed at the 'home movie' market.

Weighing just 2.48 kg, the Betamovie takes full advantage of the Beta cassette — it is more compact that the rival VHS cassette — and is only slightly larger than a deluxe 8 mm camera. It records on any standard Beta cassette, which can be played back immediately on any Beta-format video cassette recorder. An adaptor is not necessary.

The Betamovie, which measures 125 x 220 x 357 mm, features a single-head drum three-fifths the size of a conventional head, a 13 mm SMF Trinicon pick-up tube for higher resolution and sensitivity, an F1.2 lens with macrofocus and a 6x power zoom, and a through-the-lens viewfinder.

A single button on the handgrip starts and stops the recorder. The rechargeable battery pack fits into the handgrip.

For further information contact Sony Australia, 453-463 Kent St, Sydney NSW 2000. (02)266-0655.

Howdey, Partner...

TEAC Australia has released what it calls 'the ultimate personal stereo' — the TEAC Partner, Model PC-7RX.

The Partner incorporates continuous automatic reverse and a noise reduction system to eliminate tape hiss. This has previously been a problem with personal stereos, due to the very nature of headphone listening.

Available options include the LS-X3, an internally amplified mini-speaker system and the TP-7 AM/FM stereo tuner.

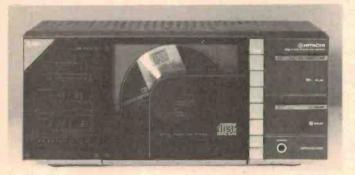
Meanwhile, TEAC Australia has been appointed Australian distributor for the Japanese range of Denon cassette and open-reel tapes. Denon is recognised as one of Japan's leading manufacturers of magnetic tapes and produces one of the largest ranges available.

For further information contact TEAC Australia, 115 Whiteman St, South Melbourne Vic 3205. (03) 699-6000.



Sight and Sound NEWS

Sight and Sound NEWS



Hitachi digital audio disc player

The Hitachi DA-1000 plays those compact digital audio discs which have been so much in the news lately.

Each compact disc has a diameter of 120 mm and has a playing time (on one side) of 60 minutes.

Since the disc is read optically through a transparent protective layer, dust and scratches have no effect on the sound. Of course, deep, concentric scratches can obscure the data to a point where any brand of CD player will not be able to conceal the errors.

To store information on a compact disc the sound is sampled at a rate of 44.1 kHz and then each sample is given a 16-bit code. The code is printed on the disc as a series of reflective pits which are read optically by a laser beam and then converted back into music.

The DA-1000 claims to have a

frequency response of 20 Hz to 20 kHz, a dynamic range of more than 90 dB and total harmonic distortion of less than 0.03%. The wow and flutter is to crystal oscillator precision and the channel separation is more than 60 dB.

The Hitachi CD player features pickup forward and reverse, cue, repeat and auto search which is a function that lets you quickly hear any selection no matter where it appears on the disc. Program playback allows you to program up to 15 selections for playback in any sequence automatically.

For more information on the DA-1000 contact Hitachi, 153 Keys Rd, Moorabbin Vic. 3189. (03)555-8722.



Philip's super VCR

Philips has launched a new top-of-the-line model for its range of VHS-format, front-loading video cassette recorders.

Designated the VR901, it has a cordless remote-control unit which duplicates the full 14 functions of the VCR's main control panel: adjustable picture quality; two picture search speeds in both forward and reverse; a clear freeze-frame picture; and a tape-elapsed indicator which is adjustable for cassettes from E30 to E240.

In addition to the usual record-

a ing facilities, the VR901 has a socketatthefrontoftheunitthat allows the direct connection of a video camera. The VR901, which retails at

\$1299, allows full stereo sound to be played through existing stereo systems and is equipped with a Dolby noise-reduction system.

For further information, contact Philips Industries, 15 Blue St, North Sydney NSW 2060.

Visionhire defends Videotex

Despite the critics' dire predictions for Videotex in Australia, Visionhire is confident that the computer-based two-way information system, which involved telephoneline access to data-storage bases, has a bright future.

Visionhire is a member of the newly formed Australian Videotex Industry Association which has set, as one of its goals, the dissemination of accurate information on Videotex, to counteract the confusion arising from misleading reports.

According to David Peers, Visionhire's technical director, "Most of the negative viewpoints have been almost word-for-word recitations of the British experience." (In Britain, the equivalent Prestel system has had a difficult beginning.) "On the other hand, Videotex found an immensely receptive market in the business community. From day one it has been aimed not as a system for public use but as a system of providing information retrieval by the business community.

"That's not to say, however, that Videotex will always remain within the narrow confines of the business market. There is absolutely no doubt that it will one day beome commonplace in our homes."



Stereo video from the General

General Electric's new Model GE-V-8900 stereo videocassette recorder incorporates two audio channels which allow playback of pre-recorded stereo video cassettes and taping of simulcast televion programmes.

The front-loading recorder has a multi-programmable timer for pre-setting the recording of up to eight programmes from any channel within a 14-day programme.

It also has a search function which moves the picture at nine times normal speed, both forward and reverse. A variablespeed playback gives double speed, slow motion, freeze-frame and frame-advance modes.

To prevent the loss of part of a progamme during a power fail-

Pocket television sets

Sanyo has started production of pocket-size LCD-based television sets with 75 mm and 100 mm screens.

Sanyo says it has overcome major problems with the driving systems, while the development of a 'stacked liquid-crystal matrix panel' has produced clear pictures. The panel combines a newly developed liquid crystal ure, the system has a clock/timer memory-hold system with a back-up time of about 10 minutes.

This is the second General Electric VCR to go on the Australian market — the first was the economy unit, Model GE-V-6900.

For additional information contact Rank Major Appliances, 19 Forge St, Blacktown NSW 2148. (02) 671-1322.

display with amorphous silicon thin-film transistors.

Reliability has also been improved, along with contrast, wider viewing angle and response speed.

The 100 mm screen model has a display size of 60×80 mm, and an overall size of $253 \times 30 \times 113$ mm. The reception bands include both UHF and VHF and the batteries provide four hours of viewing.

No other details have been released.

The Ultimate Performance.

The new Special Performance Series Components from Pioneer.

The sum of these components is probably the most technologically advanced sound reproduction system available from any one of the world's manufacturers.

A-70 Amplifier. 120 watts per channel. Dynamic Power Non-Switching Amplifier, featuring DC – Servo high-gain phono equalizer, Line Straight switch and L.E.D. peak power indicators.

F-90 Tuner. Digital Direct Decoder Tuner featuring Quartz – PLL Digital Synthesized Tuning, 8FM/8AM station presets, I.F. Bandwidth selector and Record Level Check switch.

CT-90R Cassette Deck. Three head auto reverse record and play cassette deck, featuring Pioneer's own Ribbon Sendust heads and the computer alded convenience of Music Search, Blank Search, Index Scan, Blank Skip and Reverse, Music Repeat and Real Time Counter; auto B.L.E. PLS70 Turntable. A fully automatic DirectDrive

PLS70 Turntable, A fully automatic Directorive Turntable, featuring Quartz-PLL coreless DC – Servo Hall Motor with Stable Hanging Rotor, low mass straight P.G. tone arm with Dynamic Resonance Absorber, and automatic operations of record detection/selection and repeat.

SG-540 Graphic Equalizer. Seven band graphic equalizer, featuring L.E.D. – lit slide controls, equalizer recording and tape monitor switch.



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If there's anything more you'd like to know about Pioneer's amazing Performance Series Components, please mai this coupon to 'Performance Components Brochure', Pioneer Electronics Australia Pty. P.O. Box 295, Mordialloc. 3195.	
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20FF. 240/100V, 10/23A at \$280 power supply. Contact R. Ballie on 597 1911.



Telex: AA25728. Melbourne (03) 328 2843



Sight and Sound NEWS



Digital age speaker systems

Pioneer's three new speaker systems feature two technologies developed by Pioneer to take maximum advantage of digital recordings.

The cone material is polymer graphite which is as stiff as metal and has the excellent damping qualities of paper, yet generates significantly less distortion than either

Pioneer has also developed a range of ribbon tweeters which have eliminated voice coils completely because the entire directdriven diaphragm surface acts as the voice coil. It is claimed that this results in a wider frequency range, better transient characteristics, lower distortion and far wider dynamic range than any competing tweeter designs currently available.

The S-910 is Pioneer's top bookshelf style speaker system and contains a large 30 cm woofer and a 10 cm midrange speaker. The beryllium ribbon has an

extremely low mass diaphragm for better transient response together with individual controls for adjustment of middle and high frequencies.

The S-710 bookshelf system has a large 30 cm woofer and 10 cm midrange with an aluminium ribbon tweeter which gives a wide dynamic range.

The S-510 bookshelf system has a 25 cm woofer and 4.5 cm midrange with an aluminium tweeter to handle frequencies up to 50 kHz

Retail pricing of the systems will range from approximately \$650 per pair for the S-910 down to approx. \$350 for the S-510 type.

For further information contact **Pioneer Electronics Aust. Ptv** Ltd, 178 Boundary Rd, Braeside Vic. 3195. (03)580-9911.

New Dynavector tonearm and cartridge

Concept Audio now has the Dynavector DV.501 tonearm which is a refined version of the DV.505 tonearm.

It features a dual pivot system and electromagnetic damping. It also incorporates a cueing device.

The complete package has a recommended retail price of \$498 and will fit onto most standard sized turntables.

To complement the DV.501, Dynavector has introduced the DV.10X.3 cartridge which is lower in mass than its predecessor, the DV.10X.2. It weighs three grams and the output voltage has been increased to 2.6 mV. The new DV.10X.3 does

not require a step up device and is designed specifically for use in moderate priced hi-fi systems.

The recommended retail price of the DV.10X.3 is \$148 and it is also available pre-packaged and mounted on a high quality magnesium alloy headshell at a price of \$175 complete.

Dynavector products can be obtained from Concept Audio Pty Ltd, 17/98 Old Pittwater Rd, Brookvale NSW 2100. (02)938-3700.



The Sony CDP101 The magic of digital audio becomes a magnificent reality.

Digital Audio is a revolution. The greatest advance in home music reproduction since the



gramophone record. As you'd expect, Sony is the leader of this revolution with its magnificent CDP-101 player that offers you original studio master quality at home.

For the technically minded, the specifications read more convincingly than any superlatives • flat frequency

response over the entire audible range • dynamic range and signal to noise ratio over 90dB • perfect channel separation • immeasurable wow

and flutter • negligible distortion. Sony's CDP-101 uses an

optical laser pick-up (incorporating three micro processors), it is easier to use than a conventional turntable and connects easily to your existing system.

Other features include • fully automatic linear skate front disc loading • automatic music sensor • dual function digital readout of playtime • audible fast forward and reverse • 10 function wireless remote control.

Compact Discs Last Forever

Just 12 cms in diameter, the Compact Disc plays up to 60 minutes of music. It's protected from scratches, dust and finger prints by a plastic coating; and because the pick-up is a laser beam, deterioration is non-existent. Reproduction remains perfect virtually forever.

Hundreds of titles will be available with many more to follow from major companies such as CBS.

CDP-101 Specifications

Frequency Range	5Hz-20kHz ± 0.5dB
Dynamic Range	more than 90dB
S/N	more than 90dB
Channel Separation	more than 90dB (at 1kHz)
Harmonic Distortion	less than 0.004% (at 1kHz)
Wow and Flutter	immeasurable

Contact Sony for the name of your nearest dealer.

Sydney (02) 266 0655, Adelaide and N.T. (08) 212 2877, Brisbane (07) 44 6554, Perth (09) 3238686, Melbourne (03) 419 3133, Launceston (003) 44 3078, Wollongong (042) 715777.

Jamo CBR 1703 loudspeakers

Louis Challis

on the local scene, these four-way Danish speakers are well matched to the compact disc players. Well designed, they are able to handle peak power levels of 600 W with low distortion.

MOST OF OUR READERS will probably remember 1983 as the year of the compact disc, affectionately known as 'CD' by the trade.

However, I suspect the trade, overwhelmed with problems, some of which are technical but generally financial, will not remember 1983 with so much affection.

One of the most interesting technical problems facing the high fidelity industry is the obviously wide gap between the technical performance of the new CD players and that offered by the 'average' loudspeaker of the type that you might buy in your local hi-fi shop.

CD players, with dynamic ranges of 90 dB and frequency responses that are flat from 10 Hz to 20 kHz, really need loudspeakers to be better than they have ever been before if they are to do credit to the new medium. So it is not surprising that most of the major speaker manufacturers or their importers are concerned that their cheaper speaker enclosures might not be suitable for the new demanding task.

Until these Jamo speakers turned up for review I must admit I had never heard of the firm. The speaker system that came out of the box was sufficiently unusual in its appearance to provoke a lot of comment in our office.

The 'CBR' concept, which stands for Centre Bass Reflex, but which I would describe as the Circumferential Bass Reflex, obviously has its merits. If this concept is half as good as it seems to be, it could well make this speaker stand out in the marketplace.

A four-way speaker system is not usually my cup of tea' because it makes the testing more complicated; however, I realise that this system does have obvious attributes.

It was purely coincidental that at the time of reviewing the Jamo speaker system I had the good fortune to have three different CD players in my house. It seemed that these two separate sets of circumstances were auspicious, for many people at this moment are considering whether or not their existing speaker systems are suitable for use with CD players.

It seems appropriate for me to look briefly at the specific attributes that a speaker system should possess if it is to be suitable for listening to a CD player system. Obviously, one of the first requirements is that its frequency response should ideally extend from 25 Hz or at worst 40 Hz through to 16 kHz and be as smooth as possible.

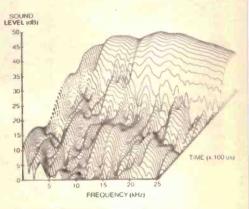
This frequency response should be supplemented by low distortion at high signal levels, low colouration from the drivers and the cabinet, low intermodulation distortion, a good transient decay response with minimal resonance, broad non-directional output in the horizontal plane and the ability to handle transient signals with peak powers of hundreds of watts.

So it was with these salient criteria in mind that I proceeded to evaluate the Jamo CBR 1703 loudspeaker system.

An overall view

After you remove the speaker grill the most unusual feature of the speaker system that catches your eye is the method of loading the low frequency driver. This makes use of a 'centre bass reflex' (abbreviated to CBR) system in which the woofer is inserted in the middle of its own circular, annular loading port.

The woofer is supported on four rectangular rubber isolating mounts which are apparently intended to vibrationally decouple the driver from the cabinet and from the loading port structure. The lack of resilience of these



Every picture tells a story. The decay response spectra for the Jamo CBR 1703 loudspeakers.

JAMO CBR 1703 LOUDSPEAKERS

Dimensions:	425 mm wide x 365 mm deep x 810 mm high
Weight: Price:	31 kg Rrp \$1695 per pair, Including
Manufactured:	pedestals In Denmark by Jamo Hi-Fi
Distributor:	Scan Audio, P.O. Box 741, Dandenong Vic. 3175.
	(03)793-5670.

four rubber mounts leaves one with the immediate impression that they can only be really effective at high frequencies and therefore they must be relatively ineffectual at low frequencies.

The woofer speaker is a 325 mm diameter unit with a large rolled plastic edge. Behind the main driver is a reasonably large ceramic magnet assembly. The woofer edge frame is designed to create the inner element of the CBR system, while the outer table element on which it is supported by the rubber blocks forms the other element of the CBR system.

The low mid-range unit is a 173 mm driver with plastic roll edge support, the mid-range unit is a 50 mm domed tweeter and the tweeter is a 25 mm domed unit.

The distribution of these individual low mid-range, mid-range and tweeter units on the top of the cabinet is asymmetrical with an L-shaped configuration. This must affect the uniformity of the sound field in the midrange region and detract from the directional attributes, rather than optimise the sound field generated by the speaker.

Thin foam plastic has been inserted around the mid-range and tweeter units to minimise the diffraction effects at the front of the loudspeaker. However, the foam that has been chosen has apparently been selected on the basis of its durability rather than its acoustical absorption.

The concept of forming the front of the speaker enclosure from a plastic moulding is delightfully simple as it saves all the expense of machining wood and all the other milling operations that are required in the construction of conventional wooden enclosures.

MEASURED PERFORMANCE OF : JAMO Model 1703

FREQUENCY RESPONSE:

35Hz - 18kHz

CROSSOVER FREQUENCIES:

280Hz/1400Hz/3500Hz

SENSITIVITY:

(for 90dB average at 2m)

5.6 VRMS - 3.9 Watts (nominal into 8:)

HARMONIC DISTORTION: (for 96dB at 1m)

	100Hz	IkHz	6.3kHz	
2nd	-43.3	-47.4	-55.2	db
3rd	-45.6	-44.9		dB
4 th	-62.2	-72.6		dB
5th	-66.0	-67.6		dB
THD	0.76%	0.71%	0.17%	

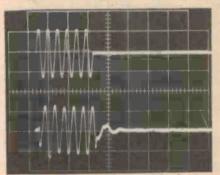
INPUT IMPEDANCE:

IOOHz	5.6 ohms
lkHz	9.5 ohms
6.3kHz	8.2 ohms
Minimum at 90Hz	5.4 ohms

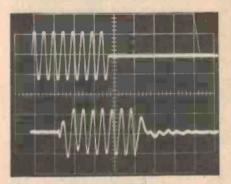
Absolute copyright in this review and accompanying measurements is owned by Electronics Today International. Under no circumstances may any review or part thereof be reprinted or incorporated in any reprint or used in any advertising or promotion without the express written agreement of the Managing Editor. Tone burst response of the Jamo CBR 1703 loudspeaker system. For 90 dB steady state sound pressure level at two metres on axis.

SOUND REVIEW

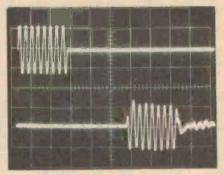
Upper trace is electrical input. Lower trace is the loudspeaker output.



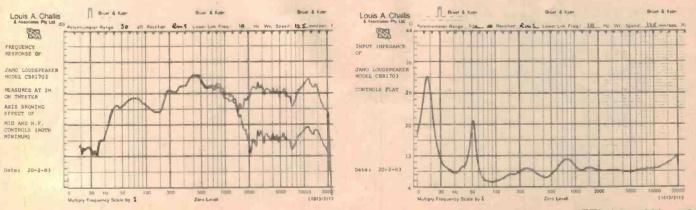
Output at 100 Hz (20 ms/div).



Output at 1 kHz (2 ms/div).



Output at 6.3 kHz (0.5 ms/div)



It also means that the complications in providing mounting holes, mounting systems, dimensions of those systems and the fabrication of the other complex parts of the system are minimised. It is all economically fabricated in one simple operation. Even the mounting holes for the fancy clothcovered front of the cabinet are moulded into the face structure and may then be forgotten.

What may have been overlooked is the resonance and damping characteristics that a plastic system provides when compared to a well-braced timber system.

The whole front of the speaker cabinet is moulded as a complete unit. This provides for all the mounting requirements of the individual speakers as well as incorporating the treble and mid-range attenuators on the top lip of the cabinet and overload indicator lights on the front edge of the cabinet.

The front clip-on panel which supports the black speaker cloth is also a plastic moulding and it seems that one of Jamo's strongest points is their moulding technology and the extent to which they are prepared to utilise their plastic mouldings and automated processes to reduce the manufacturing costs.

The cabinet structure is unusual as it features a contoured front panel. The full significance of this does not become apparent until you see the supporting steel pedestals on which each speaker is to be floormounted, or until you examine the installation instructions that show how the unit may optionally be mounted upon the wall.

The pedestals unquestionably enhance the appearance of the speakers as well as being functional. Each pedestal is a very strong, two-piece, 4 mm thick steel unit with interlocking elements. These provide a backrest which restrains the back of the speaker.

On the top of the pedestal are four selfadhesive pads which reduce the likelihood of the speaker system slipping off the top of the stand.

With typical Danish ingenuity the pedestal is very simply manufactured to provide strong and resilient support for the speaker as well as providing a small degree of vibration decoupling for low frequency energy being fed through to the floor. This is evident as the low frequency performance is substantially improved in the critical 25 Hz to 40 Hz region. The rear of the cabinet contains a terminal connection box with two colour-coded, large, spring-loaded terminals to accept bare speaker wires which may be up to 2.5 mm in diameter.

Objective testing

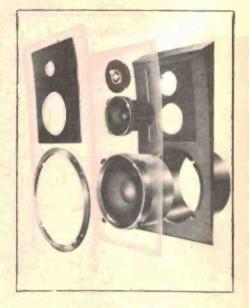
We placed one of the Jamo speakers in our anechoic room and proceeded to evaluate its objective performance. The first data we produced was the impedance curve which provides an assessment of the CBR system. This displayed two significant resonances; the first one is due to the main venting port of the CBR resonance which peaks at 12 Hz. The second resonance is the main speaker resonance which occurs at 51 Hz with a peak impedance of 21 ohms.

The minimum impedance measured is approximately six ohms at 90 Hz. The rest of the impedance curve from 100 Hz to 20 kHz is relatively smooth and consequently safe for virtually all amplifiers with which it is likely to be used.

The frequency response of the Jamo 1703, when measured on axis and without a reflective plane underneath the speaker, is not particularly smooth and lies within a range of ± 6 dB from 38 Hz to 20 kHz. Obviously with the speaker mounted on its pedestal above a reflective plane, the bottom end response would receive a significant lift as a result of the floor reflection. This does compensate for the drooping low frequency response that is visible in the level recording.

The frequency response exhibits a far greater number of peaks and bumps than I would have expected, particularly in the 350-500 Hz region and again between 2000 Hz and 7000 Hz. There is also a sharp rise in the response between 8 kHz and 16 kHz. These peak responses are measurable and, as we discovered later, audible and give the speaker a degree of frequency colouration which affects the audible response on both classical and pop music programme content. I pondered the non-linearity and suspect that the individual components are not closely monitored for uniformity of performance.

The speaker crossovers occur at 250 Hz, 1400 Hz and 3500 Hz, which are substan-

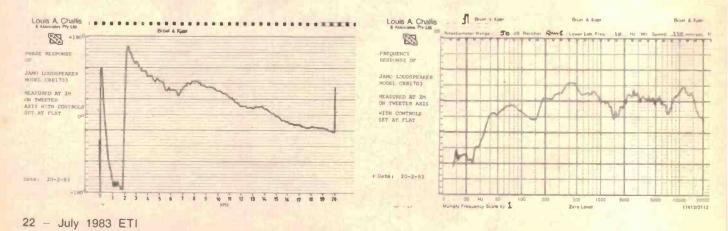


tially different from the manufacturer's claims. The lowest crossover between the bass driver and the low mid-range is not easy to pick, even with a microphone close to the speaker, and 1 gained the impression that the woofer was not delivering its fair share of the total output.

The mid-range attenuator, which is one of the 'push to activate' type, provides useful adjustment between 1500 Hz and 6 kHz, which is a maximum of -7 dB between 1800 Hz and 3 kHz. The treble attenuator, which is located next to the mid-range unit, provides effective attenuation from 2 kHz to 20 kHz with a maximum of -14 dB between 7 kHz and 14 kHz. The values and graduation of attenuation are sensible and it is clear that the designers have put a lot of thought into both the ergonomic and electronic design of the attenuators.

The overload indicator lights did not register at all during our testing and l suspect that I just did not push out enough power to activate them.

The speakers are quite efficient, requiring less than four watts to produce 90 dB at two metres. This means that with a 100 W stereo amplifier peak levels of greater than 110 dB



are possible in a normal moderately reverberant living room.

The phase response of the Jamo 1703 is extremely smooth and it is clear that, even though the unit contains four separate drivers and their associated crossovers, the designers have achieved a commendable result in precisely positioning the drivers and in designing matching crossover networks that appear to work well.

The distortion levels are particularly low, being well under 1% at 100 Hz and 1 kHz and less than 0.2% at 6.3 kHz. The low distortion levels are a primary feature of these speakers and the distortion levels do not really become significant until the power level approaches 100 W (i.e. 107 dB at two metres on axis).

I was not surprised to find that both the tone burst testing and the decay response spectra produced excellent results. Only moderate levels of decay resonance were apparent at about 5 kHz and even this was well down compared with the peak signal level.

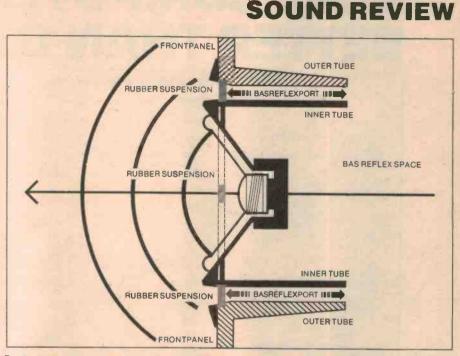
This part of the testing was particularly pleasing as it augured well for the subjective testing that followed.

Subjective testing

With a choice of three CD players, more than a dozen compact discs and many excellent conventional records, I had what I can only call an exciting time.

The music on the compact discs was exhilarating and the performance of the speakers was generally extremely good. The first thing that was apparent was the ability of the Jamos to handle an instantaneous peak power up to 600 watts per speaker without complaint and with remarkably low distortion. At these power levels the sound pressure level in my living room was over 114 dB and my family was starting to complain.

The sound was remarkably clean but by comparing the outputs of selected tracks against my reference speakers it was clear that the mid-range balance was not really perfect and that speech and singing had quite discernible colouration. This was readily observable on Classical Guitar (Albeniz Granada — track nine on the Sony CD disc YEDS4; Demonstration disc, Volume 1). The sound was not at all poor but was unques-



Fundamental construction of the Centre Bass Reflex system.

tionably different from what it should be. By contrast, on organ music the speakers were absolutely outstanding.

I played Virgil Fox's 'The Digital Fox' (Ultragroove UG9001) and was amazed how well these speakers coped with the record. The low frequencies were full-bodied with no signs of frequency doubling and remarkably free of distortion, while the mids and highs were rich and vibrant. The speakers reproduced the organ almost as well as any speakers I have heard, and they handled peak levels of over 400 W without complaint. The overload lights didn't come on and I was quite impressed.

l played an excellent Mobile Fidelity record, Earl Klugh's 'Finger Painting' (MFSL 1-025), and was rewarded with rich transients and exciting percussion. However, the response was coloured and the 'sound of the speakers' was superimposed on the sound of the music. The staccatos of the Earl Klugh were not quite as clean as they are on the monitors and I was a trifle less impressed.

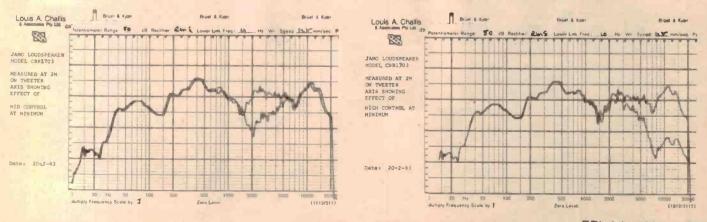
l returned to the CD players and was once

again rewarded with the magic of crystal clear digital recordings free of clicks, pops or background noise. The Jamos were really in their element and I was suitably impressed that these speakers are the first of a new generation of speakers that are well suited to the CD generation.

The Jamo CBR 1703s are an exciting speaker system which is generally well designed, well executed and which offers a performance bordering on the superlative. They are well suited for listening to both classical and rock music and are not afraid to handle power levels which would quickly destroy lesser speakers.

The only factor which detracts from an almost perfect testing and subjective assessment is the variability of the mid-band frequency response and the consequent frequency colouration.

Jamo would do well to spend a little more time and effort in matching their individual drivers if they would like a five-star rating instead of the excellent four stars that they won for this review.



SOME COMPUTERS ARE BETTER THAN OTHERS

COMPUTER COMPARISON CHART						
	SPECTRAVIDEO	APPLE IN PLUS	ATARI 800	COMMODORE 64	COM.000045	TANDY TRS-80
BASE PRICE	\$499	\$2100	\$1100	\$699	\$299	\$549
COMPUTING POWER FEATURES BUILT-IN ROM EXPANDABLE TO BUILT-IN EXTENDED MICROSOFT BASIC BUILT-IN RAM EXPANDABLE TO	32K 96K YES 32K* 144K**	12K N/A YES 48K 64K	10K 42K ADDITIONAL COST 16K 48K	20K N/A NO 64K N/A	20K N/A NO 5K 32K	8K 14K NO 4K 32K
KEYBOARD FEATURES NUMBER OF KEYS USER DEFINE FUNCTIONS SPECIAL WORD PROCESSING GENERATED GRAPHICS (FROM KEYBOARD) UPPER/LOWER CASE	71 10 YES YES YES	51 N/A NO UPPER ONLY	61 4 NO YES YES	66 8 NO YES YES	66 8 NO YES YES	53 N/A NO YES UPPER ONLY
GAME/AUDIO FEATURES SEPARATE CARTRIDGE SLOTS BUILT-IN JOYSTICK COLORS RESOLUTION (PIXELS) SPRITES SOUND CHANNELS OCTAVES PER CHANNEL ADSR. ENVELOPE	YES 16 256 x 192 32 3 8 YES	NO NO 15 280 x 160 N/A 1 4 NO	YES NO 128 320 x 192 4 4 4 NO	NO NO 16 320 x 200 8 3 9 YES	NO YES 16 196 x 184 8 3 9 XES	NO NO 8 192 x 256 N/A 1 3 NO
PERIPHERAL SPECIFICATIONS CASSETTE AUDIO IO BUILT-IN MIC DISK DRIVE CAPACITY (LOW PROFILE)	2 CHANNEL YES YES 256K YES	I CHANNEL NO NO 143K NO	2 CHANNEL YES NO 96K NO	I CHANNEL NO NO 170K NO	I CHANNEL NO NO 190K NO	I CHANNEI NO NO IS6K NO
CP/M COMPATIBILITY (80 column programs) CP/M* 2.2 CP/M* 3.0	YES	NO*** NO	NONO	NO	NO	NO

* 16K user address able plus 16K graphic support ** 128K user address able plus 16K graphic support *** Apple II can accept modified 40 or 80 column CP/ **** Commodore 64 accepts 40 column CP/M

CP/M is a trademark of Digital Research, Inc.

OURS IS <u>MUCH</u> BETTER

When you start comparing Spectravideo's SV-318 to other personal computers, you'll find there really is no comparison. The SV-318 is the only logical choice, because it does more than some computers costing 4 times as much. And its abilitles simply embarrass other computers in this price range.

The SV-318 isn't just more capable. It's much more capable. No other computer at even twice the price comes near its 32K ROM expandable to 96K. Or to its 32K RAM expandable to 144K. And no other computer has a built-in joystick/cursor control—an immeasurably useful feature when it comes to playing your favorite video game. Further, the SV-318 has, as its resident "language" Extended Microsoft Basic, the industry standard. It even has built-in CP/M (standard 80-column program), so you can immediately utilize over 10,000 existing software programs.

The SV-318 isn't just more expandable. It's much more expandable. Unlike many other so-called computer systems, all our important peripherals are available at once. That means you can get almost full usage out of your SV-318 from the day you buy it. With the Super Expander, Data Cassette, Floppy Disk Drive, Dot Matrix Printer, Graphic Tablet and SV-800 Series Expansion Cartridges, there's almost no end to the work you can do. Or to the fun you can have. The SV-318 is well designed to interface with new options as they become available, too. All this adds up to a computer you'll grow into, not out of.

The SV-318 is not only eminently affordable, it's the first real bargain of the computer age! Besides business application, home budgeting, word processing, programming and self-teaching, the SV-318 is the best entertainment value in town. Not only can you use it with your TV or color monitor to play hundreds of different video games,



FOR UNDER \$500

with the optional SV-105 Graphic Tablet you can draw pictures, graphs, charts and other visual images on your TV screen. Considering what you get for what little you pay, the SV-318 is once again the only logical choice.

Whether you're investing In your first computer, or are already well versed in today's most important machine, you'll find that the SV-318 is the only logical choice for you.



computing today

Spectravideo SV 318 personal computer

The Melbourne-based firm, Videoactive, the Australian agent for Spectravideo video games (including Planet Patrol, Master Builder and Quickshot) has just released the Spectravideo SV 318 personal computer in Australia.

Spectravideo is an Americanbased electronics company with an affiliation with Bondwell, a large firm located in Hong Kong.

Combining the research and development facilities of the two companies has provided the technological and manufacturing capabilities to produce the SV 318 personal computer.

The SV 318 computer features, as standard, 32K RAM; 16K is for graphics and 16K is user addressable memory. Memory expansion can be accommodated up to a total of 144K RAM. The 32K ROM is expandable to 96K with a custom extended Microsoft BASIC interpreter built in.

The computer is compactly and stylishly constructed in a console unit with keyboard and TV modulator. The keyboard unit can be linked up with the full range of peripheral options or user devices through the SV 601 Super Expander or adaptor to suit all requirements.

Other features include a Z80A microprocessor with 36 MHz clock, CP/M compatibility, combined cursor/joystick control, easy loading cartridge slot, arcade quality graphics and sound, a seventy-one-key, multifunction keyboard and built-in word processor keys.

With these features and priced at \$499 Videoactive claim that the SV 318 offers exceptional value.

The SV 328 personal computer

Buggy brings computers to life

A robotic vehicle, the BBC Buggy, has been designed to provide children and adults with a stimulating and versatile introduction to the world of computer-control technology.

Developed for use with the BBC microcomputer, the 127 mm square vehicle has three wheels driven by two precision stepped motors and is equipped with detectors to feel objects and seek out light sources, together with an infra-red transceiver to read bar codes.

A series of 13 programs are supplied which explore important aspects of microcomputer technology. These include a memory function, line follower, route planner (in which the buggy follows a picture of the route on the computer screen), programs for exploring objects or a given territory, working out the dimensions and transferring them to the screen and a sunseeker program (which enables it to seek out light sources among a maze of objects). It can also compose music, reading the information from a series of barcodes.

will follow later this year

More details on the SV 318 are

available from Videoactive

Electronics, 70 St Kilda Rd. St

Kilda Vic. 3182. (03)537-2000.

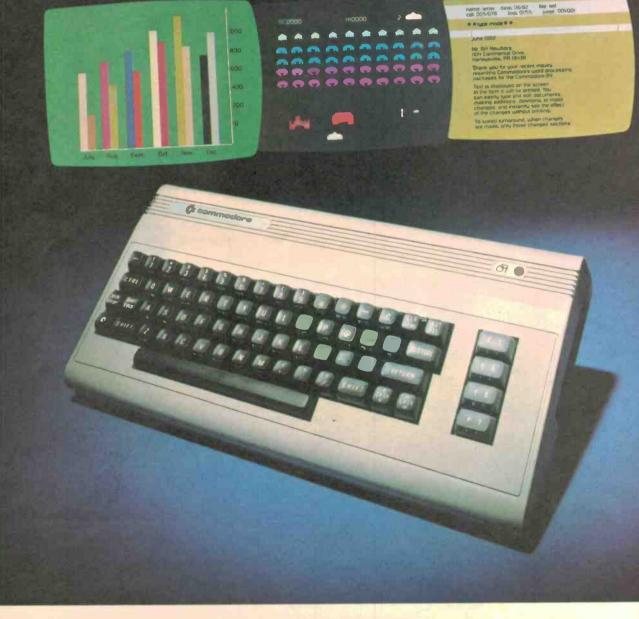
Though primarily developed for the BBC microcomputer, the manufacturer, Economatics, plans to include an interface with the vehicle so that it can be used with other microcomputers.

For more details contact Economatics, 4 Orgreave Crescent, Dore House Industrial Estate, Handsworth, Sheffield, United Kingdom.





THIS IS THE COMMODORE 64 ONLY \$699.



"THE COMMODORE 64 COULD BE THE MICROCOMPUTER INDUSTRY'S OUTSTANDING NEW PRODUCT INTRODUCTION SINCE THE BIRTH OF THIS INDUSTRY."

They're speaking to a group as interested as anyone else in the future of computers: the people who buy stock in the companies that make computers.

If, on the other hand, you're a person whose livelihood depends on a personal computer – or whose leisure time revolves around one – what follows should impress you even more than it impresses investors.

MIGHT MAKES RIGHT.

The value of a computer is determined by what it can do. What it can do is largely determined by its memory.

The Commodore 64's basic RAM is 64K. This amount of power is unusual enough in a micro at any price.

At \$699, it is astonishing.

Compared, with the Apple II+,[®] for instance, the Commodore, 64[™] offers 33% more power at considerably less than 50% of the cost.

Compared with anything less, it's even more impressive.

And it can effectively double your computerequipped work force.

PILE ON THE PERIPHERALS

Because the basic cost of the 64 is so low, you can afford more peripherals for it. Like disk drives, printers or even printer-plotters.

This means you can own the 64, disk drive and printer for a little more than an Apple II+ computer alone.

HARD FACTS ABOUT SOFTWARE

The Commodore 64 will have a broad range of custom software packages including an electronic spreadsheet; business graphics (including printout); a user-definable diary/calendar; word processor; mailing lists, and more.

With BASIC as its primary language, it is also PET BASIC compatible.

The Commodore 64 will also be programmable in USCD PASCAL, PILOT and LOGO.

And, with the added CP/M[®] option, you will have access to hundreds of exciting software packages.

SHEARSON/AMERICAN EXPRESS

THE FUN SIDE OF POWER

The Commodore 64 can become very playful at a moment's notice.

You can use Commodore's plug-in game cartridges or invent your own diversions. All will be enhanced by brilliant video quality and high resolution graphics (320 × 200 pixels, 16 available colors, 3D Sprite graphics), plus outstanding sound.

The 64's built-in music synthesizer has a programmable ADSR (attack, decay, sustain, release) envelope, 3 voices (each with a 9-octave range) and 4 waveforms. All of which you can hear through your audio system and see in full color as you compose or play back.

NOW'S YOUR CHANCE

If you've been waiting for the "computer revolution," consider it as having arrived.

Through its 25 years of existence, Commodore has been committed to delivering better products at lower prices.

Today, the company's vertical integration has resulted in the Commodore 64's price performance breakthrough heralded by Shearson/American Express.

Visit a Commodore Computer dealer and discover the 64 soon.

It will expand your mind without deflating your wallet.

CPM™ is a registered trademark of Digital Research. Inc.
Commodore Business Machines Pty. Ltd.
5 Orion Road, Lane Cove NSW. 2066. (02) 427 4888.
Please send me more information on the Commodore 64.[™]
Name
Address
Postcode Phone
BF
COMMODORE BUSINESS MACHINES PTY. LTD.
5 Orion Road, Lane Cove NSW. 2066. (02) 427 4888.

commodore

Printout

Dictionary program

Software Source has signed an agreement with Oasis Systems, California, for exclusive distribution of the powerful new dictionary program, 'The Word Plus'.

Software Source will be providing full technical support for the package which is something new to those who have used a conventional 'spelling checker' package.

The Word Plus is a series of integrated spelling check tools which perform a wide range of spelling check and word analysis functions, including the ability to find incorrectly spelt words in context and to suggest similar words and to actually carry out the spelling correction.

AED Melbourne

Elston Micro has been appointed as Melbourne dealer for the Australiandesigned and manufactured AED Universe range of computers.



The Word Plus is compatible with nearly all CP/M wordprocessors and text editors.

For further information contact Software Source, 344-348 Oxford St, Bondi Junction NSW 2022. (02)389-6388.

AED Universe's Supercomputer II features a choice of eight- or 16-bit single- or multiuser operating systems, and has full S100 buss compatibility.

For further details, contact Elston Micro, 53 Waverley Rd, East Malvern Vic 3145. (03) 211-5542.

Information Technology Week

Information Technology Week, an annual national event aimed at heightening community awareness and understanding of information technologies, is being held in New South Wales from August 7-13.

As part of the activities Sydney's Power House Museum (Mary Ann Street, Ultimo) will stage an exhibition of computers, as well as providing visitors with 'hands-on' computer experience in its new 'computer lab'.

A variety of displays and workshops will be held in shopping centres throughout Sydney, as well as at the Institute of Chartered Accountants, the OTC showrooms in Martin Place and local council chambers and

Computer games spectacular

Australia's first major computer games exhibition, Electronic and Computer Games and Toys '83, will be staged at the Sydney Entertainment Centre from August 18-21.

libraries.

Other participating organisations include the Metropolitan Water Board, the Sydney Stock Exchange, the State Government Information Office and various business houses.

For further information contact Hartley North, Executive Secretary, Information Technology Week, P.O. Box K701, Haymarket NSW 2000. (02) 218-8080.

Organised by the Industrial Presentations Group of Companies, the exhibition will be open to both the trade and the public; on the mornings of August 18-19 entry will be restricted to the trade.

For more details, contact Industrial Presentations Australia, 4/389 Victoria Avenue, Chatswood NSW 2067. (02) 412-4124.

MicroBee games software

M.B. Software is a recently established software house specialising in high quality software for the MicroBee.

Mine Drop is an exciting chase game for the standard 16K MicroBee. The program is written in machine code and is fully interactive. The operator controls the movement of Supertank as it

Pineapples are not Apples

Apple Computer has been awarded a preliminary injunction against a Los Angeles distributor of one of the copies of the Apple II personal computer being sold in the US.

Formula International, distributor of the 'Pineapple' moves around the screen picking up urgently needed supplies.

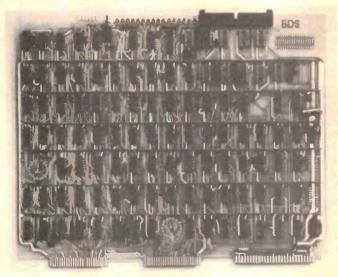
Supertank is pursued by a homing rocket, Bingle. However, Supertank has a defence. Mine Drop, a mine, can be placed and exploded on command. Bingle is destroyed if it is passing over the mine at the time of detonation.

Mine Drop is available on cassette for \$12.95 from M.B. Software, 248 Brunswick Rd, Brunswick Vic. (03)380-9805 and from MicroBee outlets.

computer, has been enjoined from copying or selling any of Apple's copyrighted software programs in the computer's memory circuits or diskettes.

This includes the Autostart ROM, Applesoft, DOS 3.3 and Integer BASIC programs.

In Australia steps are being taken to ensure that further sales of bogus Apple computers are prevented.



High speed printers for HP 3000 series

BDS has introduced a new microprocessor controlled printer system which gives users of HP 3000 series 30/33/40/44/64 computers printing speeds up to 1500 lpm.

The HPC-33L controller includes a Z80 microprocessor and an Intel 8291 talker-listener for interfacing with the HP general purpose interface lines. The printer is daisychained with other peripherals in the standard series 33 fashion.

The unit has three modes of operation, the control, transfer and print modes. It also responds to self test commands which result in remote initiation of internal diagnostics.

Long line drivers, which are standard, allow the printer to be located up to 200 metres from the system.

The range of Dataproducts' band printers, which are supplied in conjunction with the controller, have speeds of 300, 600, 1000 and 1500 lpm.

For further details contact BDS Computer Australia Pty Ltd, 8th Floor, 445 Toorak Rd, Toorak Vic. 3142. (03)241-8901.

Plato and the home computer

Platoeducational courseware, originally developed by the Control Data Corporation for use in schools through terminals connected to a mainframe computer, will be available later this year on diskette for the Texas Instruments 99/4A home computer.

There are 64 packages, offering mathematics, reading and grammar programs for students in Years Three to Eight, and 44 packages covering mathematics, writing, science, social studies and reading for Years Nine to 12.

The initial Plato package includes an interpreter solid-state cartridge and diskettes containing a survey to help parents or teachers select courseware for individual needs. For those not familiar with the operation of the computer, there is a program designed to teach beginners how to use the TI-99/4A keyboard.

These Plato packages will cost \$49.95 each.

To take advantage of the Plato software, TI-99/4A owners will need a Texas Instruments peripheral expansion system, a memory expansion card, a disk memory drive, and a disk controller card.

Also soon to be available for the TI-99/4A home computer are 15 educational software cartridges, featuring mathematics learning exercises for children in the kindergarten to Year Nine age group.

For further information contact Texas Instruments Australia, 6-10 Talavera Rd, North Rvde NSW 2113. (02) 887-1122

Colour video display processor

Texas Instruments has a new video display processor. designed to interface between a microprocessor and a PAL video colour monitor

The new TMS9929A has four display modes, graphics 1, graphics 2, multicolour and text mode, displaying information in 15 colours. As well, using an advanced planar representation, one can simulate objects in three dimensions.

Low cost 16K-type dynamic memories are used for storage of the display parameters and the TMS9929A does an automatic and transparent refresh of the dynamic RAMs.

A unique graphics feature of the TMS9929A is the special animation patterns called sprites which provide smooth motion and multilevel pattern overlaying. Up to 32 sprites can be displayed on the screen at any one time

Interfacing to most 8-bit or 16-bit microcomputers is easy, using a minimum number of external components.

The chip is available now for \$35 from Texas Instruments Australia Ltd, 6-10 Talavera Rd, North Ryde NSW 2113. (02)887-1122.



Supa edit

Supa Edit, by eei is a powerful enhancement to level II BASIC's 'edit' and 'list' commands and is designed to suit the needs of the serious programmer/ hobbyist.

It is claimed that Supa Edit will improve programming speed and is suitable for both Tandy TRS-80 and DSE System 80, Model I and Model III, 16K, 32K and 48K tape based systems. A lower case driver is also included for Model I systems.

This machine language pro-gram occupies less than 0.5K of user memory, especially important to users of 16K systems, and is simple to operate.

The introductory price, including sales tax, package and posting is \$14. Send payment to **Elite Electronic Industries** Pty Ltd, 36 Luxmoore St, Cheltenham Vic. 3192. (03) 583-1201.

Stop wasting time program editing. Start VEDIT-ing!

If you spend hours a week working on a terminal or computer, editing programs and vou're not using VEDIT full screen text editor, then you're working too hard! VEDIT's combination of wordprocessing, powerful macros and full screen (visual) mode will slash your programming and reediting time. Here's how VEDIT will help you get the job done faster:

- Blocks of your program can be moved or copied within a file.
- □Automatic indenting for structured programs such as PASCAL or 'C'.
- Reformatting of paragraphs between set margins.
- Any portion of the text may be sent to the printer.
- The 'UNDO' key will 'undo' the mistake vou just made - before it ruins your text.
- UVEDIT is fully customisable to a huge range of terminals. (Write for a full listing)
- The powerful command structure includes search and replace, macro commands for repetitive command sequences and special functions.
- Sophisticated buffering enables editing of files larger than main memory. Shownoom. 344:348 Oxord Street Bond Junction, New South Wates Bostal Address: Po Box 364, Edgecilit, NSW 2027, Pr. 1021, 369 6388

Name

Postcode

Printout

Games cartridges for Intellivision and VIC-20

Imagic has released three new games cartridges for the Intellivision system and two for the popular VIC-20.

Dracula, Ice Trek and Tropical Trouble are for the Intellivision. Dracula features the legendary count who prowls the night in search of sustenance, pursued by ravens, wolves and the law. If daylight dawns, he's finished!

Ice Trek makes you the hero fighting your way across icy wastes dodging caribou and snowdrifts. With an ice hook and fire torch you must build and defend an ice bridge.

In Tropical Trouble you are marooned on a swamp-andvolcano-infested island and must rescue the heroine. Plagued by snakes, man-eating clams, hostile gorillas and spitting lava, this game requires co-ordination, quick reflexes and stamina.

Demon Attack and Atlantis are for the VIC-20 — and it doesn't take much imagination to guess what they're like.

Further information can be obtained from Don Dennis, Imagic, P.O. Box 300, Forestville NSW 2087. (02)981-2744.

Apple gains injunction against the Wombat!

Apple Computer has been granted a court order by the Supreme Court of New South Wales, prohibiting a microcomputer dealer from selling microcomputer copies of the Apple computer.

The dealer, Microeducational, of Newcastle (New South Wales), also agreed to pay the court costs involved.

The microcomputers in question, the Golden II and the Wombat, contained substantial parts of the Autostart and Applesoft ROMs, which are Apple Computer's intellectual property.

Commenting on the order, David Strong, general manager of Apple Computer, said, "We are gratified by the Supreme Court order. As a precaution, we recommend that prospective purchasers of personal computers or accessories which might violate copyright laws check first with Apple Computer, or any of our authorised dealers. In some cases, purchasers are often misled by statements such as 'compatability with programs for the Apple'."



High-speed joystick

Discwasher, an American company previously associated with record cleaning accessories, has expanded its product range to cater for the video game and computer accessory market.

Heading the new range is the Pointmaster, a joystick that plugs into Atari and Commodore VIC-20 game centres. It features a high-speed thumb trigger, comfortable handgrip, selfcentring mechanism and a 1.5 m cord. It costs about \$29.99.

The manufacturer claims the Pointsmaster has a very fast reaction time, enabling the operator to achieve much higher game scores than with a normal joystick.

Also available is a rapid-fire adaptor (\$14.99) which allows the joystick to be fired continuously, like a machine gun.

The Discwasher range is handled in Australia by Arena Distributors, 642 Albany Highway, Victoria Park WA 6100. (09) 361-5422.

HARD DISKS AT FLOPPY PRICES

We have a limited quantity of 6.37 MB (unformatted) 5° Winchesters available at the special price of

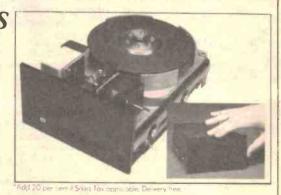
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Pennywise Peripherals, 96 Camberwell Rd, Hawthorn East, Vic, 3123. P.O. Box 398, Camberwell, Vic, 3124. Phone (03) 82 2389. Telex AA31820.

Sinclair ZX Spectrum software

Gloster Software has new software packages for the Sinclair ZX Spectrum.

Gloster Software has been devoted exclusively to the ZX series, providing original software which is both serious and entertaining.

'Display' handles over 300 user-defined characters on screen at the same time. Display and printing can now employ 64 alphas or numerics per line. Fast and smooth animation in BASIC are also possible. (\$14.90)

'Protext' is the text processor

The Gospel according to James

James Martin, the computer industry's most widely read author and cofounder of the DMW Group and of Database Design, will hostaflve-day dataprocessing seminar in Melbourne from August 1-5.

The program will open with a self-contained one-day senior

The Aussie Byte

The Aussie Byte is a new single board computer, totally designed and manufactured in Australia.

It has been designed specifically for the high performance, single and multi-user operating systems such as CP/M 3, MP/M II, CPNET 1.1, TURBODOS etc.

The board incorporates 256K of RAM, four RS-232C serial ports, Z80A CPU at four MHz.

Club Call

If any DGZ-80 computer owners in Queensland are interested in forming a users' club could you please contact Peter Grimes, 6 landra St, Strathpine, Brisbane Old. 4500. For more information phone (07)205-4597.

The Newcastle Microcomputer Club meets on the second and fourth Monday of each month at 7.30 pm in room G12 of the Physics Building at the University of Newcastle.

The club publishes a monthly newsletter. Members own a variety of microcomputers; the club is not devoted to any particular brand or type of microcomputer.

Anyone interested in further information should contact Angus Bliss on (049)67-2433 bh or Tony Nicholson on (049)52-6017 ah.

that sets up the text for 48K only. Up to six A4 pages per text file can be saved and loaded during program operation. (\$24.50)

Gloster Software also provides the means of loading ZX81 saved programs straight into the ZX Spectrum (\$14.90).

'Matcalc', the first of the ZX81 series to be upgraded for the ZX Spectrum, is a spreadsheet calculator. (\$14.90)

All software is on cassette and can be obtained from Gloster Software, GPO Box 5460CC, Melbourne Vic. 3001. (03) 232-2398 after business hours.

management seminar on August 1. The fee for the one-day seminar is \$340; the full five days cost \$1195.

Guest speakers include Dr Vance Gledhill, managing directorof Wicat Computer of Australia, and Dr J. C. Mudge of the CSIRO Division of Computing Research.

For further details, contact Doll Martin Associates, 131 Walker St, North Sydney NSW 2060. (02) 923-2233.

DMA, two Winchester hard disk interface ports to suit WD 1001 and Konan controllers, D/D, D/S floppy disk controller (max eight drives), voice synthesiser, video with 24 lines x 80 characters display and high resolution graphics and light pen input etc. More information about Aussie

Byte (The Great Australian Byte) can be obtained from RDM **Computers Pty Ltd**, 225 High St, Northcote Vic. 3070. (03)481-0136.

How can I write better software, faster? Write it in BASIC/Z!

BASIC/Z. A new standard in compilers for the CP/M system. BASIC/Z is the most powerful implementation of the BASIC language available. BASIC/Z generates executable machine code compatible with 8080. 8085, Z-80 under CP/M 80 and 8086/8088 processors under CP/M 86 and MS-DOS

Syntax testing as you type. BASIC/Z has a powerful program editor with built in syntax testing as you type. Time saving features include global search and replace, fifteen local edit commands and extensive debugging facilities. Line trace, error line retention, and the unique ability to 'single step' a program with a continuous display of selected variables are just a few of the features which will save you time.

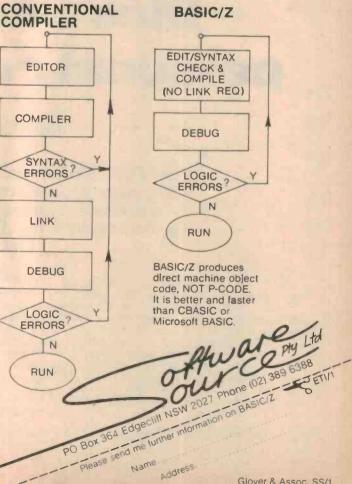
Multitiered error handling allows your program to trap logical errors. including previously fatal BDOS errors. Only BASIC/Z can trap that 'BDOS ERROR ON A: READ ONLY before it happens.

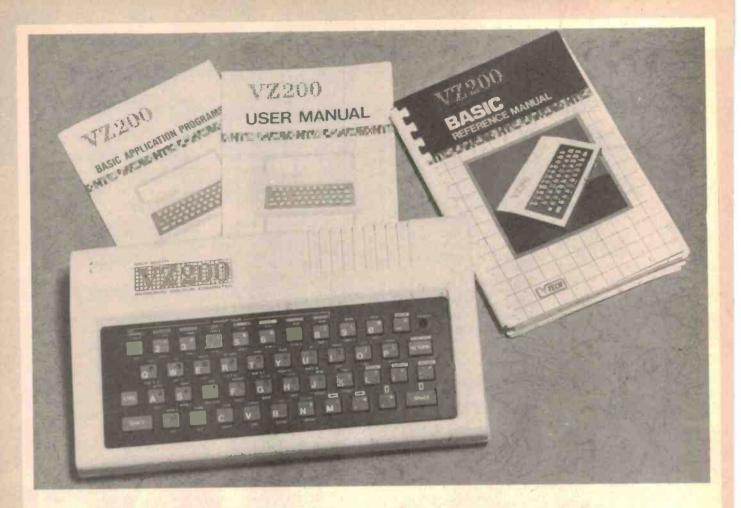
Printer/terminal customizing is built in. The runtime library of BASIC/Z (included in the package) includes installation routines for the majority of CP/M machines on the market. Your software will have near universal application without further modification. Just one set of programs will run on practically any hardware.

Unsurpassed accuracy. Floating point numerics with a range of 1E-61 to 1E+61, with a choice of precision from six to eighteen digits. All floating point maths are performed in decimal (BCD), avoiding rounding off errors. Powerful executive functions aid programming. Using SORT, it can sort 2,000 elements in two seconds. User defined functions are fully recursive, support multiple arguments and may contain an unlimited number of statements

No Royalties. BASIC/Z has no royalties nor runtime charges. The license agreement confers the right to distribute support software such as the BASIC/Z runtime module and the installation hardware configuration utility, subject only to specified copyright acknowledgements

What does it all cost? BASIC/Z documentation & Software: \$495* inc. tax. Available from your computer supplier of from Software Source direct. Available on 21 days app oval (if software seal not broken). Or clip out the coupon and send in for further details.





Dick Smith's VZ200 personal colour computer

Jamye & Roger Harrison

Since Clive Sinclair dropped his ZX80 and 81 'toy' computers on an unsuspecting and unprepared market, there's been a rush, no — a stampede, to expand the features of personal computers and contract the price. The VZ200 currently sits right at the forefront. What does it offer?

THE VZ200 packs an amazing number of features in such a tiny package: 8K bytes of memory (RAM), 16K Microsoft BASIC in ROM, colour graphics — eight colours in medium resolution and four in higher resolution, programmable sound generator with 2½-octave range and nine different note durations, 45-key moving-key keyboard (with auto-repeating keys), both RF output (to TV antenna input) and direct video (for a monitor), inverse video and on-screen cursor-controlled editing. The VZ200 measures just 290 mm wide by 163 mm deep by 50 mm high overall. The keyboard is on the sloping front apron and all the attachments plug into the rear. It is powered from a 9 Vdc plugpack. Along the rear apron are the following connectors: dc input socket, cassette recorder jack, monitor output, expansion connector, peripheral connector and TV (RF modulator) output on channel 36 UHF.

The video display only uses about threequarters of the screen (unlike the picture in the Dick Smith catalogue shows), like many of the colour home computers available. The text format is 32 columns across the screen by 16 lines down. In what they call medium resolution graphics mode you get 64 pixels (blocks) across the screen by 32 down, 128 x 64 (i.e: double) in the 'high resolution' mode.

In the medium resolution mode, you can program a block to be any of eight colours green, yellow, blue, red, buff, cyan (a blue), magenta or orange. They're what's called the 'foreground' colours. The background (i.e: the rest of the screen area) can be either green or orange in this mode.

In the higher resolution mode, you can program any block (foreground) to be any of only four colours — green, yellow, blue or red — with the background colour green, or with the background buff you can program the blocks to be buff, cyan, magenta or orange.

The programmable sound generator has a range of 31 notes over $2\frac{1}{2}$ octaves from A₂ to D#₅, plus a 'rest'. There are nine programmable note durations of 1/8, 1/4, 3/8, 1/2, 3/4, 1, $1\frac{1}{2}$, 2 and 3.

The text character set comprises 62 of the standard 64-character ASCII table, 5 x 7 dot matrix format. The two you don't get are hardly important in this application. Thirty of the keys on the keyboard have four 'shift' levels — as can be seen from the accompanying pictures. With the exception of the RE-TURN, SPACE, CTRL and SHIFT keys, the rest have three levels of shift. That is, apart from obtaining the normal character when you press a key, you can get more functions, such as a graphics character, a BASIC command, an operating command or a program statement.

Four keys act as cursor control keys in the CTRL mode, these being the four on the right of the lower rank. The L and ';' keys provide the INSERT and RUBOUT editing functions in the CTRL mode. The colour programming command keys, 1 to 8, are labelled and colour-coded.

The expansion connector will accommodate such things as a memory expansion module. A 16K module is available for just \$79, allowing expansion of the user memory to 24K.

The peripheral connector is for plugging in such things as a printer interface, and one is available for \$49.50, permitting the attachment of a standard Centronics printer, many models being widely available — and the prices are continually coming down.

The VZ200 is supplied with all cables in generous lengths, a plugpack, a User Manual, a demonstration program on cassette, a BASIC Reference Manual and a booklet of BASIC Applications Programs.

From the user's view

For all the functions packed into the keyboard, the key operation is a big let-down. The keys are rubber-buttoned microswitches and while they do have movement, the feedback via your finger can only be described as uncertain.

We've criticised this type of keyboard in the past and can't help but think that, where a cost compromise is necessary, an elastomeric keyboard (like that on the ZX81) is preferable. The computer gives a 'beep' when you press a key (except for the CTRL, SHIFT and RETURN keys), which helps, but the key action is so light that double-keying is common. The auto-repeat feature, however, is a good idea. The key will repeat the character or command if you hold it down for longer than one second.

The on-screen editing functions are very good — a real boon to the beginner programmer. The usual BASIC editing feature of simply retyping a crook line works, but that can be time-consuming, especially with long lines. The VZ200 allows you to move the cursor around and re-type incorrectly entered characters, commands or statements. With the latter two, the single-key entry feature is a real time-saver. We would rate the editing facilities as one of the VZ200's major features.

The keyboard has an enlarged SPACE key at the right of the lower rank. This is a problem if you're used to a normal typewriterstyle keyboard as you keep cracking your finger on the case below the keyboard! It takes a little getting used to. We also took a little time to learn not to confuse the SHIFT and CTRL keys. There are other problems with the keyboard that relate to its partly non-standard layout, but if you're a beginner in the personal computer stakes it's unlikely to be a worry.

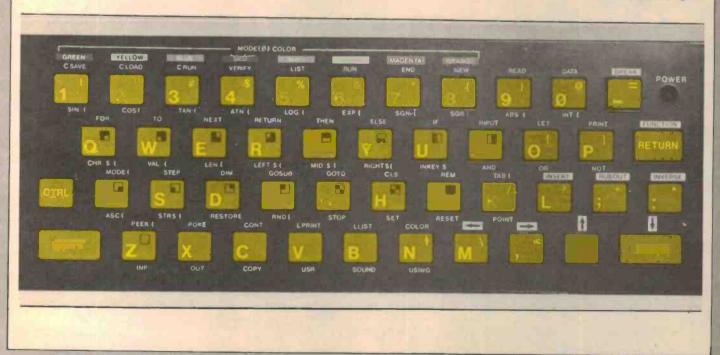
The single-key entering of statements and commands was an idea introduced by Clive Sinclair with his ZX80, forerunner to the ZX81 and Spectrum computers. It's a good idea, taken to its logical limit with the VZ200. Strictly, you need to use more than one key to enter a command, statement or graphics character, but only three at the most; e.g: to get the PRINT command you push CTRL and P together. To get the command or statement under a key, you hold down CTRL and press RETURN, then the key you want.

The direct video output into a Philips 20" colour monitor is good, but plagued by patterning that ripples seemingly diagonally across the display. The display is noticeably inferior when using the RF output into the TV set's antenna. However, it is better than some other popular colour computers around. For the price, it's acceptable.

The VZ200 uses a Z80 microprocessor, probably the most widely used microprocessor in all the personal computers produced to date. The specifications say it runs at 3.58 MHz. However, it's not all that fast, but is probably quite fast enough to manipulate simple graphics effectively.

If you really want to know, a FOR-NEXT loop takes four milliseconds, which in today's computer world is pretty slow. As it really is a beginners' machine, that's no real disadvantage. If you're thinking of ploughing through your maths homework with it, a pocket scientific calculator is faster.

Continued on page 37



The military coup



of Lieutenant Wilson. Aged 20.

When you talk to Geoff Wilson about his 18 months in the Army you can't help but feel he's a young man who's come a long way in a short time.

Already he's graduated from Officer Cadet School, Portsea, with a commission as a Lieutenant. And already he commands an Infantry Platoon of thirty men.

It's a tremendous responsibility. Especially for a 20 year old. But as Geoff says, "You get enormous satisfaction from using what you've learnt in your training to help bring out the best in your team. And in achieving results together."

"Obviously situations which I haven't had experience in crop up all the time. And that's when I have a real responsibility to myself and my men to make decisions based on sound knowledge rather than guesswork. As a result you constantly find yourself learning and mastering new skills. And that's a challenge I always enjoy accepting. You really begin to realise your full potential and believe in your ability."

"I suppose you might say my decision to take on a career as an Army Officer is my personal 'military coup'."

"It's hard to visualise another career as exhilarating, varied and satisfying. Or which provides as many opportunities to be recognised and rewarded for a job well done."

"Think about it carefully. I know I did before joining!"

Armour, Artillery, Engineering, Survey, Signals, Transport, Infantry, Intelligence and Aviation are just some of the fields open for you to enter.

If you're aged between 181/2 and 23 on entry (or up to 25 with a degree or diploma), have your HSC or equivalent, (at a level acceptable to the Army), and think you have

what it takes to pull-off your own military coup, contact your nearest Army Careers Recruiting Centre or fill in the supplied coupon.

There are two courses per year: Applications close mid-March for a July entry and early August for a January entry.

Authorised by Director-General Recruiting, Department of Defence.

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SUMMARY OF BASIC COMMANDS

Fu	nctions:
1)	Arithmetic operators
	+,-,*,/,*
2)	Relational operators >, <, =, > =, < =, <>
3)	Arithmetic functions:
	SQR – Square root INT – Integer part RND – Random number
	ABS – Absolute magnitude SGN – Sign
	COS – Cosine SIN – Sine EXP – e*
	TAN - Tangent LOG - Natural logarithm
	ATN - Arc tangent
4)	String functions:
	LEN - Length STRS - String of numeric argument
	VAL - Numeric value of string
	ASC – ASCII value CHRS – Character
	LEFTS - Left characters MIDS - Middle characters

Documentation

The BASIC Reference Manual and the two booklets supplied with the VZ200 are generally well produced, clear and understandable — which is just what the raw beginner wants.

The BASIC Reference Manual is spiral bound, which facilitates laying it open so the pages sit flat. However, the spiral binding is just slightly too small for the number of pages and it's a bit of a bind trying to turn them.

This manual covers all the functions and operations of the VZ200 in a fundamental way, with some programming examples. You are encouraged to learn by trying things for yourself. We found a number of small errors, but nothing disastrous.

For example, the method of using the IN-SERT command when editing does not work the way it's described in the book. Say you typed PRIT instead of PRINT. The book says you do an INSERT by moving the cursor up to the character before the place you want to insert a character (that is, 'I' here), type CTRL INSERT, then type the required character (that is, 'N' here). However, that gives you PRNIT!

What you really have to do is cursor up to the character *after* the place where you need to insert a character, then do the insert routine.

The reference manual lists all the available text characters and BASIC statements, or chors and commands, with some brief e planations. An error message list is given, but incredibly, no explanation of what they all mean or what to do when you get one! Grrr.

For all its good points, the manual contains no *detailed* index, which would be very useful for a beginner. The contents list is at least comprehensive, so that's a plus in its favour.

What happens when you've worked your way through the reference manual? Well, you won't be a hot-shot programmer, but you will have gained an understanding of programming and be able to tackle some programs of your own invention, plus modifications to published software.

_	
	RIGHTS - Right characters INKEYS - Check keyboard
5)	Logical operators
	AND OR Relation and logical expressions have value 1 If true, NOT 0 if false.
6)	Graphics and sound functions
	CLS - Clear screen SET - Plot a point RESET - Clear a point POINT - Return the color code COLOR - Set color SOUND - Produce tone of different frequency and duration MODE - Select graphic or text
7)	Program statements
	DIM - Dimensions STOP END GOTO GOSUB RETURN FORTOSTEP NEXT
	REM IF THEN ELSE
	INPUT

As Microsoft BASIC is used — the erstwhile 'industry standard' — there are huge amounts of published programs and many, many books on the subject that will keep you occupied for ages.

A booklet of applications programs is included with several dozen short programs that are not only interesting and amusing, but instructive and perhaps useful to boot. Many would be good 'starting points' for developing programs of your own devising or useful as subroutines within your own programs.

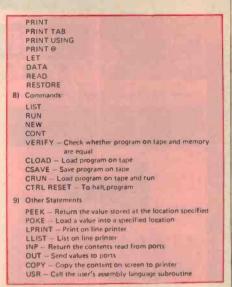
Absolutely no technical details, not even a memory map, are given, but we guess that such things might appear in some 'support' publications.

The BASIC

The 16K Microsoft BASIC included can only be described as excellent — outshining the mechanical and electronic constraints of the VZ200. But, we have to keep reminding ourselves that this is really a low cost beginners' machine. The range of commands, etc, available, and the flexibility of the language, stand out. Learning to use the facilities is a breeze. The buzzword is 'user friendly'!

All the BASIC commands, operators and statements are shown in the accompanying panel. Those of you who know will see that it's all pretty standard fare. However, it's good to see the inclusion of such things as IF...THEN...ELSE statements and the COPY statement (otherwise known as a 'screen dump'). Seeing that USR is included for the benefit of using machine code in BASIC programs, we can only hope that some suitable books or manuals on the subject, specifically for the VZ200, will appear at some later date.

Programming using graphics or sound is relatively simple. The graphics commands are simple, largely because of the 'chunky' graphics employed. You'll find no DRAW, PAINT, LINE or CIRCLE commands here, but what you do get is effective for the sort of graphics included in the machine. It's best to crawl before you walk, and it's a beginners' machine, remember. Similar sentiments apply to the sound programming.



Cassette comments

A pre-recorded cassette with cute demonstration software comes with the VZ200. For one thing, it shows that the cassette interface is quite good, as reliable loading was no problem.

As the VZ200 is not a games/computer machine, the pre-recorded software base is only going to be available on cassette, as there's no ROM socket. At present, there's no pre-recorded software available, but, from past experience, that's probably a situation that will rectify itself.

There are lots of 'freelance' software producers in the market supplying software for existing machines who will doubtless get behind the VZ200.

Conclusion

The VZ200 is very reminiscent of the Sinclair ZX81/Spectrum or National JR100 (which is sort of rare here, as yet). It has a very great deal to offer in price, functions and features. The major disappointment is the keyboard, but all low cost home computers compromise here and it's a matter of preference whether you favour one type of cheap keyboard over another.

The big question is, would you do any better at \$299. You'd almost certainly get a better keyboard, but we haven't yet seen anything in that price range to compete with the features and memory capacity of the VZ200.

Judging from the phenomenal success and popularity of other 'bottom end of the market' computers, such as the ZX81, Spectrum and VIC-20, there are huge numbers of people who want a low cost computer just to 'get started', or get their children started, in computing.

Price is all-important to people who don't want to pay a great deal of money to learn what the subject's all about before 'getting in deeper'. Compromises are acceptable therefore, and our criticisms should not be taken too much to heart. For its price, the VZ200 has a great deal to offer, and from such small beginnings one can go on to 'conquer the world', or at least a comfortable niche.

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Machine language graphic driver for the ETI-640/ETI-685 Part 2

In the first part we whetted your appetite telling you what was in the graphic driver package. Now you've read (almost) all about it, here's the real thing! ... from two of the 'slacker hackers'.

G.H. Secomb J.F. Adamthwaite

Rectangle function

This routine draws a rectangle, using the two pairs of user-supplied coordinates as opposite corners. Conceptually quite straightforward, it constructs the coordinates for the other two corners from combinations of the supplied coordinates, and plots a line from corner to corner until the full rectangle is drawn, using the appropriate pixel plotting mode of SET, CLEAR or INVERT.

The requirements for the rectangle functions were:

1. Double plotting of the corner pixels must not occur. Each corner is the end point of one line and the start point of the next. Double plotting during the INVERT RECTANGLE function would leave the corners as they were, effectively remaining unplotted. To get around this, each line must be shortened by one point (see diagram below).

2. The user must be able to specify any of the diagonally opposing coordinates, as for the window function, and achieve the same rectangle on the screen.

 $\begin{array}{c|cccc} X1,Y2 & X2,Y2 \\ X1,Y1 & X2,Y1 \\ \hline (X1,Y1)(X2,Y2) \\ (X1,Y2)(X2,Y1) \\ (X2,Y1)(X1,Y2) \\ (X2,Y2)(X1,Y2) \\ \hline (X2,Y2)(X1,Y2) \\ \hline \end{array}$

These combinations should all give the same results. Don't call us if they don't.

The line shortening and flexible coordinate specification capability are shown in the algorithm below.

Ellipse function

The brain-strain required to come up with a useful ellipse plotting routine caused nearly as many hairs to desert their post as did the LINE routine. Several techniques were tried before the final one was selected. They were:

1. Calculating sine and cosine for progressively increasing angles from 0 through to 359 degrees. This approach proved to be slow and took too much code. (Machine language trig. is a pain.)

2. Using look-up tables for sines and cosines. This was fast, but required a large look-up table, even when using a single 90 degree quadrant table backwards and forwards to create the sines and cosines for each of the four quadrants.

3. Using coarse angle increment look-up tables with interconnecting lines to fill the gaps.

This seems to be the method employed in the TRS-80 colour computer, although we are not sure of that. It was fast, but the code required to avoid double plotting at the line junctions proved to be too cumbersome. This method requires greater resolution than is possible on the ETI-640, in order to look convincing. 4. Using a version of the Variable Duty-Cycle Algorithm.

This method held promise of being the fastest of all, but it suffered from a precision problem due to the use of integer arithmetic, resulting in a noticeable distortion of the circle. It also didn't seem to know when to stop. We only wanted one lap.

Finally, a chance encounter with two bright young lads (these kids make us oldies feel so stupid) at an establishment known as 'Comp-Soft' in Swan Street in Richmond, brought to light a workable solution. I will be forever grateful to the proprietor for allowing me to waste his time and that of his workers. It saved us from needing to weave baskets.

The method has its roots in calculus. Don't cringe, we've already done it for you. It is quite novel in the method it uses to create the sine and cosine values. An explanation follows:

Looking at a sine curve or trig. table book, you will notice that at 0 degrees, sine = 0, rising rapidly with increasing angles at first, then progressively slowing in its ascent till 90 degrees is reached where the sine value is momentarily stationary, with a maximum value of ± 1 .

The 'secret' is in deriving a progressively decreasing value from somewhere to be added to the old sine to create the new one. Where can we get such a decreasing value from?

The answer is . . . trumpet fan-fare . . . the cosine!

<pre>* RECTANGLE FUNCTION * Input parameters are- (X1,Y1) coords of any corner * (X2,Y2) coords of opposite corner</pre>	SET X = X2 FOR Y = Y1 TO (Y2-YINCREMENT) STEP YINCREMENT PLOT(X,Y) NEXT Y
IF X2) X1 THEN XINCREMENT = +1.	SET Y = Y2
IF X2 = X1 THEN XINCREMENT = 0.	FOR X = X2 TO (X1+XINCREMENT) STEP (-XINCREMENT)
IF X2 (X1 THEN XINCREMENT = -1.	PLOT(X,Y)
IF Y2) Y1 THEN YINCREMENT = +1.	NEXT X
IF Y2 = Y1 THEN YINCREMENT = 0.	SET X = X2
IF Y2 (Y1 THEN YINCREMENT = -1.	FOR Y = Y2 TO (Y1+YINCREMENT) STEP (-YINCREMENT)
SET Y = Y1	PLOT(X,Y)
FOR X = X1 TO (X2-XINCREMENT) STEP XINCREMENT	NEXT Y
PLOT(X, Y) NEXT X	END.

Going back to our trig. table again it will be found that the cosine has a maximum value of ± 1.0 at 0 degrees, progressively decreasing towards a value of zero as 90 degrees is approached. If a small portion of the current cosine value is added to the current sine value, it will yield the *new* sine value.

Fine. So where do we get the new cosine value from?

Notice that the *amount* by which the cosine decreases is progressively increasing as 90 degrees is approached. (Yes, it is a mouthful.) In other words, an increasing value is being subtracted from the cosine at each step. Where can we find a number which just happens to be increasing?

The source of a suitable increasing value is found to be the sine (surprise, surprise).

We therefore take a small portion of the sine value and subtract it from the old cosine to give the new cosine value.

The pleasant feature of this approach is that the method follows through into the second, third and fourth quadrants without the need for any further trickery.

Further subtractions from the cosine value after 90 degrees is reached make it swing negative, and further additions of a portion of this negative cosine to the sine value make the sine start dropping from +1 towards zero, and so on. In this way, the values of the sine and cosine leap-frog right around the circle, eliminating the need for traditional calculations based on a known angle. The information required for the next point is extracted from the previous point values. Wake up!

Before tackling the final algorithm chosen for circle and ellipse plotting, a simplified version will be examined. This version is a 'stripped' example of the 'bells and whistles' model. It has the advantage of significantly greater speed due to the lack of multiplication operations, sacrificing the ability to draw ellipses. If only circles were needed,' and speed was vital for a particular application, it could be used in place of, or better still, in addition to the ellipse function.

It is slightly different to the previous description in that the variables SINE and COSINE, instead of containing values varying between +1 and -1, contain values varying between +RADIUS and -RADIUS. The concept is the same, just the magnitude of the variables has been changed to gain speed.

You will notice spontaneously, if nudged, that the divisions can be done in machine language as simple right shifts, as long as the divisor, in this case PORTION, is a simple power of 2. e.g. 32, 64, 128 etc.

Increasing values for PORTION give more plotted points along the perimeter of the circle, allowing circles of larger radius to be plotted without gaps. This also causes the values of SINE and COSINE to be altered in smaller increments, more closely approximating a true differentiation function, resulting in a rounder circle at the same time. Very helpful if circles are round, it is. Saves overworking your imagination. Sadly, you must sacrifice speed if you want large circles or high precision. Try experimenting.

The final value of PORTION chosen for the machine language version is 128 (2 to the power of 7). This allows circles with a radius of up to approximately 140 to be plotted without gaps. (That's huge.) If your application doesn't require such large arcs, a doubling of speed can be obtained by selecting a value for PORTION of 64. A value of 32 would further double the speed but the maximum radius of 35 (now) is starting to become restrictive, and the circle gets a little slanted to one side. There is no correct value, merely values of greater or lesser convenience.

We chose a value of 128 to keep the precision high. It gives the maximum usable precision for the ETI-640 and could be regarded as the practical upper limit here. At a radius of 140, the circle has a diameter of 280 and wraps around the screen due to byte overflow. It is a simple matter to alter the precision if desired.

The final algorithm presented now is identical to the machine code version. The penalty for ellipse plotting capability is the need for two multiply operations per loop, which slows it considerably. Unfortunately we must do true multiplies, not just left shifts.

In this version 'correct' values for sine and cosine are kept in the variables SINE and COSINE with values between +1 and -1. These values are multiplied by the vertical and horizontal radius and added to the ellipse centre coordinates to give the final plot point coordinates. Remember, this happens some 804 times per circle if PORTION is set at 128.

A comparison to the previously plotted point is performed to ensure freedom from double plotting. Some twiddling to the loop counter eased testing for the exit condition.

* ELLIPSE FUNCTION
* Input variables are - XCENTRE X coord of ellipse centre
- YCENTRE Y coord of ellipse centre
+ - HRADIUS Horizontal radius + - VRADIUS Vertical radius
- VRHDIUS Vertical radius
LET SINE = 0.
LET COSINE = 1.0
LET PORTION = 12B
LET XPLOT = XCENTRE + HRADIUS (Plot first point)
LET YPLOT = YCENTRE
PLOT (XPLOT, YPLOT)
FOR COUNTER = 0 TO 3. 14159 STEP 1/256
SINE = SINE + COSINE/PORTION
COSINE = COSINE - SINE/PORTION
NEWXPLOT = INTEGER (COSINE + HRADIUS + XCENTRE + 0.5)
NEWYPLOT = INTEGER (SINE * VRADIUS + YCENTRE + 0.5)
IF NEWXPLOT = XPLOT
AND IF NEWYPLOT = YPLOT
THEN GOTO NEXT-COUNTER (Same as last, skip plot)
XPLDT = NEWXPLOT YPLOT = NEWYPLOT
PLOT(XPLOT, YPLOT)
NEXT COUNTER
END.

CIRCLE GENERATOR ALGORITHM XCENTRE X coord of circle centre YCENTRE Y coord of circle centre -- RADIUS Desired circle radius LET SINE = Ø. LET COSINE = RADIUS LET PORTION - 128 (Speed/accuracy trade-off value) LET XPLOT = XCENTRE + RADIUS LET YPLOT = YCENTRE PLOT (XPLOT, YPLOT) (Plot first point on circle) FOR COUNTER = 1 TO 804.2477 (= 2 + PI + PORTION) SINE = SINE + COSINE/PORTION (Calc new sine) COSINE = COSINE - SINE/PORTION (Calc new cosine) XPLOT = INTEGER (XCENTRE + COSINE + 0.5) (Round result) YPLOT = INTEGER (YCENTRE + + 0.5) SINE PLOT (XPLOT, YPLOT) (and plot (t) NEXT COUNTER (Back for more?) END.

Software interfacing

The program was designed to make interfacing as easy as possible (we hate phone calls) while allowing for future expansion in the number of supported functions, or changes in the number of parameters passed to it.

The only absolute address you need to remember is the subroutine entry point 'ENTRY' at hex 6400. All the parameters which need to be communicated to or from the program are stored in a 6-byte table starting at 'ENTRY+3', hex 6403 through to 6408, inclusive. The number of parameters required to be passed depends on the desired function. See tables opposite.

Speculation

This package is by no means complete, and it can never be so, due to the diversity of requirements in different applications. It is

HEX DECIMAL	
640025600ENTRYEnter here after storing parameter640325603FUNCTIONGraphic op-code store.640425604PARAM1Parameter list.640525605PARAM2"640625606PARAM3"640725507PARAM4"640825608BITVALUEResult of "BIT TEST" function call.	5,
FUNCTION PARAMS REQ'D PARAMETER DESCRIPTION	
0 SET WINDOW 1,2,3,4 - Parameters 1 & 2 define one corr of the desired area. 1 CLR WINDOW 1,2,3,4 - Params 3 & 4 define the diasonal	ner
2 INV WINDOW 1, 2, 3, 4 - As above. 3 PROT SCRN 1 - Parameter 1 is the number of ten lines to be protected from scrol	llins.
4 SET PIXEL 1.2 - Parameter 1 is the X coordinate. 5 CLR PIXEL 1.2 - Parameter 2 is the Y coordinate. 5 INV PIXEL 1.2 - As above. 7 TST PIXEL 1.2 - Pixel status returned in location "BITVAL".	on
8 SET LINE 1,2,3,4 - Parameters 1 & 2 define the star point of the line. 9 CLR LINE 1,2,3,4 - Parameters 3 & 4 define the end. 10 INV LINE 1,2,3,4 - As above. 11 N.U.	rt.
12 SET RECTNGL 1,2,3,4 - Same parameter definitions as th 13 CLR RECTNGL 1,2,3,4 WINDOW functions. 14 INV RECTNGL 1,2,3,4 - As above. 15 N.U.	he
16 SET ELLIPSE1, 2, 3, 4- Params 1 & 2 define the X & Y control17 CLR ELLIPSE1, 2, 3, 4of the centre of the ellipse.18 INV ELLIPSE1, 2, 3, 4- Param 3 is the horizontal radiu19 N.U Param 4 is the vertical radius.	oords

designed to be easily extendable and as fast as possible. Computer graphics can never be too fast.

If you have a different memory-mapped VDU, don't despair, as only the subroutine 'CONVRT' need be changed to perform the required coordinate translation. If your VDU has a greater coordinate addressing range, the use of 16-bit arithmetic will be necessary. Those of you with Z80/ETI-640 combinations are invited to try your hand at rewriting this program.

Hardware tricks to consider are preprogrammed ROMs on I/O ports to speed up coordinate-to-address conversion or for use as look-up tables for trig. or multiply/divide functions.

We chose not to do this here, as it was judged to be out of the range of the average part-time hobbyist.

The user should try to adapt and use the program according to the application needs. It works quite well as a Napoleon hat or a fleet of paper darts.

Reader application examples are invited, as we are burnt out. We are currently threatening upgrades to the program such as macro function capability, relative coordinate addressing and string processing functions. These will most likely be added as a 'front-end' program, which preprocesses the extended functions and feeds the derived parameters to this one.

A 6809 (wot's dat) version is currently in the pipeline, and if this article makes us rich and famous, it may get some airplay as well.



Z80 S-100 BUS MICROPROCESSOR

SME is proud to present the flagship of its Unicorn series - the MPU-100.

Designed for use in commercial, industrial and engineering environments where speed and reliability are paramount, the MPU-100 features an advanced vertical motherboard system that gives it the highest reliability and lowest profile of any machine of its type.

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FEATURES

- Z80, 4Mhz, SBC 800.
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- 10 slot vertical motherboard for reliability and ease of expansion.
- Designed and manufactured in Australia.

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EXERCISE PROGRAM - 640/685 graphic driver

1000 REM ****** GRFXTEST.BAS ****** 1010 GOSUB 9980 : GOTO 1200 1020 REM Sample POINT operation driver. PDKE(F7,F) PDKE(X5,X): PDKE(Y5,Y) REM Set function 4,5,5,7 PDKE(X5,X): PDKE(Y5,Y) REM Set PDINT coords CALL(G7) REM Plot it 1030 1050 CALL(G7) 1060 RETURN 1070 REM Sample LINE & RECTANGLE driver
 1070
 REM Sample LINE & RECTANGLE driver

 1080
 POKE(F7,F)
 REM Function 0.1.2 8.9.10 12.13.14

 1090
 POKE(X5,X1):POKE(Y5,Y1)
 REM Set start coords

 1100
 POKE(X7,X2):POKE(Y7,Y2)
 REM Set end coords

 1110
 CALL(G7)
 REM Draw it
 11100 RETURN 11200 REM Sample ELLIPSE driver 11400 POKE(F7.F) REM Set function 16,17,18 11500 POKE(F7.F) REM Ellipse position 11600 POKE(X7.W) : POKE(Y7.H) REM Ellipse dimensions 11600 POKE(X7.W) : POKE(Y7.H) REM Ellipse dimensions 11700 CALL(G7) REM Draw it 1170 RETURN 1200 PRINT ******* GRFXTEST. BAS ******** * PRINT 1320 GOSUB 1080 1330 NEXT Y2 1340 F=9 # X1=0 # Y1=0 : Y2=63 : FOR X2≈126 TO 0 STEP -1 1350 GOSUB 1080 1360 NEXT X2 1380 REM Set alternate Horiz. Lines from top to bottom. 1390 F=S: X1=0: X2=127 : FOR Y1=63 TO 0 STEP -21400 Y2=Y1 : GOSUB 1080 1410 NEXT VI 1410 NEAT VI 1420 REM INVERT alternate vert. lines from right to left. 1430 F=10 : Y1=0 : Y2=63 : FOR X1=127 TO 0 STEP -2 1440 X2=X1 : GOSUB 1080 1450 NEXT X1 1460 FOR D=1 TD 500 : NEXT D 1480 REM Clear screen 1490 POKE(F7.3) : POKE(X6.0) : CALL(G7) 1510 REM Random horizontal & vertical lines. 1520 REM Do 1 horiz. & 1 vert. line per loop. 1540 F00 2-1 TO 10 1450 NEXT X1 1540 FOR I=1 TO 40 1550 X1=INT(RND(0)+128) 1550 Y1=INT(RND(0)*128) 1550 Y1=INT(RND(0)*64) 1570 X2=INT(RND(0)*128) 1580 Y2=Y1 1590 GOSUB 1080 1600 X1=INT(RND(0)*128) 1610 Y1=INT(RND(0)*64) 1620 X2=X1 1620 X2=X1 1630 Y2=INT(RND(0)*64) 1640 GOSUB 1080 1650 NEXT I 1660 FOR D=1 TO 500 : NEXT D 1680 REM Invert state of entire screen twice 1690 F=2 1700 X1=0 : Y1=0 + X2=127 : Y2=63 1700 X1=0 : Y1=0 : X2=127 : Y2=63 1710 GOSUB 1080 1720 FOR D=1 TO 500 : NEXT D 1740 X1=127 : Y1=63 : X2=0 : Y2=0 1750 GOSUB 1080 1790 REM Sceen clear, rotating from top corner. 1800 F=9 : X1=127 : Y1=63 : Y2=0 : FOR X2=127 TO 0 STEP -1 1810 GOSUB 1080 1920 NEVT Y2 1820 NEXT X2 1830 F=9 : X1=127 : Y1=63 & X2=0 & FOR Y2=1 TO 63 1840 GOSUB 1080 1850 NEXT 1860 REM Random angled lines, joined end-to-end. 1892 F=8 1900 FOR I=1 TO 4
 1900
 FUN
 I=1
 IO
 4

 1910
 FOR
 L=1
 TO
 3

 1920
 X1=X2
 Y1=Y2
 1930
 X2=INT(RND(0)*128)

 1940
 Y2=INT(RND(0)*64)
 1940
 Y2=INT(RND(0)*64)
 1950 GOSUB 1080 1960 NEXT L 1970 FOR D=1 TO 500 : NEXT D 1980 NEXT I 1970 1980 NEXT I 2000 FOR D=1 TO 500 : NEXT D 2020 REM Clear Screen, 0 lines protected 2030 POKE(F7.3) : POKE(X5.0) : CALL(G7) 2050 PRINT " This may be a sood opportunity to adjust" 2060 PRINT " Thi 2150 W=C*5 : H=W GOSUB 1140 2170 NEXT C 2200 REM Draw another circle to produce Moire fringe 2210 X=46 2220 FOR C=0 TO 15 2230 W=C*5 : H=W 2240 GOSUB 1:40 2250 NEXT C 2280 REM Clear screen 2290 POKE(F7.3) : POKE(X5.0) : CALL(G7) 2310 REM Draw random Ellipses. 2320 F=16

2330 FOR I=1 TO 3 2340 FOR C=1 TO 10 X=INT(RND(0)*128) Y=INT(RND(0)*64) 2350 2360 W=INT(RND(0)*100) H=INT(RND(0)*85) 2370 2380 GOSUB 1140 NEXT C FOR D=1 TO 500 : NEXT D POKE(F,3) : POKE(X6,0) : CALL(G7) 2390 2400 2410 2430 NEXT I 2450 REM Use 'INVERT LINE' function to create Moire Fringe. 2450 FE10 2460 FE10 2470 FOR I=1 TD 2 2480 X1=0 : Y1=0 | X2=127 2490 FOR V2=0 TD 63 2500 GOSUB 1080 2510 NEXT Y2 X1=0 : Y1=0 : Y2=63 253Ø 254Ø FOR X2=126 TO Ø STEP -1 GOSUB 1060 NEXT X2 X1=0 : Y1=63 : Y2=0 2550 2560 2570 FOR X2=0 TO 127 2580 GOSUB 1080 NEXT X2 X1=0 : Y1=63 : X2=127 FOR Y2=1 TO 63 GOSUB 1060 2590 2600 2510 2620 2630 NEXT Y2 2640 NEXT I 2660 REM Draw random rectangles using 'INVERT RECTANGLE' 2670 F=14 2690 FOR T=1 TO 3 2590 FOR R=1 TO 10 2700 X1=INT(RND(0)+128) 2710 Y1=INT(RND(0)+64) X2=INT(RND(0)+128) 2730 Y2=INT(RND(0)*64) 2740 GOSUB 1080 PEXT R FOR D=1 TO 500 : NEXT D PORE(F7,3) : POKE(X5,0) : CALL(G7) 2750 2760 2770 POKE (1 2780 NEXT I 2800 REM Use 'INVERT WINDOW' function to draw random blocks. 2810 F=2 2810 F=2 2830 FOR I=1 TO 4 2830 POKE(F7,3) : POKE(X5,0) : CALL(G7) 2840 FOR R=1 TO 8 2850 X1=INT(RND(0)*128) 2850 V1=INT(RND(0)*128) Y1=INT(RND(0)*64) X2=INT(RND(0)*128) 2860 2870 2680 Y2=INT(RND(0)+64) 2890 GOSUB 1080 NEXT R 2900 2910 FOR D=1 TO 500 : NEXT D. 2920 NEXT I 2940 REM Invert whole screen with 'INV WINDOW' function. 2950 F=2 2960 X1=0 : Y1=0 : X2=127 + Y2=63 2970 GOSUB 1080 2980 FOR D=1 TO 500 : NEXT D 2990 X1=127 : Y1=63 : X2=0 : Y2=0 3000 GOSUB 1080 3010 FOR D=1 TO 500 : NEXT D 3060 PRINT 3050 PRINT " The display will show any pattern sensitivity in" 3050 PRINT " your BASIC random number generator function, if" 3050 PRINT " left running for several minutes." 3100 PRINT SIDD PRINT " This is the last example. It is a "near-infinite"" 3120 PRINT " loop. Press 'BREAK' when you wish to stop." 3130 PRINT : PRINT 3140 FOR D=1 TO 1000 : NEXT D 3160 REM Avoid subroutines here to gain speed. 3170 POKE(F7,4) 3120 FOR I=1 TO 1E4 3190 PUK 1#1 10 1E4 3190 PUKE(X6, INT(RND(0)*(28)) 3200 PUKE(X6, INT(RND(0)*64)) 3210 CALL(G7) 3220 NEXT I 3240 PRINT : PRINT " My, we are patient, aren't we." 3240 PRINT I PRINT " My, we are patient, aren't we 3250 END 3260 REM 9980 REM Standard GRFXPAK Parameter Definition Module 9981 G7=25600 REM srfxpak s/r entry address 9982 F7=57 + 3 REM srfx function code store 9983 X6=F7 + 1 REM X1 co-ordinate store 9985 X7=X76 + 1 REM Y1 " " 9985 Y7=X76 + 1 REM Y2 " " 9985 Y7=X7 + 1 REM Y2 " " 9987 T7=Y7 + 1 REM Y2 " " 9987 RT-Y7 + 1 REM Y2 " " 9988 RC Chear Screen (text mode), protect 0 lines 9989 POKE(F7,3) : POKE(X6,0) : CALL(G7) : RETURN 9990 REM 0. SET WINDOW (A)PARAMS 10, INVERT LINE 9991 REM 1. CLEAR " (4) 11.-9992 REM 2. INVERT " (4) 12. SET RECTANGLE 1. CLEAR (4) 2. INVERT (4) 3. PROT/CLEAR (1) 4. SET POINT (2) 5. CLEAR (2) 6. TANGER (2) 14) 9992 REM 9993 REM 12. SET RECTANGLE (4) 13. CLEAR (4) 14. INVERT (4) 9994 REM 5. CLEAR " 6. INVERT " 7. TEST " 8. SET LINE 9. CLEAR " (2) 15. (2) (2) 9996 REM 16. SET ELLIPSE 17. CLEAR 18. INVERT (4) (4) 9998 REM 9999 REM (4) (4) (4) 19.

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SOURCE LISTING -

	SOURCE LISTING -	6450 3F6534	IWINDOW BSTA, UN DIRTST WORK OUT STEPPING DIRECTIONS IWLOOP LODA, R3 INDEX FETCH INDEX TO POINT HANDLER
	640/685 graphic driver	6463 BF642B	BSXA FNTBL+0C.R3 PLOT POINT
	GRFXPAK	6466 0D04DC 6469 0E04DD 646C 0402 646E 93	LODA,R1 XPLOT RECOVER COORDS LODA,R2 YPLOT LODI,R0 LCOM LOG. COMP, NO CAR LPSL
	** PERFORMS ONE OF SEVERAL GRFX OPERATIONS. ** THE ACTUAL OP. PERFORMED IS SELECTED BY A CODE BYTE ** POKED INTO LOCATION "FUNCTN" BY THE CALLING PROGRAM.	646F EE0407 6472 1805 6474 8E05B1 6477 1867	COMA.R2 Y2 FINISHED VERT LINE VET? BCTR.EQ WXTST ADDA.R2 YINCR NO. STEP VERTICALLY & CONT. BCTR.UN WLOOP
	:* FUNCTIONS-+ :* GENERAL SCREEN OPERATIONS	6477 1007 6479 ED0406 647C 14 6470 BD0580	IWXTST COMA,R1 X2 FINISHED WHOLE WINDOW? RETC.ED YEB.RETURN TO CALLING PROGRAM. J ADDA,R1 XINCR NO. MORE TO DO
	10 00. SET WINDOW WITH OPP. CORNERS (X1,Y1) & (X2,Y2) 10 01. CLEAR WINDOW 10 02. INVERT WINDOW 10 03. SCROLL PROT. (X1) LINES, CLR SCREEN TO TEXT MODE	6480 0E0405 6483 1858	LDDA. R2 Y1 RESET Y TO Y1 BCTR. UN WLOOP Screén clear and protect
	<pre>** GINGLE POINT OPS ** @4. SET (TURN ON) SCREEN POINT @ X1.Y1 ** @5. CLEAR (TURN OFF) " ** @6. INVERT (REVERSE) " ************************************</pre>	5485 0C0404 6488 440F 6488 C888	PROTOT LODA, R0 X1 FETCH SCRL PROTECT VALUE ANDI. R0 OF TRUNCATE ANY SILLY VALUES STRR, R0 +PROT
	<pre>* 05. INVER! (AEVERSEAN POINT (X1.Y1) AND LEAVE A 00 * 07. TEST STATEOP SCAN POINT (X1.Y1) AND LEAVE A 00 * (IF BIT OFF) DR "FF" VALUE AT LOC. "BITVAL". * 08. INVERT A LINE FROM (X1.Y1) TO (X2.Y2) * 08. DRAW (SET) A LINE FROM (X1.Y1) TO (X2.Y2)</pre>	648C 0420 648E 0700 6490 CF7800 6493 CF7900	LODI, RØ M'20 SPACE CODE LODI, R3 Ø INDEX INT2 STAA, RØ VDU, R3 CLEAR SCREËN J STAA, RØ VDU-100, R3
	10 09. DANA (SET) # LINE NOT CATTLE TO CALLER 10 09. EASSE (CLEAR) * 10 10. INVERT (REVERSE) LINE 10 11. RESERVED 10	6496 CF7800 6499 CF7800 649C CF7C00	1 STRA, RØ VDU-200, R3 1 STRA, RØ VDU-300, R3 1* 1 STRA, RØ VSTAT, R3 CLEAR ATTRIBUTES
	I* RECTANGLE OPS I* 12. SET RECTANGLE WITH OPP. CORNERS (X1.Y1) & (X2.Y2) I* 13. CLEAR I* 14. INVERT " I* 15. RESERVED	649F CF7000 64A2 CF7E00 64A3 CF7F00 64A8 D066	1 STRA. R0 VSTAT+100. R3 2 STRA. R0 VSTAT+200. R3 3 STRA. R0 VSTAT+200. R3 1 BIRR. R3 INT2 1* 2 LODI. R0 (VDU SET UP POINTERS
	<pre>1* CIRCLE OPB 1* CIRCLE OPB 1* 16. SET AN ELLIPSE AT CENTRE (X1.Y1), 1/2 WIDTH DF (X2) 1* AND 1/2 MEIGHT OF (Y2) 1* 17. CLEAR ELLIPSE</pre>	64AA 0478 64AC C830 64AE C82F 64B0 CF0408 64B3 17	ELDIARO KODO SEI DE POINTERS I STRR.RO POINT+1 I STRA.RS POINT+1 I RETC.UN
	1+ 19, INVERT ELLIPSE 1+ 19, RESERVED 1/VDU EQU 7800 BASE OF VDU	64B4 0437	PROT ACON M'8437 BINBUG SCROLL PROTECT STORE Single pixel functions
	IVSTAT EQU VDU+400 VDU ATTRIBUTE RAM IXMAX EQU 7F MAXIMUM HORIZ, COORD.		** SET A POINT ON SCREEN @ R1.R2
	IVMAX EQU 3F MAXIMUM VERT. COORD.	6496 3828 6486 14	:* SETBIT BSTR. UN CONVRT CONVERT POINTERS TO AN ADDRESS I RETC. EQ IF CC=EQ, COORDS OUT OF RANGE.
6400 1909	I ORG H'6408' IENTRY BCTR, UN ENT2 NORMAL ENTRY POINT. I.e ASSUMES ALL NECESSARY VALUES MAVE BEEN POKED INTO STORAGE LOCATIONS.	6489 6843 6488 C841 6480 17	I DORN. RØ «POINT BIT-MASK IN RØ STRR. RØ «POINT RETC. UN
6402 0B	IVERSN DATA ØB VERSION NUMBER		** CLEAR POINT ON SCREEN
6403 00	1. STORAGE AREA FOR EXTERNAL ACCESS 1. FUNCTN DATA 2 FUNCTION NUMBER	64BE 3B23 64CØ 14 64CI 24FF 64C3 4899	ICLRBIT BSTR.UN CONVRT CONVERT POINTERS RETC.ED TEST FOR CLIPPING (CC=ED) EORI.R0 0FF INVERT MASK ANDR.R0 *POINT
6484 88 6485 88 6486 88	IXI DATA Ø START CO-DRDINATE STORE	64C5 C897 64C7 17	STRR.RØ +POINT RETC.UN
5407 00 5408 00	PY2 DATA B BITVAL DATA D TESTED BIT VALUE		** INVERT A POINT ON SCREEN (REVERSE IT)
	1* INTERNAL USE FUNCTION INDEX	64C8 3819 64CA 14	INVEIT BETR, UN CONVRT CONVERT POINTERS RETC, EC TEST FOR CLIPPING EORR.RO +POINT
6409 00 6400 0402	INDEX DATA 0 FUNCTN ND. + 3 IENT2 LODI, R0 LCOM LOG, COMP, CLEAR C, WC	64CB 2891 64CD C88F 64CF 17	STRR.RØ «POINT RETC.UN
640C 93 640D 0975 640F 0974	LODR, RI X1 PICK UP CO-ORDS LODR, R2 Y1		** TEST STATE OF POINT ON SCREEN ** IF POINT "ON" RETURNS ROUFF CC-LT
5411 0970 6413 E713	LODR. R3 FUNCTN GET FUNCTION NUMBER COMI, R3 D'19' SILLY?		14 " " "OFF" " 00 " ED 14 IF POINT OUT OF RANGE, RETURNS "OFF" STATE CONDITIONS.
6415 15 6416 Ø3	RETC. GT YES! LODZ, R3 MULTIPLY FUNCTION BY 3	6400 3811 6402 4889 6404 1802	ITSTBIT BSTR.UN CONVERT CONVERT POINTERS ANDR.R8 *POINT BCTR.ED TSTB2
6417 DØ 5416 83 6419 C3	I RRL, RØ FOR USE AS INDEX I ADDZ.R3 I STRZ.R3	64D6 04FF 64D8 CC0408 64D8 17	E LODI, RO 2FF POINT WAS "ON" STRB2 STRA, RO BITVAL RETC. UN
641A CB6D 641C 9F641F	STRP.RS INDEX BXA FNTBL.RS JUMP TO FUNCTION		LOCAL STORAGE FOR SINGLE POINT OPERATIONS
641F 1F645D	: FUNCTION JUMP TABLE IM IFNTBL BCTA, UN WINDOW 00, FLOOD AREA WITH 1'S	64DC 00 64DD 00 64DE 7800	INPLOT DATA 0 COORDS OF POINT IVPLOT DATA 0 POINTER TO VDU POINTER TO VDU
6422 1F645D 6425 1F645D 6428 1F6485	BCTA, UN HINDOM 01. CLEAR AREA TO ZEROES I BCTA, UN WINDOM 02. INVERT AREA BCTA, UN PROTET 03. CLR SCREEN, PROT (X1) LINES	64EØ 7CØØ	ATTPNT ACON VSTAT POINTER TO VDU ATTRIBUTES BITHAP DATA & MASK FOR "BIT WITHIN BYTE"
6428 1F6486 642E 1F648E	BCTA, UN SETBIT 04. SET BIT 0 (XI, YI) BCTA, UN CLABIT 05. CLEAR BIT BCTA, UN INVBIT 06. INVERT BIT		Coordinate to address conversion
6431 1F64CB 6434 1F64D0 6437 1F636D 643A 1F636D 643D 1F636D	BCTA, UN TSTBIT 07. TEST BIT BCTA, UN LINE 08. DRAW LINE (X1,Y1)-(X2,Y2) BCTA, UN LINE 09. CLEAR LINE BCTA, UN LINE 09. CLEAR LINE		10 Converts X.Y coords in R1.R2 to VDU memory address & bitm to Coordinates 0.0 are at bottom-left of VDU. Coords greater to than XMAX & YMAX return a bitmask of 00 & CCRED. I*
6440 1F6459 6440 1F65F1 6446 1F65F1	BCTA, UN RESRVD ØB. RESERVED BCTA, UN RCTNGL ØC. SET A RECTANGLE	64E3 7702 64E5 C975 64E7 CA74	CONVRT PPSL LCOM STRR.R1 XPLOT SAVE A COPY OF CO-ORDS STRR.R2 VPLOT
6449 1F65F1 644C 1F645B	BCTA, UN RCTNGL ØF. INVERT " BCTA, UN RESRVD ØF. RESERVED	64E9 E57F 64EB 193D 64ED E63F	COMI.RI XMAX XCOORD) MAX? BCTR.GT CONV3 COMI.R2 YMAX YCOORD) MAX?
644F 1F662A 6452 1F662A 6455 1F662A 6458 1F662A	BCTA, UN ELIPSE 10, SET AN ELLPSE BCTA, UN ELIPSE 11, CLEAR " BCTA, UN ELIPSE 12, INVERT " BCTA, UN RESRVD 13, RESERVED	64ED E63F 64EF 1939 64F1 26FF	EDETR. GT CONV3 EDETR. GT CONV3 EDETR. R2 DFF INV YCOORD TO SUIT SCRN ADDRESS
6458 1F645B	Reserved function handler	64F3 040A 64F3 93 64F6 51	LODI, RØ WC+LCOM CLR C, SET WC & LCOM LPSL RRR, R1 GET L/R BIT OF X VALUE
545B CØ	RESEVD NOP NO OPERATION	64F6 51 64F7 D2 64F8 7508 64FA 453F	I RRL.RI DELET ON BACK OF Y VALUE I RRL.RC PUT ON BACK OF Y VALUE I CPSL WC I ANDI.RIJSF RI HAS BYTE-ON-LINE COUNT
645C 17	Function handler routines	64FC 0407	: LODI. RO 07 EXTRACT BIT NO. FROM R2 : ANDZ, R2
	I+ SCREEN MINDOW FUNCTION	64FF 0C652C 6502 C85E 6504 C3	STRR. RO BITMAP SAVE
	USES (X1,Y1) & (X2,Y2) TO DEFINE OPPOSITE CORNERS OF A SCREEN AREA TO BE SET, CLEARED, If OR INVERTED.	6505 D2 6506 D2 6507 D2	RRL, R2 CONTINUE ROTATION OF R2 RRL, R2 RRL, R2
	1*	6508 0400	S LODI, RO BCO FINISH LOW BYTE OF POINTER

6450 3F6534 IWINDOW BSTA. UN DIRTST

WORK OUT STEPPING DIRECTIONS

ETI July 1983 - 43

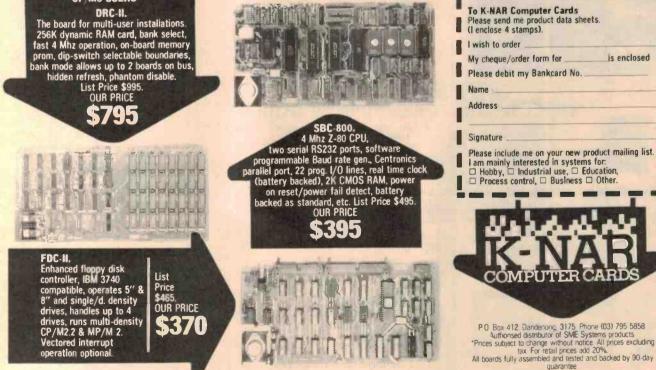
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650A 42 650B 61	1	ANDZ. R2 IDRZ. R1		658D 096D	1.00	LODR. R1 TEMPX	RECOVER POINTERS
650C C851 650E C851	1	STRR. RØ POINT+1 STRR. RØ ATTPNT+1	SAVE IT SET ATTR. POINTER	658F BASC	1 1+	LODR, R2 TEMPY	
6510 4603 6512 6678	-	ANDI. R2 03 IORI. R2 (VDU STRR. R2 POINT	BUILD HIGH BYTE OF POINTER ADD VDU ADDRESS	65C1 EE0407 65C4 181E	1	COMA, R2 Y2 BCTR, EQ ENDLIN	IF YPLOT=Y2 THEN STOP
6514 CA48 6516 8604 6518 CA46 6518 0402	8 8	ADDI. R2 04 STRR. R2 ATTPNT LODI. R0 02	SET ATTR. POINTER HI BYTE	65C6 8A69	1 10		IF V2) V1 THEN INCR Y BY 1 V2=V1 IGNORE V2(V1 THEN DECR Y BY 1
651C C2 651D 4AC1	8	STRZ . R2	TEST IF GREX ALREADY SET	65C8 0864	1*	LODR. RS DELTAK	SET R=R+DX
651F 1907 6521 CCB4E0 6524 20	1	BCTR. GT CONV2 STRA. RØ +ATTPNT EORZ. RØ	NO, SO SET IT NOW AND ERASE ASCII CHAR	63CA 8966 63CC CB64 63CE 085F	1	ADDR. R3 RUNCHT STRR. R3 RUNCHT LDDR. R0 DELTAY	DECIDE IF X IS TO BE CHANGED
6525 CC84DE 6526 03	CONV2	STRA. RØ +POINT	FETCH BITMAP & SET CC NON-ZERD	6500 A3 6501 E77F	1	SUBZ, R3 COMI, R3 7F	RO-DY-R If RUNCNT) 127, do next
6529 17 652A 20	t CONV3	RETC. UN EORZ. RØ	BAD COORDS, SET RØ & CC-00	65D3 195E 65D5 7502 65D7 E3	2	BCTR.GT LLOOP2 CPSL LCOM COMZ.R3	
6528 17 652C 01020408	I +	RETC. UN DATA 1. 2. 4. 8	BIT-MASK TABLE	65D8 7702 65DA 1957	X = 1	PPSL LCOM BCTR,GT LLOOP2	IF R(DY-R THEN GOTO STEP 10A
6530 10204080		tion test subr		65DC 8952		ADDR. RI XINCR	IF X2) X1 THEN INCR XPLOT BY 1 X2=X1 IGNORE
	1+	by WINDOW. LINE & R		65DE AB4F	2+		X2 (X1 THEN DECR XPLOT BY 1
	1+ Tests 1+ Tests	X1 assingt X2 to s	DELTAY & YINCR	65E0 C050 65E2 104F	1	SUBR. R3 DELTAY STRR. R3 RUNCNT BCTR. UN LLOOP2	SET R=R-DY
6534 7509 6536 000404	I+ DIRTST	CPSL WC+CAR		65E4 000406 65E7 0E0407 65EA CD0404	ENDL IN	LODA, R1 X2 LODA, R2 Y2	SWAP X2, Y2 INTO X1, Y1 FOR FAST CONSECUTIVE LINES
6530 000404 6530 0700	2 2	LODA, R1 X1 LODA, RØ X2 LODI, R3 Ø	Compare X1 against X2 Prepare XINCR	65ED CE0405 65F0 17	2	STRA. R1 X1 STRA. R2 V1 RETC. UN	RETURN TO CALLING PROGRAM
653E A1 653F 180A	1	SUBZ. RI BCTR.ED CALDX2	Determine DELTAX X2=X1. XINCR=0			angle function	
6541 1906 6543 07FE	3.	BCTR. GT CALDX1 LODI.R3 -2	X2)X1, XINCR=+1 X2(X1, XINCR=-1		1+ 1+ PLO 1+ AS	TS A RECTANGLE USING	CODRDS (X1. Y1) & (X2. Y2)
6545 24FF 6547 8401 6549 8701	: : :CALDX1	EDRI.RØ ØFF ADDI.RØ 1	DELTAX negative, make it positive	65F1 3F6534	5.00	BSTA. UN DIRTST	Determine time directions
6549 CC05AE 654E CF05B0	CALDX1	ADDI. R3 1 STRA. R0 DELTAX STRA. R3 XINCR	Save ABS(DELTAX) Save X stepping increment	65F4 3027 65F6 800500	RECT1	BSTR. UN RPLOT	Plot pixel at R1.R2
6351 ØEØ4Ø5 6354 ØCØ4Ø7	:	LODA, R2 VI LODA, R0 Y2	Compare VI assingt V2	65F9 ED0406 65FC 9876	1	ADDA, R1 XINCR COMA. R1 X2 BCFR. ED RECT1	NEXT X Up to star; of next line yet? No. Neep Logelns.
6557 0700	1	LODI, R3 0 SUBZ, R2	Prepare VINCR	65FE 3810 6600 8E0581	RECT2	ADDA R2 YINCR	R1=X2 NEXT Y
655A 180A 655C 1906	3	BCTR. EQ CALDY2 BCTR. GT CALDY1	Determine DELTAY Y2=Y1. YINCR=0 Y2)Y1. YINCR=01	6603 EE0407 6606 9876	5	COMA, R2 Y2 BCFR, ED RECT2	More? Yes.
6556 07FE 6560 24FF 6562 8401		LODI, R3 -2 EORI, R0 0FF ADDI, R0 1	V2(V1, VINCR=-1 DELTAY negative, make it positive	6608 3813	RECT3	BSTR. UN RPLOT SUBA. R1 XINCR	R2=V2 NEXT X, STEP(-XINCR)
6564 8701 6565 CC05AF	CALDY1	ADDI. RS 1 STRA. RO DELTAY	Save ABS (DELTAY)	6610 9876	3	COMA. RI XI BCFR. EQ RECTS	More?
6569 CF0581 656C 17	1 8	STRA. R3 VINCR	Save V steppins increment	6612 3809 6614 AE0581	RECT4	BSTR. UN RPLOT	R1=X1 NEXT Y. STEP(-YINCR)
	Line f	unction		6617 EE0405 661A 9876	1	COMA- R2 Y1 BCFR-EQ RECTA	Hore?
	I+ I+ Plots	straight lines from	(X1. V1) to (X2. V2)	661C 17 661D 0F0409	RPLOT	LODA, R3 INDEX	Finished, return to caller. Pick up function index
656D 20 656E CC0582	IA ILINE	EDRZ. RØ	Clear slope prediction counter	5620 BF5407 5523 0D04DC 5625 0E04DD	1	BSXA FNTBL-H' 18' LODA, R1 XPLOT LODA, R2 YPLOT	
6571 3841 6573 E839	1.4	BSTR. UN DIRTST	Set up DELTAX, DELTAY, XINCR, YINCR	6629 17	Elling	RETC, UN	Return for next point,
6575 1D6583	5	BCTA. GT LLOOP2	Determine (f slope) 45 desrees Yes, do LLOOP2		1.0	e and circle fu	
6578 C932	2.4	must be (= 45 degre			1+ R2 co	ntains X1 centre co- ntains V1 centre co- ntains Horiz width/	ordinate
6578 C932 6578 CA31 657C ØF0409	ELCOP1	STRR.RI TEMPX STRR.R2 TEMPY LODA.R3 INDEX	SAVE POINTERS FOR LATER PICK UP INDEX FOR POINT OP'S.		10 V2 co	ntains Vert helent/	2
657F BF6413	1		RECOVER POINTERS	662A 800406 662D CD06EC 6630 CE06ED	ELIPSE	ADDA, RI X2 STRA, RI FIRSTX STRA, R2 FIRSTY	Add horiz width to Ri Save copy of first point
6584 0A27 6586 ED040E	1	LODR. R2 TEMPY		6635 20		BSXA FNTBL-H' 24'	R3 and plot first point
6589 1C65E4	1	BCTA. EQ ENDLIN	IF XPLOT=X2 THEN END	6637 0708 6639 CF46E4	: ELIP2	LODI. RS 8 STRA. RO CNTR. R3-	Initialise CNTR to 0.00
658C 8922	2 2.0 1.00		1F X2)X1. THEN INCR XPLOT BY 1 X2=X1. IGNORE X2(X1. THEN DECR XPLOT BY 1	663C 5878 663E 0501 6640 CD06E9		BRNR, R3 EL IP2 LODI, R1 1 STRA, R1 COSINE	Init SINE to 8.00 00 Init COSINE to 1.00 00
658E ØB1F	1+	LODR. RS DELTAY	SET R=R+DY		In MAIN	LOOP STARTS HERE	
6592 CBIE	1 19	STRR. R3 RUNCHT			: + SINE	SINE + COSINE/128	
6594 Ø816 6596 A3 6597 E77F	:	LODR. RØ DELTAX	DECIDE IF YPLOT IS TO BE CHANGED	5643 0D06E9 6646 0E06EA 6649 0F06EB	NEWSIN	LODA, R1 COSINE LODA, R2 COSINE+1 LODA, R3 COSINE+2	Get old cosine
6599 1950 6598 7502	1	COMI.R3 7F BCTR.GT LLOOP1 CPSL LCOM	If runcht) 127, do next	664C 3F66FØ	1	BSTA, UN SHR	Calc cos/128
659D E3 659E 7702 6540 1956	1	COMZ, R3 PPSL LCOM	IF ROX-R THEN GOTD STEP 2	664F 8F06E8 6652 7706 6654 8E06E7		ADDA+ R3 SINE+2 PPSL WC ADDA+ R2 SINE+1	Add to sine
65A2 8A0D	To:	ADDR. R2 VINCR	IF V2) VI. THEN INCR VPLOT BY 1	6657 8006E6 665A CD06E6	4	ADDA, RI SINE STRA, RI SINE	Save new sine
	1 0 1 0 1 0	IF	Y2-Y1, IGNORE Y2 (Y1, THEN DECR YPLOT BY 1	6650 CE06E7 6660 CF06E8	1	STRA. R2 SINE+1 STRA. R3 SINE+2	
65A4 080C	1	LODR, R3 RUNCHT	SET R=R-DX		1.00	E = COSINE - SINE/12	8
65A8 CB08 65AA 1B4C	1	STRR. R3 RUNCHT		6663 3F66F0	: NEWCOS	LODA, RD COSINE+2	Cate sin/128
65AC 00	1 .	VARIABLES USED IN L	INE ROUTINE	6669 A3		SUBZ. R3 STRZ. R3	Subtract R123 from cosine
65AD 00 65AE 00	DELTAX	DATA Ø	ABS(X2-X1)	6668 7788 6660 80868A 6678 A2	1	LODA, RO COSINE+1	
55AF 00 6580 00 6581 00	IXINCR	DATA 20 DATA 20 DATA 20	AB5(Y2-Y1) Horiz, direction flag Vertical """	6671 C2 6672 0006E9		SUBZ.R2 STRZ.R2 LODA.R0 COSINE	
6582 00	RUNCHT	DATA 0	SLOPE PREDICTION COUNTER	6675 A1 6676 C1 6677 CD06E9	3	SUBZ, R1 STRZ, R1	C
	34	STRR. RI TEMPX	The second se	6670 CE06E8		STRA, R1 COSINE STRA, R2 COSINE+1 STRA, R3 COSINE+2	Save R123 as new cosine
6585 CA75		STRR. R2 TEMPY	SAVE POINTERS FOR LATER		IN MULTI	PLY COSINE BY HIDTH	
658A 8F6413		BSXA FNTBL-0C.R3	ADD IT	6680 0C0406 6683 3FE707	: COSBYW	LODA. RO X2 BSTA. UN HLTPLY	Mult width in RØ by CDS on R1.2 (Returns RSLT in R1.23)

6686 8E 0404		ADDA, R2	¥1	Add centre X coord	66FØ	8488	SHR	LODI, RØ	WC	Divides R123 by 128
6689 83	-	LODZ , R3		Test fractional portion	66F2	93		LPSL		by shifting R123 right 7 bits
				D7 = rounding flag						
668A 9A02	3	BCFR, LT			66F3	07		RRL.R3		Set up CAR from MSB R3
668C 8601	R	ADDI, R2		Round Upwards				LODZ, R2		Set of CHR From the No
668E CEØ6EE	: NORND	STRA. R2	NEXTX	New Xplot coord (tentative)	66F4					Children di succe des stabilit
					66F5			STRZ.R3		Shift 1 byte to right
	5.00				66F6	D3	1	RRL, R3		and back 1 bit (pick up CAR)
		TOLV RINE	BY HEIGHT							
	:+	11-61 91146	DI NEIGHT		66F7	81	1	LODZ . R1		
			OTAF	Fetch sine into R12	66F8		2	STRZ. R2		
6691 0D06E6		LODA, R1		Petch Sine Into H12	E6F9			RRL, R2		Put cid sign bit in CAR
6694 ØEØ6E7	:	LODA, R2			Gor 3	24				
6697 000407	1	LODA, RØ	Y2	Mult height in RØ by SINE in R1.2	66FA	20		EORZ, RØ		Perform sign-extend thru into R1
669A 3F6707	1	BSTA, UN	MLTPLY	(RSLT in R1.23)						Assume the number
					66FB		1	STRZ.R1		
669D 8E0405	1	ADDA, R2	¥1	Add centre Y coord	66FC	DØ	1	RRL, RØ		Test old sign bit
66A8 83		LODZ, R3		Test fractional portion	66FD	7508	1	CPSL	WC	
		BCFR, LT	NORNES	D7 = rounding bis	66FF	14	8	RETC, EQ		Return if sign = 0
66A1 9A02				D7 = 1						
66A3 8601	1	ADDI. R2		07 - 1	6700	OSFF		LODI.R1	-1	Number was nesative
66A5 CEØ6EF	INORND2	STRA, R2	NEXTY		6702			RETC, UN		
					6/182	. /		HE ICION		
6648 ØDØ6EE	ITSTSAM	LODA, R1	NEXTX	Test if new coords same			1.0	D	Due Due	Multipulsetion Sub-foutipe
56AB ED04DC	1	COMA, R1	XPLOT	as last point plotted.				BIE # 8	pit pinari	Multiplication Sub-routine
66AE 9805	:	BCFR, EQ					2.0			
66BØ EEØ4DD		COMA, R2							1 = Muitin	
				Same, dont plot it			3+ R1	.2 = OPR	2 = Multi	pilcand
66B3 1814		BCTR, EQ	2146114	Samer addite provine			2.41			
and the second second					6783	0.0	10PR1	DATA	0	Input operand storage
6685 CD04DC	INOTSAM	STRA, R1		Update old coords		0000	+OPR2	DATA	0,0	
6608 CE04DD	:	STRA, R2	YPLOT					DATA	8	Multiply loop count
66BB E92F	1	COMR, RI	FIRSTX	Test If same as start	6706	96	MONT	DHIH	0	
66BD 9804	:	BCFR, EQ	NOTSM2	to avoid double operation			3 A M A			
668F EA2C		COMR, R2			6707	7708	IMLTPLY	PPSL	WC	
		BCTR.EQ			6709	C878	1	STRR. RØ	OPR1	Save OPR1
66C1 1806				and plot new point	670B	C977	ā	STRR, R1	OPR2	Save OPR2
66C3 0F0409	:NOTSM2					CA76	:	STRR. R2		
66C6 BFE3FB	1	BSXA	FNTBL-H' 24'	, K3	6700	CHIO				
					C 705	20		EORZ, RØ		Ciear result
6609 0919	SINCTR	LODR. R1	CNTR	Fetch old loop counter value	670F			STRZ, RI		
66CB 0A18	1	LODR, R2	CNTR+1		6710					
66CD DA02		BIRR, R2	6+4	Add D. DI (HEX) to It	6711		1	STRZ, R2		
66CF 0900	1	BIRR, R1			6712	C2	1	STRZ, R3		
66D1 C911		STRR. R1		Save new loop count	6713	8486		LODI, RE	8	Loop count
				Dare new root count						
66D3 CA10		STRR, R2	Claima 1		6715	7501	:MLOOP	CPSL	CAR	
						C86D		STRR, RE		Save loop count
66D5 7702	IENDIST		LCOM	Compare loop count to end value				LODR, RE		Shift OPRI Right one bit
66D7 E503	1	COMI, R1		of PI = 03.24 3F 6A 88 (HEX)		0868		RRR. RØ		
66D9 1E6643	:	BCTA.LT	NEWSIN	More to do.	6718				0001	Save It
66DC 1905	3	BCTR, GT	DONE	Over-shot?		C865		STRR. RE		
					671E		2	EORZ, RE	,	Test 14 add to be done
66DE E624		COMI . R2	H' 24'	MS Byte equal, test next	671F	DØ	3	RRL, RØ		
66E0 1E6643		BCTA LT		More to do.		1804		BCTR.EC	MNDAD	If CAR was 0, no add
66E3 17	I DONE	RETC UN		FINISHED, return to caller.	6700	8461	: MADD	ADDR. R	OPR2+1	Add OPR2 to RSLT
						895E	1	ADDR. RI		
	I+ Loc	al storas	e for circle	routine	8724	GODE		MUDRIT RI		
								000 01		Shift entire result risht
66E4 0000	I CNTR	DATA	0,0	Loop counter store	6726		: MNDAD	RRR, R1		
	ISINE	DATA	0, 0, 0	Sine value store	6727		1	RRR, R2		by one bit
66E6 000000			1,0,0	Cosine store	6728	3 53	1	RRR, R3		
66E9 010000	COSINE			First plot point store						
66EC 00	FIRSTX		0	FITSE PIDE POINT BEORE	6729	9 085B	1	LODR, RE	MONT	Fetch current loop count
66ED 00	FIRSTY		8			5 F868	-	BDRR, RE		More to do?
66EE 00	*NEXTX	DATA	0	Tentative new plot point	6/41					
66EF 00	INEXTY	DATA	0			7500		CPSL	WC+CAR	
			SUB-ROUTINE			7509				Finished
	IT DIVI	DE DT 120	DOD-HOUTIN		672	- 17		RETC, UP	4	r in (Brittin :

S100 Z80 System Cards

CP/M3 USERS





YOUR No. s. 2 **ROD IRVING ELECTRON**

THIS MONTHS KITS FROM EA

ETI-163 LAB SUPPLY

ETI May 1983



Fully variable 0-40 V current limited 0-5 A supply with both voltage and current metering (two ranges: 0-0.5 A/0-5 A). This employs a conventional series-pass regulator, not switchmode type with its attendant problems, but dissipations is reduced by a unique relay switching system switching between taps on the transformer secondary.

EPROM PROGRAMMER

\$43.00

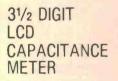
No need for a Micro with EA's great Eprom Programmer suitable for 2716/2758 Eproms.

With Textool Sockets \$55.00 EA January 82

ELECTRONIC METRONOME

\$16.90

Great new Metronome Circuit with iow current drain (less than one milliamp) drives a Loudspeaker and a Led Indicator. EA January 82



Handy pocket size Digital Capacitance Meter, runs off a 9V battery and measure 1pF to 19.99uF in just three ranges. EA March 82



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\$79.00



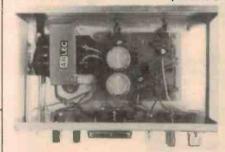


For a low cost Storage CRO with Synchronised Display, Flectronic Graticule, One-Shot Triggering and Optional Storage of up to four Screen Displays it can't be beaten. EA February 82



ETI-688 Every digital workshop should BIPOLAR have one! Can be used to PROM program the popular fusible-ROGRAMMER link PROMs like the 74S188/ ETI June 1983 288, 82S23 and 82S123 etc.

DUAL TRACKING POWER SUPPLY \$83.50



Built around positive and negative 3-Terminal Regulators, this versatile dual tracking Power Supply can provide voltages from $\pm 1.3V$ to $\pm 22V$ at currents up to 2A. In addition, the Supply features a fixed $\pm 5V$ 0.9A output and is completely protected against short circuits, overloads and thermal runaway. EA March 82

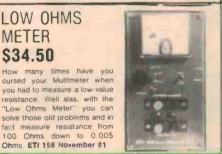
VOICE OPERATED RELAY \$14.95



EA's great new Voice Operated Relay can be used to control a tape recorder, as a VOX circuit for a transmitter, or to control a slide projector. EA April 82

FUNCTION GENERATOR \$79.50

This Function Generator with digital readout produces Sine, Triangle and Square waves over a frequency range from below 20Hz to above 160kHz with low distortion and good envelope stability. It has an inbuilt four-digit frequency counter for ease and accuracy of frequency setting. EA April 82



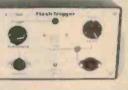


MICROBEE EPROM PROGRAMMER Simple, low cost programmer for the MicroBee can program 2716s, 2516s, 2732s and 2764s. P.O.A. ETI Feb 1983

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This easy to build sound or light operated flash trigger has many feature. Catch those

spectacular and h u m o r o u s moments like that time your motherin-law slipped on the moss covered patio and broke her neck. ETI 568 October 80



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\$14.95

The "Le Gong" Doorbell with those unmistakable chimes generated by the LSI. A must for the man who has everything! EA February 81

LE GONG

LED LEVEL METER

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Build a Led level Meter with simultaneous peak and dynamic range. This kit is Ideal for any application requiring a wide dynamic range level display. ETI 458 June 81



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You have to be in it to win it Take the chance out of winning the Pools as well as Lotto, and build the great new Pools/Lotto Number Selector. EA July 81

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Have great fun creating your own recording effects with music and voice. The Sound Bender can receive from Electric Guitar, Microphones, etc. ETI February 82



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POWER AMPLIFIER

Please note that the "Superb Quality" Heatsink for the power amp was designed and developed by Rod Irving Electronics and is being sup-plied to other kit suppliers. This product cost \$1,200 to develop so that your amplifier kit would have a professional finish as well as sound. We also have a new range of rack mounting boxes which will be released soon.

SPECIFICATIONS 100W RMS into 8 ohms (\pm 55 V supply). 8 Hz to 20 kHz, \pm 0-0.4 dB 2.8 Hz to 65 kHz, \pm 0-3 dB. NOTE: These figures are determined solely by passive filters.

Power output: Frequency response:

Input sensitivity: Hum: Noise:

Stability:

No.

Expirt Date.

Signarure

2nd harmonic distortion: 3rd harmonic distortion:

Total harmonic distortion: Intermodulation distortion ngures are determined solely by passive linters. 1V RMS for 100W output. — 100dB below full output (flat). -116 dB below full output (flat). < 0.001% at 1 kHz (0.0007% on prototypes) at 100 W output using a ±56 V supply rated at 4 A continuous. < 0.003% at 10 kHz and 100 W. < 0.003% for all frequencies less than 10 kHz and all powers below clinoing clipping. Determined by 2nd harmonic distortion (see above). < 0.003% at 100 W. (50 Hz and 7 kHz mixed 4:1). Unconditional

THIRD OCTAVE GRAPHIC EQUALIZER

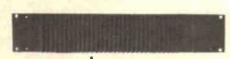
MX-1200 MICROPHONE/AUDIO MIXER

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NIX 1200 SASE this month only



Price Slashed \$299 \$279

SPECIFICATIONS E.T.I. Dec. 1982 Bands: 28 Bands from 31.5 Hz to 16 kHz Noise 20 kHz bandwidth Distortion:

< 0.008 mV, sliders at 0, gain at 0 (-102 dB),

0.007% at 300 mV signal, sliders at 0, gain at 0; max. 0.01%, sliders al minimum. 12 Hz-105 kHz, +0, -1 dB, all controls flat. 14 dB

Frequency Response Boost & Cut:

1 unit \$189 2 units \$359

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- only \$425 *All parts available separately for both kits.

PREAMPLIFER

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 English "Lonim" Switches are supplied to substitutes as
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- We have built and tested this unit and so know what needs: is, op into every kit Specially imported black anotised auminium knobs Again as with the power and we are offering this kit A & T at a price which we do not believe there is a commercial unit available that sounds as good. Same derivery as the PA • ň only \$425



FADER & CONTROLLERS 12 channel fader; Silde, 80m/m, LOG 28% Valate: fader; Silde, 80m/m, LOG 28% 12 /78 Valame; 300; LIN 12 /78 Valame; 300; LIN 12 febra 300; LIN 12 febra 300; LIN 12 febra 300; LIN 12 febra 300; LIN 14 febra 30 IDB JENCY RESPONSE: 20-20 KHz L HARMONIC DISTORTION, LONG han 0.1% METER: 2 duminated VU Meters 0db =

This unit features: 12 microphone line inputs with pan, bass, treble, effect and fold back controls for each channel • LED peak inditators for each channel • 2 furntable inputs with cross-lade and individual output controls • master equaliser for bass, midrange and treble • variable headphone output etc. etc. • complete with carrying case.

75V IX INDICATOR: 12 LED Peak indicators TAGE 240 VAL 50Hz WER CONSUMPTION: 7.2 watts IENSIONS, 620 (W) x 386 (D) x 106 (M) mm pythed complete with carrying case)

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The 'pfs:Graph' package reviewed

Ever tried to tackle the problem of producing a really flexible graph routine? Difficult isn't it? Well, this package should interest you — it runs on the Apple II and produces graphs to order from your own data. What's more, it's compatible with PFS and VisiCalc files!

AS ANYONE who has ever tried to produce a flexible all-purpose graph routine will know, graphs wasn't meant to be easy — to paraphrase a well known aphorism.

The first problem comes when you try to work out what the scale is going to be. That's fairly easy. Then comes the problem of labelling the axes — and if you want to do it properly (i.e: with the vertical axis labelled using letters on their side), then that means a *lot* of fiddling.

Then of course comes the question of how to show the line — just as a solid white line, or with symbols to show the data points?

When your data starts to come in the form of *comparitive* graphs — i.e: more than one graph on the same set of axes — then you are in real trouble.

The *pfs*:Graph package is the third in a series of packages designed for the business user — the sort of user who is not interested in staying up to three in the morning polishing software routines. What he wants is results without headaches.

The 'mother' package, 'pfs:' itself, is a 'data aquisition' system — that is, it allows you to design forms on the screen and then fill them in. Packages like this are becoming very common these days, as people realise that computers are not only fun — they can work for their living, too.

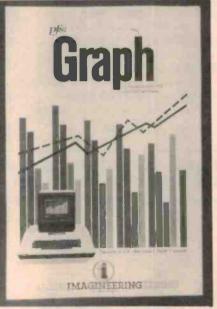
Now, pfs: allows you to generate 'files' of information, and to add to and change that information at will — more, it allows an *operator* to do this sort of thing, integrating the system into an office environment.

The second package produced by Software Publishing Corporation (the publishers of the pfs: series) was 'pfs:Report' — a 'report generator' package, which allowed that same no-headaches business user to generate printouts formatted neatly without worrying about the details of writing a program to do so.

Cunningly, *pfs*:Report took its data from the 'files' produced by the mother *pfs*: system — *pfs*:Graph does the same thing, only instead of reports, it produces graphs.

Note that *pfs*:Graph can also be used without the mother system, entering data directly to produce a graph.

The *pfs*:Graph package runs on an Apple II (with disk drive, of course), and produces the graphs in colour on the screen. The package is also capable of interfacing with a variety of printers and plotters for hard copy.



Recommended. The 'pfs:Graph' package is distributed by Imagineerng, 3/579 Harris St, UltImo NSW 2007. (02)212-1411.

How it performs

Starting up the *pfs*:Graph package is something of an experience. The disk drive whirrs for what seems like rather a long time, with the head doing what sounds like a polka across the face of the disk.

After a while, all that stops and the main menu is seen, giving you the option of retrieving/saving the graph, altering the data, altering the format of the graph or printing the results.

Operating the menu is also rather an experience — for some reason the programmer has decided to break with convention, and use CTRL-C (which, on most systems is used to mean 'interrupt') at the end of each input.

So if (as I did on a number of occasions, by mistake through long habit) you enter the menu selection followed by RETURN, nothing happens! Worse still, going back to a 'normal' system after playing with *pfs*:Graph for some time results in several unnecessary reboots!

Another problem I noticed was that, in going from the main menu to the various other parts of the system, each time I went

Jonathan Phillips

into or out of the main menu, the disk was used.

Now this fact, coupled with the usual protection against copying the disk, will give the whole system rather a limited life — until the disk wears out, in fact.

Problems apart, the system works well — it will produce three types of graph: line graphs, bar charts and pie charts.

Bar and line graphs can be generated using up to four sets of data on the same axes, and bars and lines can be mixed on the same set of axes.

Labelling of each line, bar and pie segment is clear and easy to follow — lines are labelled either by colour, or by a symbol at each data point (depending on an option). Bars and pie segments are identified by shading or colour.

All of the graph examples that I fed into the system resulted in neat, easy to read graphs — in fact, it would be very difficult to produce results that were otherwise, using this system.

By making a few simplifying assumptions (e.g. that the first set of data on a graph using colour will always be in red), the system allows you just enough choices to define what you want the *reader* of the graph to get out of it, without burdening you with unnecessary choices.

All in all, the package is easy to use and produces good results — and now we come to the most important part of the whole deal, the manual.

In a system which is designed for use by 'business' people, excellent documentation is a must. The *pfs*:Graph manual is very much up to scratch.

It is easy to read, easy to use for reference and includes a short 'guided tour' of the features of the package, along with 'monkeysee, monkey-do' instructions of how to start the thing up.

The package also comes with a disk full of example graphs, so that the naive user can get a few ideas of what is possible. All in all, the publishers have done a very good job of making sure that the documentation is up to standard.

In summary — yes, a good product; especially for those with either the base *pfs*: system, or VisiCalc (the spreadsheet package with which *pfs*:Graph is also compatible). *Recommended* for those who are into computing for profit, as well as for fun.





TV MESSAGE MAKER/INVENTORY

I initially designed this program in black and white but now I've learnt more about colour and programming techniques such as Machine Code Sub-Routines (MCSRs).

The program writes messages on a TV or monitor and has four separate memories in which you can record a full screen of writing and call it up at any time with a two-key code. With the options I have listed it is possible to print a page over a page that is already on display, for special effects.

As well as writing four pages, there is also a stack in which you can record messages. The messages can be put in the stack one after another and, after resetting the stack to the start again, they can be displayed by pressing key B. The first message will be displayed and, if key B is pressed again, the first message will be cleared and the second message in the stack will be displayed. And so on.

If you want to display a message for the second time, or a message further up in the stack, type in 'CD' followed by its address, which you have to note when recordina

The vowels, being most commonly used, have been assigned 11, 22, 33, 44, 55, i.e. A, E, I, O, U. There are several MCSRs starting at 0862 and all of them end in D4. Before calling some of these I have used the Chip 8 Axxx instruction to set register A of the 1802 to the required address. This method is slower but saves memory and so may not be suitable for those who have expanded the memory as the most significant high four bits of register A could be any value (which is not important in the previous case).

Other MCSRs manipulate the monitor's Chip 8 variable stack (0470-047F) so they communicate between each other.

The use of machine code was necessary to make the program as short as possible and colour the whole screen or line in a shorter time than the Chip 8 equivalent would do.

The MCSR shown in ETI April 1982 had to be modified so that the memory data pointed to by the SEXed register was replaced, after the OUT3 was performed, to its original value.

To prevent writing over page boundaries (not a page as 00-FF) I had to place a maximum of letters/ codes per page/display. I settled on a figure of 119, which will still leave room for the BB if you want to record what is on display. This should be sufficient and any attempt to go over this limit will prove frultless; no harm can be done to the recorded pages.

You'll know when this limit has been reached as a strange shape will be displayed after only the first key press, and a letter requires two. If this limit is reached you either have to record what's on the display, erase the last entry/line or clear the screen before you display any more messages.

The display can still be recorded even though this sign has come up. The strange shape will not be recorded as a BB is placed over it automatically

Any Invalid command will be greeted with a long beep.

Once you've got the program up and running, record from 0600 to 0970. If you are planning to demonstrate it, record all your messages and check them to see that they are what you wanted. Before recording press C00, then after reloading into the computer it will not be possible to accidentally record a message. Don't forget the colouring. Then record on another tape from 0600 to 0FFF. When you want to demonstrate your program you reload the computer, all your messages are set, and you can recall them as required

Seven lines will be displayed with an average of 14 letters, subject to a maximum of 119 letters/codes per page/display, counting each code in tables A and B as one

The code recorded in memory for cursor forward (key F) is recorded as one code for each time this key is held down, even though the cursor moves forward more than one space. The cursor can move to a maximum of 16 spaces before the key will have to be released but will stop short at the start of the next line. Four separate pages can be called up at any time.

The screen will be cleared first before a page is displayed

There can be an average of 60 lines in the Inventory. More lines are possible if only parts of a line are used; i.e: the message

BY W.F. KREYKES 1983

on three lines will only take up 24 codes + BB = 25 codes; that's 17 less than the average of 14 per line, but didn't include any colouring.

Options: Less than a page or one extra full page.

Data at	No page	1 page	2 pages	3 pages	5 pages	2x1/2 and 1 full	
0662 and 0686	6100	6f01	6102	6f03	6f05	6103	at 0930
078A	6009	600a	600a	600b	600c	600a	aa24
078C Start of	61e8	6160	61d8	6150	6140	61d8	at 0932
Inventory	09E8	0A60	0AD8	0850	0C40	0AD8	aa60

For 2x1/2 pages, with a full 58 codes per 1/2 page (1&2), there is no provision to prevent recording over page three, so be careful.

To display a page over a page, change 0690 to 00ff. If you want to clear the screen before displaying another page use code A8 at the start of the page that requires the screen to be cleared first. This feature creates special effects when used carefully.

If you only want to recolour the bottom three lines using A9, change 175c at 0758 to 1760; this leaves the top four lines in their original colour.

After recording into the stack, the address showing where the next message will be recorded will have to be removed by pressing any key. This will then return the cursor to the start of the display.

Attempting to display a message in memory that has not been recorded will bring up rubbish; you may have to restart the program.

To change the screen colour to something other than white, change the value of V0 at 075A to 60aX; X is the colours' code 0-7. To have different colour combinations when using A9, experiment with the value added or deducted from V0 at 0757, or replace it with OOff

Table 'B' - These codes are recorded as for Table 'A' 94

used to signal inventory to reset to the start (endless loop)

- AO black
- A1 red A2 blue A3 violet
- **A**4 areen A5 yellow
- A6 p. blue A7 white

- **A8** clear screen, leave cursor in present position
- A9 change back/foreground colour **
- show? AA
- AB white screen
- AC black
- AD green
- AE red AF blue

F

- D move cursor down one line
 - move cursor forward until key is released. subject to a maximum of 16 advances and/or the start of the next line

A0 and A7 colour the whole line that the cursor is currently on

AC and AF set background colour

"Not activated when clearing stack or screen.

- Table 'C' Commands/instructions not recorded. В show next message in stack (inventory),
- clear previous one C1-4
 - display selected page, clear whole screen first

CC	clear screen immediately, reset cursor to
	start of display
CD	go to address specified by next three key
	presses, and display message
C00	enforce safety lock which prevents
	recording
C01-4	record display into selected page
CF	reset stack/inventory to start immediately,
	clear previous
E	erase last entry
EE	erase last line

Note:

1. Record all your messages into the stack at once. To record into the stack after the first few messages have been recorded and the stack has been reset, simply bring up each message and when the message comes up that is previous to the one you want to record over, clear the screen (CC) and then you can record your message. Remember that recording into a page resets the stack to the start again, so be careful. 2. Don't forget to colour the lines and background as these will not change colour until instructed. 3. Code A8 should be used with care.

4. Don't record into the stack if the address shown after a recording is over FFF, e.g: 014. In this case you will have to see if the whole of the previous message recorded has been able to fit in.

5. You can display a page and then record it into the stack but you cannot display the stack to record into a page

6. The first message recorded into the stack will be recorded at BC8. After each recording an address will be shown at the top of the screen; this is the address that the next message will be recorded at. If you note these as with a tape recorder you'll be able to recall them with the CD command later.

TV MESSAGE MAKER/INVENTORY

								AT	BL	E 'A
0600 0610 0620	086e 6100 1670	6caf a8fa 440e	278 a d121 1816	275a f40a 4377	26a0 d121 18c4	6200 440b 8444	6 b 00 179e 8444	83b0 440c 8444	A B	1† 77
0630 0640 0650	8444 2644 1055	8040 1612 34f0	44d0 4100 26ce	1640 8 b 30 3377	4410 a404 7301	18c0 1355 00ee	ff0a a96e f60a	8014 131e 4600	CD	38 6d
0660 0670 0680 0690	1766 f60a 165c	6f04 460c 460f	8f65 1608 2786	4f01 460d 6f04	16a6 183a 8f65	6040 460b 3f01	f018 16 <u>a6</u> 166a	1612 4600 277c	F	22
06A0 06B0 06C0	26a0 00e0 182c f129	aXXX 6d16 a96e ddf5	f31e 00ee 087d 7d06	f065 60bb 0862 f229	40bb a96e 26a0 ddf5	17a6 f31e f029 f00a	2644 f055 ddf5 1608	1692 360b 7d06 659f	G H I	73 14 33
06D0 06E0 06F0	8505 4010 40d0	3f00 00ee 1856	1716 70ff 40aa	65 1 0 16de 185c	8502 40a9 40a8	35f0 1744 1744	16e8 40ab 65af	2730 175a 8052	J K	85 7c
0700 0710 0720	65ac 4cb0 6559	8502 6cac 8575	35ac 1708 3f00	17d4 a8d0 172e	90c0 f01e 7101	00ee d125 4790	08bd 67f0 7101	7 c01 8702 7 102	L M	29 96
0730 0740 0750	7102 6100 6cac	6f3d 00ee 08bd	8f15 4dee 80c0	3f00 00ee 70fd	00ee 30a9 175c	7207 16a0 60a7	4231 7c01 ac88	6200 4cb0 089f	N O	80 44
0760 0770 0780 0790	ad68 166a 6300 1782	0891 a92c f155 a402	00ee 8664 173e f165	606a f61e 3d16 a400	a67d f165 2792 6dee	f055 00ee 600b 27b0	a681 2772 61c8 16a2	f055 a692 a400 3d16	P Q R	50 68 8a
07A0 07B0 07C0	2792 2782 17b2	6daa a470 6100	27ac 0881 83b0	613f 4094 9830	622f 178a 160e	1612 40bb a96e	a400 00ee f31e	f165 26ce f065	ST	48 32
07D0 07E0 07F0	2654 ad28 00ee	17c6 421c 4100	ac88 ad68 180e	4207 4223 7001	acb8 ad98 d121	420e 422a 2730	acf8 add8 a8fa	4215 08a3 d121	U V	5 5 90
0800 0810 0820	7f01 264c 9830	3f3f a8fa 17c2	1800 1616 26ce	40ff 9b30 a404	180e 1612 1365	efa1 8830 1612	17f2 a404 2772	73ff f365 a400	W X	99 92
0830 0840 0850 0860	f155 8ee4 a400 172c	a96e 8ee4 f155	087d 8ee4 17a2 a78e	278a f00a 340e	1606 8e04 173a	f90a 2786 00ee	fe0a 8090 a8ca 3099	8ee4 81e0 d126 f839	Y Z ?	18 59
0870 0880 0890	a7e7 47be 00a7	f872 f82c 07ae 47be	a76e 5762 ee4a 07ae	5727 27f8 5ef8 4e5a	f6f6 2057 bbf7 8e57	f6f6 62e2 1e3a 279e	d4f8 8530 57e2	1009 00a7 98f8 d4f8	-	aa 00 03
08A0 08B0 08C0	c030 9f5a 2644	a5f8 632a 17fa	20af 8e5a 6094	f870			bfea	foae 61d4 0020	•	0b 07
08D0 08E0 08F0	4040 00e0 e020	0000 a0e0 e080	00e0 a0a0 e080	0000 e0a0 e080	4000 a0a0 c080	4000 e040 8080	0000 4000 a0e0	0080 e020 2020	Q 0 1	4c 44 33
0900 0910 0920	2020 e0a0 e0a0	e040 e020 e080	4040 e0a0 80a0	4040 a0a0 a0a0		8080 e020 2040	e080 e0a0 80e0	e0a0 e000 a9e8	234	20 1e 2a
0930 0940 0950 0960	9000	aad8 e0f0 b090 8850	80b0 9070	2020	f090 5070 a0e0 a888	50f0 f090	f0e0 90a0 f0a0 a850	9090 c0a0 9000	567	48 3c 2d
	xx @ ()692 s	set by	rout	tine a	at 077	7C	0.0.	89	3e 40

660 SOFTWARE

MEMORY DISPLAY UTILITY

L. Chubb, Kingsford NSW

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I wrote this program because I was tired of debugging programs that didn't work, looking for errors byte by byte. It also makes finding typing mistakes a lot easier. It's useful for those people who like to develop their own programs as it sits high in memory, out of the way of other programs (location 0F00) until it is needed. To use this utility replace the first instruction of your program with 1F00, then run it. It will display eight instructions at a time with the

It will display eight instructions at a time with the address next to each one, up to 512 bytes above the base address. The program initially displays memory locations from 0600 to 07FF.

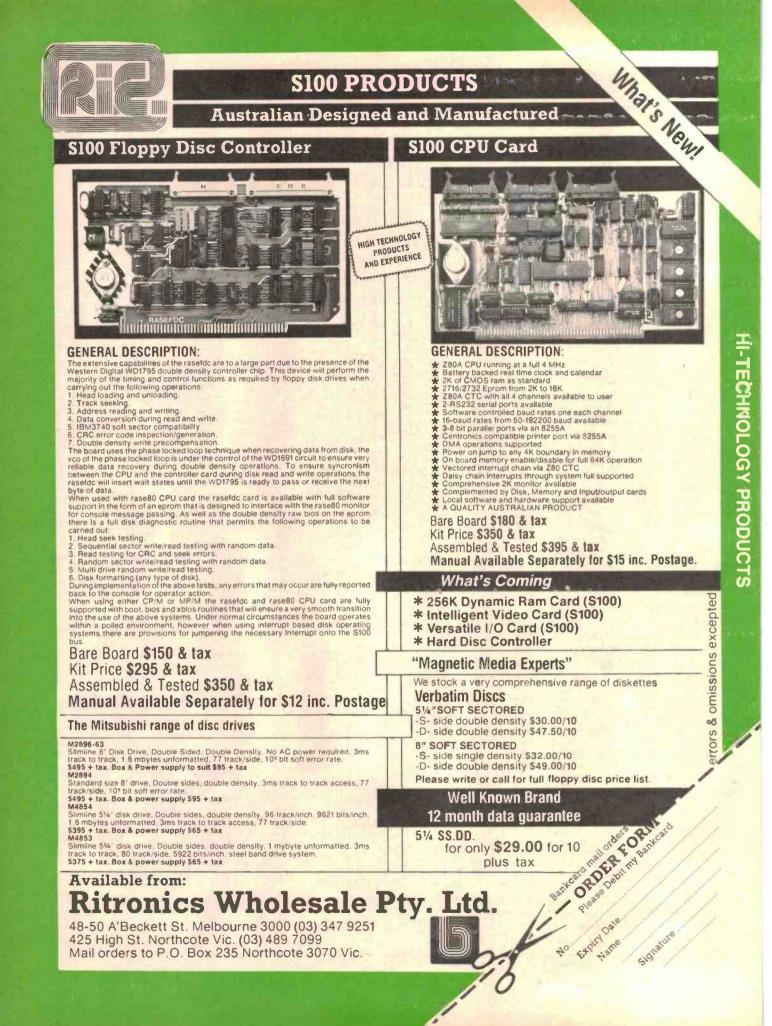
Key F will display eight instructions from the base address (at first the base address is 0600). Key D will display the next eight instructions on from the ones currently displayed. Key 5 will display the previous eight Instructions. Key C will allow the base address to be changed. To do this press key C, the screen will go blank, then press a key between 0 and F. That will then be the new base address.

If you press any other key the screen will go blank; press any correct key to bring back the display.

MEMORY DISPLAY UTILITY

0F00	6E10	1F0A	F50A	00E0
0F08	450F	2F1A	450D	2F26
0F10	4505	2F20	450C	2F7A
0F18	1F04	6700	2F26	00 EE
0F20	87E5	2F26	00EE	6B00
0F28	6A00	4B30	00EE	2F34
0F30	7B06	1F28	6F00	8070
0F38	8074	6906	4F01	7901
0F40	F929	DAB5	7A06	2F5C
0F48	7A08	A600	F71E	F71E
0F50	F165	7701	2F5C	8010
0F58	2F5C	00 EE	8200	63F0
0F60	8232	6300	82E5	7301
0F68	3200	1F64	F329	DAB5
0F70	7A06	F029	DAB5	7A06
0F78	00EE	FOOA	AF3B	F055
0F80			0	





at the leading edge

CMOS SINGLE CHIP 1200 BAUD MODEM

(U.K.) The FX409 from C.M.L. is hailed as the first such device to conform to the European mobile radio data link standard. This 22 pin part operates in full duplex and its low power consumption makes it suitable for hand held radio communicators.

ECONOMY 5V EEROM

(U.S.A.) General Instrument has dubbed its 128 x 8 EEROM "a challenge to the dip switch". At around \$5.60 in OEM quantities, they could be right. G.I. predicts we will see them in diallers, terminals, locks, traffic lights, keyboards... the list goes on.

SHARP 5KV OPTOS GET NOD FROM TELECOM

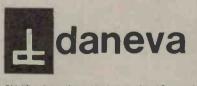
(MELBOURNE) Approval No RA83/107 is good news for opto users and even better news for former Fairchild fans. Sharp has published a handy cross reference to cover the gaps left when the big 'F' quit the coupler game.

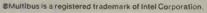
WINCHESTER CONTROLLER SPORTS SASI BUS

(U.S.A.) Western Digital Corporation are also building their brilliant WD 1002 Winchester/Floppy controller with ®Multibus and STD interfaces. At around \$388 for SASI and \$950 for Multibus, engineers will save heaps on development dollars.

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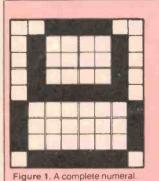
But if you load this program into your Microbee you'll be able to measure audio tones up to 12 kHz or so, with the frequency displayed in nice big characters on your television screen.

What can you use it for? Are you building the ETI-905 polyphonic organ? You'll need a frequency counter to set it up. A computer modem? An RTTY modulator? Testing your stereo? A frequency counter is almost essential for all of these projects.

Now for some details of how it works...The frequency to be measured is fed into the Microbee's cassette port via the interface shown in Figure 3. You insert the plug that normally goes to the 'earphone' socket into the interface socket instead. The tone is applied at the other end. Perhaps some enterprising manufacturer will offer the interface as a kit.

Within the Microbee is an IC amplifier/protection circuit that squares up the signal and presents it to bit 0 of data port 2. The program inputs the contents of the port and compares it with the result of the previous input. If bit 0 went low the carry flag is set. The carry flag, regardless of its state, is rotated into the 1° position of the BC register, which is then added to HL, the frequency count. So HL increments for each cycle.

Meanwhile register DE has been loaded with a number that takes precisely one second to count down to zero. Each time the port is tested, DE is decremented as well. When DE runs out of steam and the program exits the loop, the HL register contains a



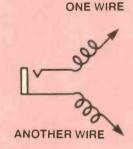
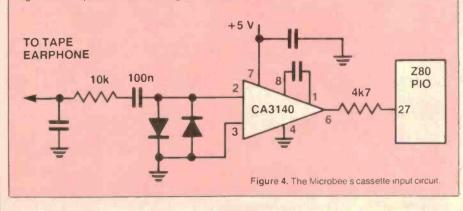


Figure 2. Graphics character format. Figure 3. The interface



OPERAND LINE LABEL MNEM CODE ADDR 00100 MICROBEE AUDIO FREQUENCY COUNTER 00110 1 - Tom Moffat 00120 DEFR 8488 00130 8488 88148 ORG LORES INT 8488 CD2788 88158 CALL 8827 88168 iGet frequency count into HL 00170 1000 +r 00100 00190 START1 00200 0403 118569 0406 210000 0409 DB02 DE,6985 ITIME FOR COUNT HL,0 INUMBER OF CYCLES A.(2) IBEGIN COUNT B ISET CARRY IF NEW CYCLE I.D. LD 00210 START 040B B9 88228 040C 010000 040F CB11 BC . 8 00230 ISHIFT IN CARRY, MAKE BC=1 OR 8 88248 RL 0411 09 0412 47 0413 10 0414 7A 0415 83 ADD HL . BC 00250 LEAVE PREVIOUS STATE 00260 00270 00280 00290 00290 I D B.A DEC DE LD OR JR A.D NZ . START : END COUNT IF ZERO 8416 28F1 00310 00320 iDisplay the result. 66336 0418 CD3E04 0410 1110F3 041E 011027 0421 CD4F04 0424 01E003 0427 CD4F04 042A 016400 042D CD4F04 CLS ICLEAR SCREEN DE.0F3101SCREEN LOCATION 00340 CALL CL S 00350 LD DE,0F31015CREEN BC,2710 110000D CONV BC,03E6 11000D CONV BC,64 1100D CONV BC,00H 110D CONV 00350 00360 00370 00390 00400 LD CALL LD CALL LD CALL LD 00410 8438 818A88 8433 CD4F84 88428 CALL 00430 CONV 0433 CD4F04 0436 010100 0439 CD4F04 043C 10C5 88448 LD BC.1 11D 00450 00450 00460 00470 00400 00490 CALL .18 STARTI :Clear the sc en routing. 00500 CLS 043E 810004 LD BC.400 043E 010004 0441 E5 0442 2100F0 0445 3620 0447 23 0448 08 0449 78 0444 01 0449 20F8 0444 E1 044E C9 00510 PUSH HL . OF000 00520 (HL),20 00530 FILL LD 00540 INC HIL DEC LD OR JR POP RET 00540 00550 00560 00580 RC A.B NZ .FILL 88598 HL 88088 91499

binary number that is the number of audio cycles that occurred in one second. The binary number is converted to decimal and displayed on the screen by a

successive subtract' division routine. Then the whole procedure repeats. The characters that appear on the screen are spe-

in the Microbee, although not widely known about. They're produced after you call the 'LORES' initialse subroutine. You then write characters to the screen with bit 7 set high.

There are six pixels in each graphics character, and 12 graphics characters are used for each numeral, giving 72 pixels for each character. Figure 1 shows, how an '8' is formed from the 12 graphics characters. Figure 2 shows how the data is developed for each graphics character.

The resulting numerals stand 35 mm high on a 300 mm (12 inch) TV monitor; they're bigger than any LEDs you'll find on a 'proper' frequency counter.

For what it's worth, this 'graphics character' business will work on a TRS-80 as well.

The accuracy of the Microbee as a frequency counter is excellent, with the timebase being derived from the Microbee's 12 MHz crystal. The quantity for register DE in line 190 has been calibrated against a you-beauty super accurate Systron Donner counter, and should hold for all Microbees if the 4 MHz crystals are any good.

Sensitivity is also quite good, about 100 mV peakto-peak at 1 kHz. At 10 kHz, it's 200 mV and at 5 Hz, 800 mV (that 100n capacitor is obviously getting in the way at such a low frequency).

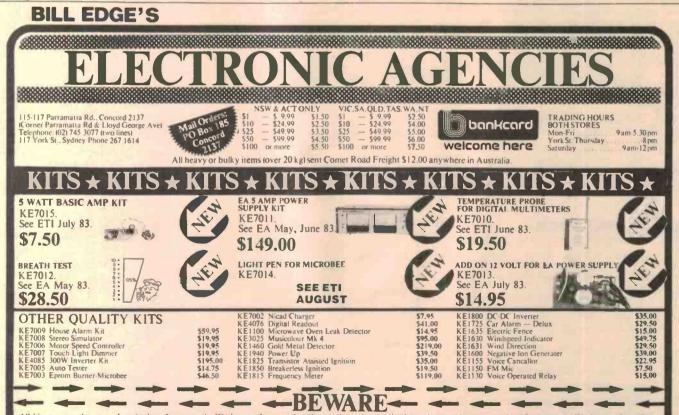
The program has been written, by the way, so you can use the clear screen and big number procedures in your own programs. They we been presented as separate subroutines.

All in all, it's not a bad little frequency counter, even though it's limited to audio frequencies.

If you don't feel like wrestling with all that nasty machine code, the usual offer applies: twelve dollars to the author brings the program on a cassette which can be loaded under BASIC, and an instruction sheet, all post paid.

MICROBEE COLUMN

ADDE CODE	LINE LABEL MINEM	OPERAND	0000 0000			
		git to decimal and show it.	ADDR CODE	LINE LABEL	MNEM OPERAND	
	00020 100010 0 010	git to decimal and show it,				
044F BEFF	08648 CONV LD	A.OFFH ICLEAR A FOR COUNT	0490 8E83	01140 THREE	DEFW 838EM	
0451 3C	00650 CONVI .INC	A	849F 834F	01150	DEFW BAF83	
8452 A7	00660 AND	A	04A1 8088	01160	DEFLI 8880	
0453 ED42.	68678 SBC	HL.BC	84A3 9C87	01170	DEFU 0878CH	
0455 30FA	88688 JR	NC, CONV1	04A5 AC80	01180	DEFU OBOACH	
0457 09	00690 ADD	HL.BC	04A7 809E	01190	DEFW YEBU	
0458 87	99799 ADD	A.A :X2	04A9 80A0	01200 FOUR	DEFW BARSO	
8459 87	00710 ADD	A.A ::X4	04AB 8695	01210	DEFW 9586	
845A 4F	00720 LD	C.A	04AD 8881	01220	DEFW 08188	
045B 87	00730 ADD	A.A 1X8	84AF 8885	01230	DEFU 08580	
0450 81	00740 ADD	A.C :X12	0481 8080	01240	DEFW 8888	
045D 0600	00750 LD	B.0	0483 8095	01250	DEFW 9580	
045F 4F	00750 LD	C.A	0485 9783	01260 FIVE	DEFW 8397	
0460 E5	00770 PUSH	HL	04B7 8381	01270	DEFW 8183	
8461 217984		HL.ZERO	0489 8083	01280		
3464 89	00790 ADD		0488 83AC	01290		
8465 8683	00800 LD	HL.BC HL POINTS TO NUMBER B.3 13 LINES OF CHARACTERS	0480 A480	01300	DEFW 0AC83 DEFW 0B0A4	
0467 C5	00810 SHOW PUSH	B, J J LINES OF CHARACTERS	048F 8086	01300		
0468 010408			04C1 A09E		DEFW SABO	
0468 ED80		BC,4 :4 ROWS OF CHARACTERS	0403 8380	01320 SIX	DEFW PEAR	
0460 78				01330	DEFW 8083	
046E C63C		A.E	84C5 878C	01346	DEFW 8CB7	
0470 5F		A. 3CH	04C7 8CB0	01350	DEFW 080SCH	
0471 CI		E.A	04C9 ADB0	01360	DEFW BBBHDH	
0472 10F3	00870 POP	BC	04CB B08E	01370	DEFW SEB0	
0474 C647	00880 DJNZ 00890 ADD	SHOW	04CD 8783	01380 SEVEN	DEFU 8387	
8476 5F		A.47 IPOINT TO NEXT SCREEN LOC'N	04CF 8398	01390	DEFW 9893	
0477 E1		E.A	04D1 80A0	01400	DEFW 0H080	
0478 C9	00910 POP	HL	04D3 8688	01410	DEFW 8086	
0470 6*	00920 RET 00930		0405 80AA	01420	DEFU BAASB	
			0407 8080	81430	DEFW 8080	
	00940 (Data for Dig	numerals, as graphics characters.	0409 A883	01440 EIGHT	DEFW 83A8	
0479 9683		000/	04DB 8394	01450	DEFW 9483	
6478 83A1	00960 ZERO DEFW	9396	84DD 968C	01460	DEFW BCPAH	
0470 9580	00970 DEFW 00980 DEFW	04193	84DF SCAS	01470	DEFU BASBCH	
0470 9580 047F 80AA		8095	04E1 A580	01480	DEFW 08045	
0481 A180	00990 DEFW	04480	04E3 8094	01490	DEFW 9AB0	
0483 809A	01000 DEFW	886A1	04E5 9E83	81500 NINE	DEFW 839EM	
0485 8088	01010 DEFW	9AB0	94E7 83AD	01510	DEFW 84D83	
0485 9580	01020 ONE DEFW	8888	04E9 838C	01520	DEFW 9C83	
0489 8088	01030 DEFW	6095	04EB SCAB	01530	DEFW BABSCH	
0488 9586	01040 DEFW	8080	04ED 8040	01540	DEFW BARSO	
0480 80A0	01050 DEFW	8095	04EF 8887	01550	DEFW 8788	
848F 8588	01060 DEFW	04080	0000	01560	END	
	01070 DEFW	8985	00000 Total e	rrors		
0491 8E83	01080 TWD DEFW	838EH				
0493 83AF 0495 A088	01090 DEFW	0AF83	NINE 04E5	EIGHT 04		\$IX 04C1
0497 8E81	01100 DEFW	08840	FIVE 0485	FOUR 84		TWO 0491
0497 8E81	01110 DEFW	818EH	.ONE 0485	SHOW 04		CONV1 0451
0498 8088	01120 DEFW	08087	FILL 0445	CONV 04	4F CLS 043E	START 8469
DALO ORES	01130 DEFW	68886	START1 0403			



All kits are not the same. Lets look at front panels. We know of two major kit suppliers who supply those cheap and nasty scotchcal front panels. We know of another two who supply aluminium panels. But only from Electronic Agencies can you expect in your kit a SILK SCREENED. DRILLED, ALUMINIUM panel. Ever tried to drill a 5 mm hole for a LED for instance and then get rid of the burrs without scratching the panel, its not easy. Our panels sell for up to \$14.50 each separately, so add that on to the cost of a kit from another supplier, and you'll see you pay no more for Electronic Agencies' quality kits.

We also include FREE in kits \$20 and over our very own KIT CONSTRUCTORS REFERENCE MANUAL, which is crammed full of invaluable information. (The exception to the aluminium panel is our new Breath Test Kit, it has a stick on panel, not aluminium). The Edge family have been in the component retail business for 50 years. We'd call it experience!

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THE VIC-20 COLUMN

There are obviously a lot of people who own VIC-20 computers. We're getting some really good programs coming In. No doubt the possibility of winning something Is an added stimulation to making a contribution. The best software item submitted to this column each month will win a VIC-20 expansion board, donated by Ozi-Soft, in conjunction with Computer Technics.

We do have an apology to make to Computer Technics, who distribute the board, as we printed their address incorrectly. It should have been Computer Technics, P.O. Box 25, Kogarah NSW 2217.

Your submissions should be original and every article published will be paid for, so send them off to the Editor now.

The first winner of a VIC-20 expansion board is (gasp, gasp! the invaders are coming) Peter Bagust, who wrote GASP.

GASP Peter Bagust, Sans Souci NSW

This is a 1.5k game program

Hostile invaders are destroying the earth's atmosphere and our only hope is the photon-guided missile hidden deep in the mountainside. Unfortunately the missile eats up air as it passes so use it carefully.

There are seven invaders, each getting faster and trickier. The game ends when all the invaders are destroyed or you have less than five per cent of air remaining.

Lines 10-50 are introduction.

Line 60 is initialisation, line 70 creates the atmosphere and line 80 makes the mountain range.

Line 90 reinitialises before each frame. Lines 100-130 are game finishes. Line 140 prints frame number and air remaining.

Lines 150-250 create and move the invader, 260-370 create and move the missile and 380-400 are sound effects.

Variables are 'F' = frame, 'A' = air left, 'M' = missile, I' = invader, 'K' = invader hit flag and 'T' = time interval.

MUSIC GENERATOR Mark Smith, Laverton Vic

The music generator allows you to play a tune by using the table of music notes and table of decimal equivalent values in the VIC—20 Personal Computing book supplied with the computer.

You simply type in the number of notes in the music then put in the decimal equivalent value for the required note. Then you decide which speakers you want on (1-4) and how long you want the notes held for.

F1 will change the speaker values

F2 lets you change the volume.

F3 allows you to change the period for which the notes are held on.

F7 changes the value of a note.

S varies the speed at which the music is played. Y allows you to play again.

Ozi-Soft, in conjunction with Computer Technics, is offering to donate a VIC-20 expansion board for the best software item submitted to this column every month.

The board is Australian designed and manufactured and simply plugs into the VIC-20's expansion stot. It features three sockets that can be independently switch-selected, plus an on-board reset switch. With it you can plug in up to three separate expansion units to your VIC-20 and avoid the hassle of plugging things in and out and turning the computer on and off each time. It is distributed by Computer Technics, P.O. Box 25 Kogarah NSW 2217 and costs \$59.95.

All submissions must be accompanied by a signed letter from you stating lhat it's your original work. The winning submission will be judged by the Editor and no correspondence will be entered into. All published submissions will be paid for.

Send entires to: The Editor, VIC-20 Column, ETI Magazine, P.O. Box 21, Waterloo NSW 2017.

GASP

10	PRINT
50	PRINT DESTROY THE INVACERS BEFORE MORE THAN 95% DOF YOUR
	AIR IS EATEN"
30	PRINT TOTOLKEY B LAUNCHES MISSILEKEYS Z AND C STEERS ITKEY M WILL
	DETONATE ITA QO"
	PRINT TO BEGIN HIT ANY KEY "FORT=1T01000 NEXT
50	PRINT TO BEGIN HIT ANY KEY "FORT=1T01000 NEXT: IFPEEK(197)
	=64 THENGOTO40
60	F=0:A=415:PRINT"""
	FORZ=1T019:PRINT"DE
	PRINT D HE T
	M=8130:1=7680:F=F+1:POKE36879,14:K=0:T=INT(75/F)
100	IFF = 8 THENPRINT " LINDING CONGRATULATIONS ALL ENEMYS
110	DESTROYED "IPOKE3687 9,204
	IFF=8THENPRINT DICAN YOU DO IT AGAIN7 ": FORL=1T010000:NEXT:RUN
150	IFAC21THENPRINT CONTINUES *** GASP *** UNDER 5% OF AIR
120	IFA(21THENPRINT" I GIVE UP YET LOSER?":FORL=1T010000:NEXTL:RUN
	PRINT "SUBMINIMUMUMUMUMUMUMUMUFRAME "F "MAIR LEFT"(INT(A/4.18))"
	R=INT(RND(1)*6):POKE1,32:I=I+1
	IFE=3ANDR(2THENI=I-R
	IFF=4ANDR<3THENI=I-R
180	IFF=5ANDR=4ANDIX8142THENI=1+22.
190	IFF = 5ANDR = 5AND 1)7702THEN 1 = 1 - 22
200	IFF JSANDM= I +22THEN I = I -22
210.	IFF SANDM=I+2THENI=I-2
220	IFF >SANDM= I - 1THEN I = I + 1
230	IFPEEK(I)=160THENA=A-1
240	IFPEEK(I)=42THENPOKE36879,138:FORL=1T0999:NEXT:POKEM,32:POKEM-1,
	32:POKEM+1,3 2:GOTO90
250	POKEI,0:POKEI+30720,F
	POKEM, 32
	IFK=1THENPOKEM+1,32:POKEM-1,32:M=8130:K=0
	IFMK B130THENM=M-22
	IFPEEK(197)=34ANDM(8108THENM=M+1 IFPEEK(197)=33ANDM(8108THENM=M-1
	IFPEEK(197)=35ANDM)8129THENM=M-22
-	IFPEEK(197)=36ANDM(8108THENK = 1
	IFK=1ANDPEEK(M)=160THENA=A-2
	IFPEEK(M)=160THENA=A-1
	POKEM, 30 1 POKEM+ 30720, 5
	IFMK 7702THENK = 1
370	IFK=1THENPOKEM-1,42:POKEM,42:POKEM+1,42:POKEM+30720,1:POKEM+30721,
	1: POKEM+30 719,1
380	IFK=1THENPOKE36877,220:FORL=15T00STEP-1:POKE36878,L:FORN=1T0200:
	NEXTN:NEXTL: GOTO400
390	POKE36878, 15 POKE36877, 220 FORL = 1 TOT : NEXTL
400	POKE36877,0:POKE36878,0:FORL=1TOT:NEXTL:GOT0110
REA	DY.

CIRCUITS 140 PRINT"DRE YOU GOING TO SAVE THIS DATA": GOSUB98: IFAS="N"THEN133 145 PRINT"WHAT IS THE NAME OF THE FILE ": INPUTBS 150 PRINT"PREPARE THE TAPE": GOTO300 THIS PROORAM?" : COSUB98 : IFAS="N" THEN123 201 FRINT"THIS IS WHAT IT SOUNDS LIKE":POKES;N(R):FORT=1TOP(R):NEXTT:POKES;0 205 FRINT"WHAT NEW VALUE FOR THE PAUSE DO YOU WANT":INPUTP(R) 210 FRINT"THIS IS WHAT IT SOUNDS LIKE":POKES;N(R):FORT=1TOP(R):NEXTT:FOKES;0 217 PRINT DO YOU WANT THE MUSIC FASTER : COSUB98: IFRS="Y"THENY=1: COTO220 218 FRINT DO YOU WANT IT SLOWER?": OOSUB98: IFAS= "Y" THENY=0: 00T0220 220 PRINT"BY WHAT FACTOR DO YOU WANT THE CHANGE" : INPUTO : 1FY20THEND-1/G 225 PRINT "WHICH NOTE DO YOU WANT TO CHANGE" : INPUTU: IFUDATHEN225 230 PRINT"THIS IS WHAT IT SOUNDS LIKE": POKES,N(U):FORT=1TOP(U):NEXTT:POKES,2 235 PRINT"WHAT IS THE NEW VALUE THAT YOU WANT": INPUTN(U):IFN(U):253THEN235 240 PRINT"THIS IS WHAT IT SOUNDS LIKE":POKES,N(U):FORT=1TOP(U):NEXTT:POKES,2 **ETI CIRCUITS**

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ETI July 1983 - 59

18 INPUT"VOICE #1";K1:INPUT"VOICE #2";K:GOTO21LIST20 20 INPUT "WHICH VOICE" ;K 21 INPUT VOLUME ";V 25 PRINT"ANY PRUSES?" : GOSUB98 30 IFAS="Y"THEN100 35 FORT=1TOA:P(T)=200:NEXTT:B0T0125 40 FORT=1TOA:P(T)=P:NEXTT:00T0125

100 PRINT"IS IT THE ONE PRUSE" : GOSUB98

120 FORB-ITOR PRINT PERIOD #"; B; INPUTP(B) NEXTB

127 POKES, 0: POKES1, 0: BOSUB330: NEXTC: POKES, 0: POKES1, 0

10 PRINT "THOW MANY NOTES WILL YOU REQUIRE" : INPUTA

17 PRINT"DO YOU WANT HARMONY?": GOSUB98: IFAS-"N"THEN20

15 PRINT NOTE #"; B; INPUTN(B) IFN(B)255THEN15

DATA?": GOSUB98: IFA#="N"THEN18

2 PRINT "WHAT IS THE FILE NAME" : INPUTBS : PRINT "PREPARE THE FILE" : GOSUB310

126 FORC=1TOR: POKE36878, V: POKES, N(C): POKES1, N(C): PRINTC; " "; N(C), O*P(C): FORQ=1TO

130 PRINT"RUN RORIN? XTEREFISER F1-SPK, F3-VOL, F5-PRU F7-NOTE, Y-YES, S-SPD"

98 GETAS: IFAS=" "THEN98

125 S=36873+K:S1=36873FK1

134 IFA\$=CHR\$(136)THEN225 135 IFAS="Y"THEN125 136 IFR\$="S"THEN217

131 GOSUB98: IFR\$=CHR\$(133)THEN260 132 IFR#=CHR#(134)THEN278 133 IFR\$=CHR\$(135)THEN200

155 PRINT NORDO YOU WANT TO END

200 INPUT "WHICH PRUSE"; R: IFR)ATHEN200

215 PRINT "ANY MORE?" : COSUB98 : IFAS "N"THEN130

245 PRINT "RNY MORE?" : BOSUB98 : IFAS= "Y"THEN225

310 OPEN1, 1, 0, 05 312 PRINT"FILE "; 05;" FOUND": PRINT"LOADING"

260 INPUT"D VOICE #1";K1:INPUT"VOICE #2";K:00T0130 270 PRINT"DHOW LOUD IS THE SOUND REQUIRED '0-15" INPUTY

304 PRINT#1, R: FORF=ITOR: PRINT#1, N(F): PRINT#1, P(F): NEXTF

314 ZZ=R:FORM=1TOZZ:INPUT#1,N(R):INPUT#1,P(R):NEXTR 318 FORC=1TOR:PRINTC;"_";N(C),P(C):NEXTC

105 IFAS="N"THEN120 110 INPUT"PRUSE"; P:00T040

MUSIC GENERATOR

11 DIMN(R), P(R):0=1 14 FORB-1TOR

3 GOTO125

16 NEXTB

99 RETURN

OMP(C) NEXTO

160 ENT

216 8070289

219 0010125

221 0070130

250 GOTO130

275 0010130

308 CLOSE1 : END

334

336

300 OPEN1, 1, 1, 85 302 PRINT"SAVINO ";85

313 INPUT#1, A: DIMN(A+1), P(A+1)

320 0=1:V=15:K=2:K1=3:RETURN 330 0ETAS: IFAS=""THENRETURN

IFRS=CHR\$(135)THEN200 335 IFRS=CHR\$(136)THEN225

332 IFAS=CHR\$(133)THEN260

333 IFA#=CHR#(134)THEN270

IFAS="Y"THEN125 337 IFR#="S"THEN217 338 RETURN READY.

1 PRINT DO YOU REQUIRE OLD

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ZX COLUMN

TYPOGRAFIX

This program can be used on an unmodified ZX81 or ZX80 8K ROM, plus 16K RAM and a ZX printer, printing up to 32 lines of text containing both upper and lower case characters.

When the lines have been entered type 'RUN' to initialise. A string prompt (" ") will appear and a max-imum length of 42 characters may be entered each time.

Z ... 1... 9

(Note: represents a space, A = inverse A, 1 = inverse 1, etc.)

Hence, the normal output is lower case letters and numbers 0-9 and the inverse symbols are used to obtain capital letters and other symbols.

After each string is entered approximately 15 seconds are needed to place the appropriate data from C\$ into the array A\$. A total of 32 strings must be

TYPOGRAFIX 1 REM UMAND7/UMANDER ?TAN 2 IF PEEK 16355+255+PEEK 1635 9=31744 THEN GOTO 5 3 POKE 16389,124 4 GOTO 1 5 FOR I=0 TO 112 6 POKE 31744+I,PEEK (2161+I) 7 NEXT I 6 POKE 31857,201 10 POKE 16524,79 11 POKE 16524,79 12 DIM A#(32,256) 13 LET C\$="0620970890690620000 02127000000090381073073070067065 050000570600746730730480970170090 0300 14 LET C\$=C\$+"1120640641200640 TYPOGRAFIX 004002" 17 LET C\$=C\$+"0000000360360000 00001031030000000176112000000 000132113000000000176112000000 21270420180200200106100900603604 21270420180200200200200200540730 35034080000020254065065000065085 054002" 054002" 13 LET C\$=C\$+"125009009091251 27073073073054062065065065034127 05506506506212707307306506512700 0000010010620650730411221270050 00000127000065127065000032054064 19 LET C\$=C\$+"127006020120540654 19 LET C\$=C\$+"127006020120340651 2705405405405424127002012002127127 02200400512706206506506506212700

Ralph E. Morgan, Killarney Heights NSW

entered before printing commences; use a space followed by 'NEWLINE' to quickly 'fill in' any excess lines. Lines 1-12 are machine code for the hi-res printing

facility (see ZX printer manual). Lines 13-19 accumulate the data string CS. Each letter is stored as five consecutive three-digit decimal numbers which are interpreted as follows:

character	decimal	binary	
	062	00111110	
	097	01100001	
.0.	089	01011001-	
	069	01000101	
	062	00111110	

Hence the user can substitute his/her own symbols if desired.

Lines 20-35 calculate x and y coordinates of the symbol to be printed. Lines 49 and 50 enter data into AS. Lines 80-120 find the appropriate data in CS. Lines 9988-9999 print out AS (see ZX printer manual).

900900900606205081030941270090 25041070006073073070050001001127 00100106306406406406301504806404 80151270320240321270902000060200 99003012112012003097081073069067
00012705505500000005505512700000 50200340550000000055034020006" 20 FOR L=1 TO 32
21 LET X=33-L 22 INPUT T\$ 23 IF LEN T\$>42 THEN GOTO 22 24 LET N=LEN T\$
24 LET N=LEN T\$ 25 IF N=0 THEN GOTO 65 30 FOR I=1 TO N 35 LET Y=(I-1) +6
40 GOSUB 60 45 FOR J=1 TO 5 49 LET Z=15*(PO5-1)+3*(J-1)
50 LET A\$[X, Y+J]=CHR\$ (((CODE C\$(Z+1)-23)+100+(CODE C\$(Z+2)-23))+10+(CODE C\$(Z+3)-28))) 55 NEXT J
60 NEXT L 65 NEXT L 70 GOSUB 9988
75 STOP 80 LET M=0 85 LET C=CODE T\$(I) 90 LET C=CODE T\$(I)
90 IF C>=128 THEN LET M=128 91 IF C=0 THEN GOTO 60 95 IF M=128 AND C<153 THEN LET P05=C-M+57
100 IF M=0 AND C<28 THEN LET PO S=C+16 105 IF M=128 AND C>=153 THEN LE
T POS=C-M+13 110 IF M=0 AND C>=28 THEN LET P 05=C-27 115 IF POS(1 OR PDS)80 THEN LET
POS=1 120 RETURN 9988 REM PRINT A\$
9389 FOR 1=0 TO 246 STEP 8 9990 FOR J=1 TO 32 9991 FOR K=1 TO 8 9992 POKE 32255+K+8*(J-1),CODE A
\$(J.K+1) 9993 XEXT K 9994 XEXT J
9995 FOR H=0 TO 31 9996 POKE 16444+H,H 9997 NEXT H 9998 LET HPRINT=USR 31744
9998 LET HPRINT=USR 31744 9999 NEXT I

ENCODER

This program accepts a text or numerical data string and then encodes it using a coding key entered by the user

The key is then erased from RAM and the program

ENCODER

20	PRINT "ENTER STRING"
30	CLS PRINT "ENTER REY"
	LET A=0 CLS LET A=A+CODE 65
50	IF LEN BS=1 THEN GOTO 100 LET BS=BS(2 TO)
100	GOTO 50 Rem Encryption Rand A
115	LET N=LEN AS FOR I=1 TO N LET X=CODE AS(I) + INT (RND+1
00) 119 120	IF X>255 THEN LET X=X-256
125	

Ralph E. Morgan, Killarney Heights NSW

can be saved with the encoded data. When the program is later reloaded the data can be retrieved by entry of (GOTO150) and the correct key. Hence improper retrieval is rendered impossible.

```
135 LET B$=""

137 CL5

138 LET A:0

139 PEINT A:

130 STOP

150 STOP

150 REH DECIPHERING

155 REINT "ENTER KEY"

160 INPUT B$

175 LET A:0

175 LET A:

160 IF LEN B$=1 THEN GOTD 200

185 LET B:

200 RAND A

204 FOR I:

200 RAND A

206 RAND A

206 FOR I:

200 RAND A

207 RAND A

208 RAND A

2
```





CBC Bank Bidg, 661 George Street, Haymarket, Sydney. (02) 212 4815

80 4654, TAS: D & A Agencies 23 2842, QLD: Baltec Systems (07) 36 5183; Solex Electronics (07) 72 2015.

Equipment NEWS

Open-frame switching power supplies

Power-One has gone into full production with its International Series open-frame switching power supplies, distributed in Australia and New Zealand by Warburton Franki.

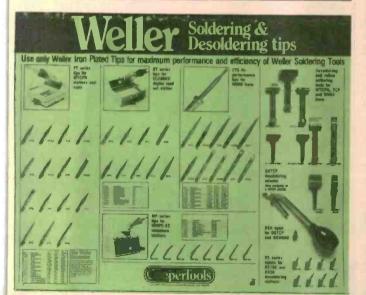
This new line incorporates the latest state-of-the-art switching technology, in eleven off-the-shelf models.

Five package sizes are available, ranging from 40 W to 250 W. Depending on the model selected, the International Series offers up to five outputs, selected to accommodate applications such as small computers, terminals, peripherals, word processors, disk drives, Winchesters, printers and other devices using microcomputer technology.

The new switches are designed to meet VDE, UL, IEC, CSA as well as most other world-wide regulatory safety agencies. They also meet the emissions limits of FCC Docket 20780 Class A and VDE 0871/6.78 Class A.

Another feature is the ability to operate from the wide variety of ac input voltages found throughout the world. The International Series ac input ranges include, as standard, both 90 V and 132 V for domestic and Asian applications, and 180 V to 264 V for European applications.

For further details contact Warburton Franki, 372 Eastern Valley Way, Chatswood NSW 2067. (02)407-3261.



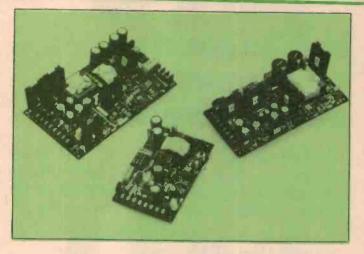
Weller's wall poster

Cooper Tools, the manufacturer of Weller soldering equipment, has published a colour poster of its Weller soldering and desoldering tips.

The poster depicts — by number, temperature and range

- all Weller tips available for the WTCPN, EC2000D and WMCP-EC temperature-controlled soldering stations and the W60D line-voltage temperaturecontrolled soldering iron. Accessories are also featured.

For a complimentary copy of the poster contact the Cooper Tool Group, P.O. Box 366, Albury NSW 2460. (060) 21-5511.



Tektronix logic analysers

The new Tektronix 1240 logic analyser is now available in configurations which range in price from \$10 500 to \$28 000.

A card-modular mainframe lets the user select the data acquisition channels and speeds, data analysis and communications interface support. The mainframe accepts up to four data acquisition modules; the 1240D1 has nine channels and acquisition speeds up to 100 MHz and the 1240D2 has 18 channels and speeds up to 50 MHz.

The 1240 has a dual timebase and sophisticated triggering and software analysis tools which are optimised for a broad spectrum of software debug problems.

The functions are presented as rectangular graphic fields containing the menu label, each selectable by touching the appropriate soft key on the display.

The Sony/Tektronix 318 and 338 logic analysers are additions to the 300 series of 'ultra portable' logic analysers.

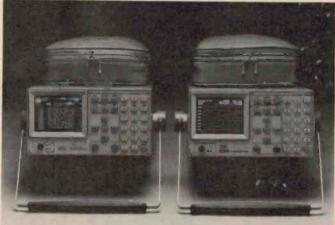
The 318, priced at \$8190,

provides 16 channels of data acquisition at up to 50 MHz. The 338, priced at \$8960, delivers 32 channels at up to 20 MHz. They both provide glitch capture and three levels of triggering.

In the serial mode both synchronous and asynchronous data acquisitions can be executed at baud rates from 50 bps to 19.2 kbps. Captured data can be displayed in hex, binary, octal, ASCII or EBCDIC which make these analysers ideal for such applications as testing line links, analysing protocols and data checking in local area networks.

An option which includes an RS-232 serial interface and nonvolatile memory is priced at \$1860.

More information about these logic analysers can be obtained from Tektronix Australia Pty Ltd, 80 Waterloo Rd, North Ryde NSW 2113. (02)888-7066.



Component NEWS

H-P expands 10-element bargraph array

Hewlett-Packard has added a new highperformance green 10-element array to the standard-red, high-efficiency red and yellow 10element LED array family.

In addition, prices have been reduced by up to 20% for the HDSP-4800 bar-graph array family, which achieves both analogue and digital indication.

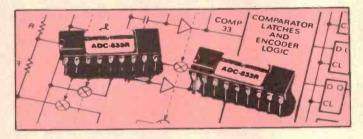
H-P bar-graph arrays are very bright, typically 1.2 mcd/segment at 10 mA for high-performance green, and also feature high segment-to-segment and on/off contrast. The segments are large (5.08 mm x 1.52 mm), which gives a wide viewing angle.

An exclusive feature is the

package interlock, which ensures correct alignment for the longer arrays. A low package profile saves space between the display board and filter.

Yellow and green bar-graph arrays are categorised for dominant wavelength and all colours are categorised for luminous intensity. This assures a consistent front-panel appearance.

For more information contact Hewlett-Packard Australia Ltd, 31-41 Joseph St, Blackburn Vic. 3130.(03)890-6351.



Six-bit video flash A/D converter

The ADC-833 available from Datel-Intersil is a low power, six-bit video flash analogue to digital converter.

It can digitise an analogue signal at conversion rates up to 15 MHz with a power consumption of 200 mW. Two ADC-833s may be connected in parallel to increase the conversion speed from 15 MHz to 30 MHz.

The ADC-833 is available for operation over the industrial, -25°C to +85°C temperature range and is packaged in an 18-pin ceramic DIP.

The analogue input voltage range is +2.5 V to +10 V with minimal linearity error.

The digital outputs are buffered three-state and include an overflow output which allows the user to cascade two units to achieve seven-bit resolution. The buffers are controlled by two enable signals with a typical output enable delay of 20 ns.

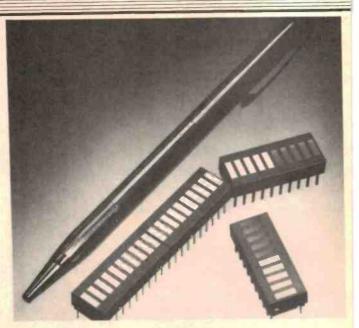
More information is available from Elmeasco Instruments Pty Ltd, 15 Macdonald St, Mortlake NSW 2137. (02) 736-2888.

Telcon takes over Amtex Electronics

National Electrical Wholesaler, Telcon Australia Pty Ltd, has announced the acquisition of Amtex Electronics.

The range of products includes photovoltaic systems and components, switching power supplies and high voltage power supplies. Telcon is located at **36 Lisbon**

St, Fairfield NSW 2165. (02) 727-5444.



Tabtek's circuit-design course

Tabtek Electronics is to hold a two-week full-time seminar, covering all aspects of both NMOS and CMOS small scale to very large scale digital integrated circuit design.

To be held from August 15-26 at the University of Adelaide, the program of lectures, tutorials and project work will aim to teach participants how to rapidly implement digital systems in silicon.

The course outline is given under the headings: NMOS VLSI Design; CMOS Logic Design; CMOS Processing; Design Rules and Scaling; VLSI Design Methods; Subsystems and Chip Floor plans; Interfacing; Design Automation and Chip Assemblers; Simulation; VLSI Testing Methods and Reliability.

General Manager for National Semiconductor

National Semiconductor (Australia) Pty Ltd has announced the appointment of John Eccleshall as General Manager.

Mr. Eccleshall will report to Curtis Reid, Managing Director.

In this newly created position, Mr. Eccleshall has responsibility for the component marketing operations in Australia, in addition to the company's research and development activities.

National Semiconductor (Australia) customises microprocessors to the specifications of the growing number of Australian manufacturers.

The principal lecturers will be Dr K. Eshraghian and Dr D. A. Pucknell; both are well-known as the authors of a series of lectures in VLSI system design. They will be supported by Tabtek engineers.

The \$850 course fee covers design notes and use of the computing facilities, as well as luncheon and morning and afternoon tea.

For further information contact the VLSI Systems Manager, Tabtek Electronics, P.O. Box 49, Rundle Street, Adelaide SA 5000. (08) 223-7267.

Texas Instruments Semiconductor guide

A new master selection guide of Texas Instruments' semiconductor products is available free of charge from Texas Instruments distributors, VSI Electronics and Rifa.

The 128 page guide covers all product areas including memory, digital logic, linear, telecommunications and microcomputers.

More information about the guide can be obtained from Texas Instruments Australia Ltd, Semiconductor Marketing, P.O. Box 106, North Ryde NSW 2113. (02)887-1122.

Audio remote control products

J A Wells Electronic Distributors has introduced a range of electromechanical audio frequency notch filters and tuning fork oscillators.

Both products, manufactured by the Swiss company Institut Straumann AG, are based on a tuning fork made from the temperature independent steel alloy Nivarox (as used in Swiss watches).

The tuning fork oscillator, OSC-204, is a low distortion oscillator meeting MIL specifications for shock vibration and environment. It is available at select frequencies within the range 1000-6000 Hz, with special frequencies from 960-6400 Hz on request. Ancillary modules for frequency division are also available.

The tuning fork is temperature compensated from -55° C to 85° C. Supply voltage is selected between 5 and 15 V.

The F-304 electromechanical filter employs two gas coupled metallic precision tuning forks with electromechanical transducers and is available at any frequency within the range

FILTRE F-304 H/2 1488 M M CHI-4437 Weldenburg - 84 800-5000 Hz. The centre frequency tolerance may be specified from 50 ppm to 500 ppm with a temperature tolerance ± 1 ppm/°K to ± 5 ppm/°K.

We had the opportunity to briefly review the F-304 filter and a 1488 Hz OSC-204 oscillator here at ETI and can report that both performed very well indeed.

The F-304 had a -3 dB bandwidth of 2 Hz (at 20°C) and an insertion loss of only 9.5 dB (quoted spec. - 12 dB, +2, -3 dB). The centre frequency was better than 100 ppm (limit of our measurement).

The OSC-204, measured at 20°C was within 100 ppm (spec. $-\pm100$ ppm) and the frequency variation for a supply voltage variation from 5 V to 10 V was only ± 0.1 Hz.

The 'standard' F-304 sells for \$170 (+S.T.) in quantities of 10-19, while the OSC-204 quality code C (pre-aging, greater than 100 hrs) sells for \$62 (+S.T.).

Enquiries to J A Wells Electronic Distributors,8 Rainsford Court, Dingley, Vic. 3172. (03) 551-5979.



ILP's latest transformer

ILP Electronics, of England, has added a 15 VA transformer to its range. It is fully encased in an ABS shell, with easy fixing by an M4 bush at the base.

Test runs have confirmed the demand for both the lower VA rating and the encased toroid, and the company is now planning to extend the facility to cover transformers up to 120 VA.

For further information, contact the Australian agent, Electromark, 40 Barry Avenue, Mortdale NSW 2223. (02)533-4896.







A laboratory standard function and pulse generator Part 1

Here is another of our laboratory standard test equipment projects. This function and pulse generator covers the range from 1 Hz to 1 MHz, generates sine, square, triangle ramp and pulse waveforms. It features digital readout and six output voltage ranges, from 10 V down to 30 mV peak-to-peak. It has positiveand negative-going pulse outputs and the pulse width can be set from one second to 100 ns in seven ranges. In addition, it can be swept by an external sweep generator (to come ...).

APART FROM a good multimeter and a variable, protected power supply or two, every electronics workshop worthy of the title, and every electronics enthusiast of serious intent, *needs* a function generator of some sort with performance adequate for the various tasks the operator is likely to engage in.

Now, that's a pretty vague specification These days, the 'various tasks' one can tackle might range from amplifier system performance checks to the design and construction of a microprocessor system. At some stage of the task being tackled, a signal source of some description will be a necessity, like as not. I had to set some sort of performance specification when setting out the initial requirements of this project and cast around for a good starting point.

The best starting point I could find was right under my nose: the Wavetek function and pulse generator we've had in the lab here at ETI for some years. Now, Wavetek is to function generators what Tektronix is to oscilloscopes, so I figured our generator was a fair place to start the process.

Our model 166 Wavetek covers 0.0001 Hz to 50 MHz. It has sine, triangle, ramp, square and pulse outputs (TTL level, positive- and negative-going). The function output can be varied from 0 to 30 V peak-topeak (open-circuit, 15 Vp-p into 50 ohms). The output dc offset can be varied over ±10 V open circuit, ±5 V into 50 ohms. The frequency can be swept over a 1000:1 range, log. or linear. The output can be amplitude modulated and the waveforms can be triggered. On sinewave output, the distortion is less than 0.5% between 10 Hz and 100 kHz, rising outside those limits. The triangle linearity is greater than 99% between 0.005 Hz and 100 kHz.

The pulse output can be triggered, doubletriggered, gated or swept. The pulse period is variable between 20 ns and 10 000 seconds while the pulse width is variable between 10 ns and 100 ms in seven ranges. The transition time is variable, too, from 7 ns to 50 ms in seven ranges. The amplitude, dc offset and frequency stability in linear mode (to 500 kHz) is $\pm 0.05\%$ over 10 minutes, $\pm 0.25\%$ over 24 hours.

They are the main performance parameters of the Wavetek 166. It has a circular frequency-setting dial with calibration markings around the skirt. The dial frequency also indicates the start frequency of a sweep range. It has served us extremely well to date. However, quite a few of the performance functions and features have not been needed.

David Tilbrook

The next question I posed was, what functions and features of the Wavetek could be done without or would not be required by the serious enthusiast or general electronics workshop?

For a start, the frequency range is probably far too wide. I settled on 1 Hz as being a practical lower limit after considerable discussion among staff and associates. This is low enough for a great many *slow* digital operations and in linear applications is good for checking loudspeaker drivers for 'poling', etc. The upper limit, I knew, would be somewhat dictated by the sort of technology I would be restricted to using, given the restraints of component availability and cost. Without looking at the latter too closely, I thought an upper frequency range of 1 MHz was a desirable goal.

Function outputs	sine triangle square sawtooth
Frequency range	1 Hz to 1 MHz in seven ranges (square to 100 kHz only)
Pulse outputs	positive going TTL level with pullup to 5 V (repetition rate set by frequency control, 1 Hz to 1 MHz)
Pulse width range	1 sec. to 100 ns in seven ranges
Frequency display	31/2-digit, 1 Hz gate time, 1.000 Hz to 1.000 MHz
Sinewave distortion	typically less than 2% THD; diode shaped
External sweep input	sweeps generator over one-decade range; 1 V peak
Output voltage ranges	10 V, 3 V, 1 V, 300 mV, 100 mV, 30 mV (peak-to-peak)

GENERAL SPECIFICATIONS - ETI-166 FUNCTION/PUL SE GENERATOR

function/pulse generator

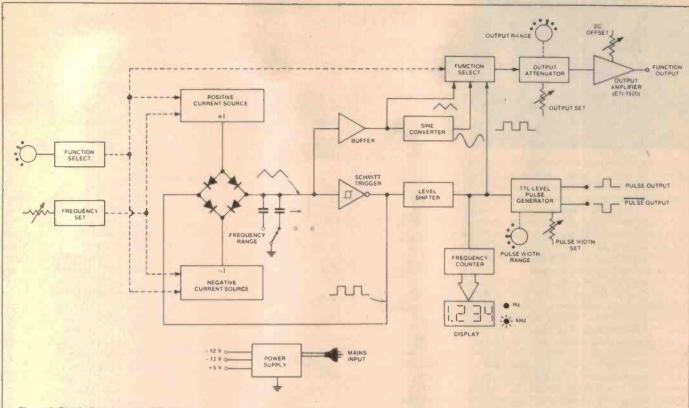


Figure 2. Block diagram of the ETI-166.

Two constant-current sources, +1 and -I, are at the heart of the main oscillator. A diode 'steering' network, diodes A-B-C-D, allows +1 to charge the Integrating capacitor and -1 to discharge it. The frequency range switch selects the integrating capacitor required. The oscillator works like this: let's assume the Schmitt trigger output is high (+ve) to start with. Diode D will be reverse-biased and diode C will be forward-biased, reverse-blasing diode B. Diode A is forwardbiased and thus the integrating capacitor will be charged by +1.

The voltage on the capacitor will rise linearly until it reaches the upper threshold of the Schmitt trigger. The Schmitt's output will then swing negative, reverse-biasing diode C and forward-biasing diode D. This reverse-biases diode A, and diode B will thus be forward-biased, allowing -1 (a current sink) to discharge the capacitor.

The voltage on the capacitor will fall linearly until it reaches the lower threshold of the Schmitt trigger, whose output will then revert to the high state, commencing the process once agaIn.

As the diagram shows, the capacitor voltage is a triangular wave, while the Schmitt trigger output is a square wave.

By varying the actual current sourced and sunk from +1 and -1, the oscillator frequency can be varied. Thus, voltage or current control of the frequency can be incorporated and the frequency set control is a potentiometer. To produce a savtooth wave, the charge and discharge currents have to be different.

For a ramp-up wave, the Integrating capacitor is discharged quickly by increasing the current sunk by -I. For a ramp-down wave, the capacitor charge current supplied by +I is increased.

This is achieved by the function select control.

The capacitor voltage signal is buffered and the triangle/sawtooth wave passed to the function select module. To produce a sinewave, the buffered triangle wave is passed to a sine converter, whose output passes to the function select module.

The Schmitt trigger output is passed to a level shifter so that the square wave signal swings between 0 V and the +ve supply rail. The frequency counter takes its input from this point, as does the TTL-level pulse generator module. The latter produces complementary pulse outputs with very fast rise and fall times. This employs a TTL monostable multivibrator, a switch selecting appropriate capacitors for the pulse width range and a pot. providing pulse width set.

The output of the function select module passes to the output amplifier via an attenuator. The output amp, includes a dc offset adjustment that can be varied with a pot.

Sweep-generation circultry has not been included in the ETI-166, simply because we couldn't fit it in a reasonably-sized off-the-shelf case! Apart from that, we realised that a stand-alone sweep generator could be used with many of the commercially available signal generators so would be suitable as a project in its own right.

Constructionally, the ETI-166 has been divided into logical modules — the main generator, the frequency counter and display and the output amplifier. The latter has been dubbed Project 1520 and appears elsewhere in this issue. The other modules will appear in following issues.

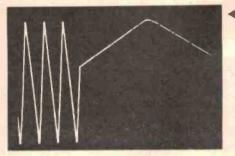
What about the function outputs? They were obvious: sine, triangle, square, and ramp-up/ramp-down. The pulse outputs were a little trickier. The general consensus was that it would be desirable to set the pulse width over a wide range and that variable transmission time was rarely required. As TTL level outputs with resistive pull-up suit both CMOS and TTL circuitry, TTL level outputs only were settled on. Complementary outputs (negative- and positive-going pulses) were considered to be desirable.

The function output level, and the desirability of dc offset adjustment, came in for much discussion. Should the generator be capable of driving a 50 ohm load at some tens of volts peak-to-peak, or was some other arrangement tolerable and easier to achieve? A search through the data books and a little quick experimentation soon settled the question. The former was certainly possible and considered generally desirable.

The goal was to get a maximum output level of 10 Vp-p into 50 ohms (20 Vp-p open circuit) right across the frequency range — a tall order, because the output stage is required to deliver around a quarter of a watt. In addition, it would need to have a bandwidth of at least five times or more than the frequency range of the generator to accommodate the harmonics that go to make up the non-sinusoidal functions.

The ability to vary the output range between zero and a defined maximum is desirable. The Wavetek has an 'output attenuator' variable in 20 dB steps, but from experience we had found the intervals too great on many occasions so I settled on having 10 dB steps for the output attenuator on this generator. As the dc offset facility had come in handy - particularly when working with digital circuitry - provision had to be made for that, too. Much digital work is done with high-level signals, while much audio work is done with low- and high-level signals, ranging from millivolts to volts. The output attenuator has to cope with those requirements.

Project 166 -



Then came sweep. The ability to sweep a generator over a given frequency range, particularly where you can preset the 'start' and 'stop' frequencies, can be extremely handy. It might be something rarely used, but when you need it, there's no satisfactory substitute. Some applications, particularly in the audio field, require sweep facilities often. There was no getting away from it the sweep facility was necessary.

Triggered and gated signals are used less often, except in certain specific fields — such as loudspeaker evaluation. It was decided that this generator could do without the nicety of triggering and gating facilities.

An important consideration on any generator is frequency readout. There are two fundamental ways one can approach this the mechanical way or the digital way. The mechanical way has the advantage of being cheap and pretty direct. The Wavetek 166 has a calibrated skirt on the frequency knob. That's fine, except in those (increasingly numerous) applications where you need to know the frequency to, say, 1 Hz in 1 kHz. Out with the digital counter! That's all right, except where you need to use the counter for other things while you're using the generator.

I opted for digital frequency display. This gave me an additional option — the ability to use a 10-turn potentiometer in place of a conventional pot. for the frequency setting control. You don't have to, but the possibility is there.

The basic functions and features settled on turned out to be:

- 1Hz to 1 MHz frequency range
- sine, triangle, square, sawtooth (ramp-up and ramp-down) and pulse outputs
- TTL level positive- and negative-going pulse outputs with variable width
- sweep facility
- output voltage range from millivolts to at least 10 V
- dc offset provision
- digital frequency display.

As the project was to be part of the 'labstandard' series, I next had to define 'lab-standard' with regard to function generators.

The definition

This proved a difficult task as there are conflicts between *desirable* performance and *reasonably achievable* performance. It's like saying a desirable goal for a jogger would be to jog up Mount Everest and back, but it's one

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 A real smoothie. Output of the triangle wave generator at 1 MHz; expanded trace on the right.

that is hardly reasonable to achieve. Keeping that in mind, I set about putting numbers to the various performance figures required of the generator project.

An obvious first one is sinewave distortion. Wavetek's lower cost, lower performance function generators quote a sinewave distortion figure of 1% to 500 kHz for a generator going to 5 MHz. Other brands quote 2% to 5% distortion figures. If you

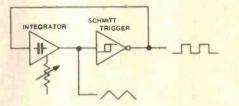


Figure 1. Fundamental arrangement of a function generator oscillator.

want to measure the total harmonic distortion of an audio amplifier, then you really need a special low-distortion oscillator. Thus, it seems to me that a sinewave output distortion of 1% to 2% would be tolerable, if not more than adequate.

Triangle and sawtooth linearity should be at least 95%, preferably close to 99%, right across the frequency range.

Frequency readout is important. When messing around with audio filters, modems and the like, the ability to measure at least 1 Hz in 1 kHz is highly desirable. A simple 3½-digit counter will allow you to read 1 Hz in 2 kHz (for argument's sake . . .), which is quite sufficient in a huge variety of applications. If you want more accuracy than that, then it's time to dig out your six- or eightdigit counter!

The output attenuator, as stated earlier, needs to be able to provide signals down in the millivolt region on one hand, yet the maximum output level desirable was set at 10 Vp-p. Then there's the 10 dB steps. Well, all that's pretty simple. A six-range attenuator will give steps of 10 V, 3 V, 1 V, 300 mV, 100 mV and 30 mV. A pot. simply allows infinite variation over whatever maximum level range is set on the attenuator.

The pulse width range was not so difficult to define. From experience, pulses greater than one second wide are not often encountered or required, so that fixed a reasonable upper limit. Pulses around 100 nanoseconds wide are encountered more often, as well as a whole host in between. A vernier control would permit variation up to a set maximum pulse width. Thus, I made the pulse width ranges 100 ms-1 s, 10 ms-100 ms, 1 ms-10 ms, 100 us-1 ms, 10 us-100 us, 1 us-10 us and 100 ns-1 us.

For the sweep function, being able to sweep over a three-decade range (1000:1) is desirable, but problematical to achieve. I settled for being able to sweep the generator over any one-decade range.

We found the modulation facility of the

lab. Wavetek was rarely used. That sort of thing's essential on an RF signal generator (..., all right, all right — we'll get round to it), but of rare application in a function generator such as this. Hence, no modulator.

I decided that providing gating facilities on the function output was a specialised application. If you need it, then the ETI-124 Tone Burst Gate (Nov, '75) should suit most applications.

The technology

The fundamental circuit of a function generator is shown in Figure 1. This consists of an op-amp integrator followed by a Schmitt trigger with a feedback path between the Schmitt trigger's output and the integrator's input. Such a circuit is arranged such that, at power-on, the Schmitt trigger output goes hig and charges the integrator's capacitor. The integrator's output rises linearly until the Schmitt trigger's upper threshold is reached, where the output goes low. The integrator's capacitor then discharges linearly until the Schmitt trigger's lower threshold is reached, where the output reverts to the high state once again.

The variable resistor in the diagram is there to indicate that the integrator's charge and discharge rate can be varied (the rates will be the same), thus varying the oscillation frequency as the time taken to reach the Schmitt trigger's upper and lower threshold is varied.

To obtain a sinewave, the triangular wave output is rounded, or 'shaped', with a special circuit. Sawtooth waveforms are generated by having different charge and discharge times for the integrator. A 'ramp-up' waveform is generated by charging the integrator's capacitor slowly and discharging it quickly. A 'ramp-down' waveform is obtained by charging the capacitor quickly and discharging it slowly.

There are a number of special ICs available that provide the fundamental circuit blocks to make a function generator. The XR2206 and the 8038 are probably the most well-known ones. Unfortunately, peither meets most of the performance requirements set down earlier in this article. At best, the XR2206 will only get to 150 kHz and the waveform 'trueness' begins to fall off once it gets past about 20 kHz. It's a fine IC for noncritical applications, such as a low-cost 'knockabout' function generator. The 8038 is better — it will comfortably get to 500 kHz, but, again, waveform trueness is nothing wonderful well below that limit.

To meet the requirements set down earlier, I had to tackle the project in discrete 'blocks'. The block diagram in Figure 2 is the result. That is the overall block diagram of the ETI-166 Function/Pulse Generator.

Constructionally, the ETI-166 has been divided into logical modules — the main generator, the frequency counter and display and the output amplifier. The latter has been dubbed Project 1520 and appears elsewhere in this issue. The other modules will appear in following issues.

Wideband amplifier module

Here is a wideband amplifier capable of driving a 50 ohm load with an output capability of 10 volts peak-to-peak. It has many applications, e.g: improving the sensitivity of a CRO or a counter, as a video amplifier or boosting the output of a signal generator.

David Tilbrook

THIS PROJECT is, in effect, the first part of the ETI-166 Function Generator, the remainder of which will be described over the next few issues. We decided to publish this section of the function generator separately so that a separate pc board could be designed and made available for any application needing a low distortion, wideband amplifier, such as a transmission line driver or video amplifier.

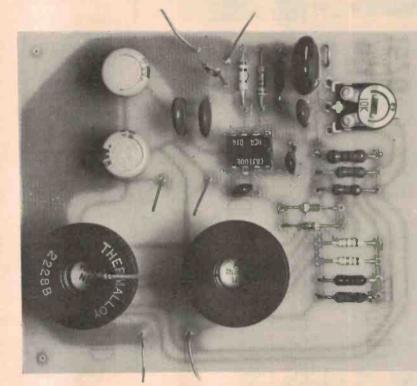
In the Function Generator this project is used as the output amplifier since its wide frequency response and 50 ohm output impedance enable it to provide large signal level sine and triangle waves (up to around 8 V peak on a 12 V supply) to a broad range of load impedances.

In order to keep both the design and the construction relatively simple it was decided to use an integrated circuit, if possible, to form the basis of the project. The main problem was to find a device capable of the necessary speed.

In order to produce a 10 V peak-to-peak triangle wave at a frequency of around 1 MHz, a slew rate of approximately 30 volts/microsecond would be a necessary minimum. Although this is not beyond the capability of a number of high speed operational amplifiers, some of these devices achieve a high slew rate by means of a relatively complicated 'feed forward' technique which can make the device prone to instability on some types of load. Furthermore, some of these devices have poor gain at high frequencies even though their slew rate figures are good.

After a look at the devices available, the CA3100 wideband op-amp was chosen. This device is manufactured by RCA and is almost ideally suited to this application. It combines both high slew rate and good gain at high frequencies. It is also convenient to use since it has a standard op-amp pinout.

The output drive capability is, of course, not sufficient but this is easily solved by following the op-amp with a unity gain emitter follower stage. This results in a general purpose wideband amplifier capable of driving a 50 ohm load and having a slew rate around 40 V/us. The bandwidth of the circuit is approximately 15 MHz for a 1 V RMS output signal level.



- SPECIFICATIONS ETI-1520 WIDEBAND AMP MODULE-

Gain

Maximum output voltage swing

Frequency response before slew rate limiting (power bandwidth)

Frequency response

Distortion (at 1 V RMS output) Noise (20 kHz bandwidth) Input terminated by 1k Input short circuited Current consumption +/-15 V supply rails) 9.2 x (approx. 20 dB) depends on R4

10 V peak-to-peak into 50 ohm load with +/--15 V supply (20 Vp-p no load)

 $\begin{array}{c} dc = -1.3 \ \text{MHz} \ (20 \ \text{Vp-p output}) \\ dc = -2.2 \ \text{MHz} \ (10 \ \text{Vp-p output}) \\ dc = -4.5 \ \text{MHz} \ (10 \ \text{Vp-p output}) \\ (below 5 \ \text{Vp-p output}) \\ (below 5 \ \text{Vp-p output}) \ (below 5 \ \text{Vp-p output}) \\ dc = -5 \ \text{MHz} \ +0, \ -3 \ \text{dB} \\ dc = -30 \ \text{MHz} \ +0, \ -10 \ \text{dB} \\ dc = -30 \ \text{MHz} \ +0, \ -20 \ \text{dB} \\ \hline (20 \ \text{Hz}) \ (20 \ \text{kHz}) \ (20 \ \text{kHz}) \\ (0.03\% \ 0$

10 nV/ Hz 8 nV/ Hz 25 mA (no load)

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Project 1520

Since the amplifier is dc coupled, a dc offset control has been included to enable the output dc level to be set to any voltage required.

In the ETI-166 Function Generator the output dc level is adjustable by means of a front panel potentiometer and this is accommodated using the optional dc offset input provided on the amplifier.

If this is not required in a specific application the on-board dc offset preset can be selected by fitting the appropriate link on the pc board.

Construction

Construction is fairly simple using the ETI-1520 pc board layout. Since relatively high frequencies are encountered, the layout and construction will affect performance. For this reason, an IC socket should not be used. The pc board layout should not be altered unless you have access to the test equipment required to ensure that performance has not been degraded.

Commence construction in the usual way by soldering the resistors and capacitors in place first. Be careful to orient C4 and C5 correctly since these are electrolytic capacitors. Keep all lead lengths as short as possible.

Solder the diodes, transistors and the IC in place, again ensuring that these devices are inserted with the correct orientation. Pin 1 of the IC is shown on the diagram accompanying the main circuit diagram.

A diagram showing the pinout of the transistors has also been included. The emitter of these devices is the pin closest to the metal tag on the case of the device.

If the internal dc offset preset is to be used solder this in place after bending the leads so that the preset will lie flush against the pc board. Fit a link as shown on the component overaly to select the preset.

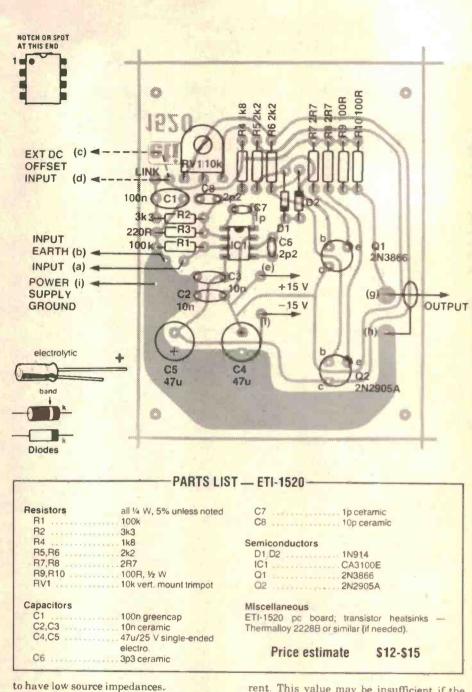
Once all the components are soldered correctly make a final check of the orientation of all polarised components. If all is well connect a ±15 V supply and power up.

Usually the dc offset will be required to be zero to ensure maximum output signal level and symmetrical clipping. Connect a voltmeter between the output and earth and adjust the offset pot so that the dc voltage on the output is zero.

The amplifier will work correctly on lower supply voltages than the ±15 V recommended, except that slew rate and frequency response are decreased. The RCA data sheet includes a curve showing the relationship between open-loop bandwidth and supply voltage (see opposite page). The supply voltage should not be allowed to increase above ±15 V since this is the maximum voltage recommended for the op-amp.

If the amplifier is to be operated from the same supply as other circuitry, ensure that the supply is reasonably free of noise. Although the amplifier has supply decoupling capacitors on the pc board most devices responsible for noise on supply lines tend

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The noise generators are therefore capable of charging and discharging the filter capacitors at the frequency of the noise. Increasing the value of the capacitors is usually ineffective since a very large increase is usually required.

The only effective solution is to increase the source impedance of the noise generators, and this is best done by fitting small value resistors in series with the supply lines. These resistors have not been included on the pc board since the value required (if they are required at all) is determined entirely by the particular application.

A good choice for most applications would be around 3R3, producing only a 1 V drop at the maximum possible output cur-

rent. This value may be insufficient if the supplies are extremely poor. In most cases these resistors will be entirely unnecessary and should only be used if a definite problem exists

In some applications it may be necessary to fit the output transistors with small heatsinks. If, for example, the amplifier is used to drive a 50 ohm load at low frequencies the instantaneous power dissipation in the output transistors will be around 1 W and a small heatsink should be used.

The best way to determine if a heatsink is required is to power up the unit and check the temperature of the output transistors. The transistor cases should be kept at a temperature that is comfortable to touch. and this will correspond to around 40 or 50 degrees Celsius.

wideband amp.

HOW IT WORKS ETI-1520

The project employs a wideband, high-speed op-amp (IC1) operated from split supply ralls, followed by a unit gain emitter follower stage (Q1, Q2) that provides some output current gain so that the amplifier can drive low impedance loads.

The op-amp chosen is a CA3100. Input signals are applied to its non-inverting input via terminal (a). R1 determines the overall input impedance. The output of the op-amp, pin 6, drives the emitter follower output stage. As split supplies are used, a complementary pair of transistors is employed, Q1 and Q2. Diodes D1 and D2 maintain a potential of about 1.2 V between the bases of Q1 and Q2.

As pin 6 of IC1 swings positive G1 turns further on and Q2 turns further off and the output voltage goes toward positive. As pin 6 of IC1 swings negative, Q2 turns further on, Q1 turns further off, and the output voltage goes toward negative.

Resistors R5 and R6 provide a small amount of bias for the emitter follower stage, while resistors R7 and R8 ensure current sharing between Q1 and Q2. Resistors R9 and R10 establish the output impedance at 50 ohms.

Overall feedback is provided by R4 and R3 and the gain is equal to:

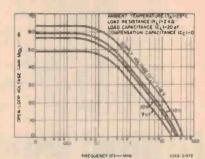
(R4 + R3)/R3

If you need to increase the gain, increase R4. Capacitor C8 provides frequency compensation in the main feedback loop. Capacitor C7 provides phase compensation for the op-amp for overall stability, while C6 provides a little negative feedback at high frequencies to ensure high frequency stability.

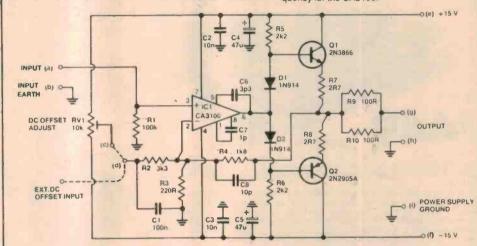
To provide dc offset adjustment, a small bias voltage is applied to the op-amp's inverting input via R2. This can be obtained onboard or off-board.

On-board dc offset is provided by linking terminals (c) and (d). This connects the wiper of RV1 to R2. Thus, adjusting RV1 will adjust the output dc offset voltage. Capacitor C1 is a bypass for this input.

Supply rail bypassing is provided by a 10n ceramic capacitor on each supply line, near the op-amp, and a 47u electrolytic capacitor on each supply rail.

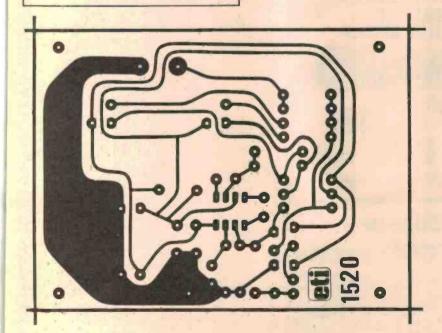


Performance. Open-loop gain versus frequency for the CA3100.



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Q1 Q2



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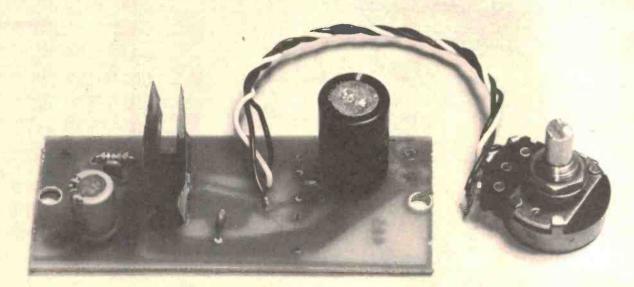
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A general purpose IC audio amplifier module

This is an ideal project for the beginner. In fact, most constructors will find it handy to have at least one around. A general purpose small audio amplifier finds many useful applications. Just two described here are an intercom and a 'baby minder'.

A VERY USEFUL object for any electronics enthusiast to have around is a simple audio amplifier. It can be used to test the operation of many circuits or employed in some practical item of equipment — such as an intercom.

This simple, yet versatile, module is easy to construct and can be powered from a variety of supply voltages, depending on your application. It will drive loudspeakers of 4, 8 or 16 ohms impedance and can deliver a maximum output of five watts.

The project has been designed around an integrated circuit audio power amplifier, the LM380 (from National Semiconductor) or the uA380 (from Fairchild). This is quite a versatile little IC and, using it, an audio amplifier is very simple indeed to make.

The '380 is generally available in a 14-pin dual-in-line package, and this is what I have employed here. An 8-pin version is available, but cannot be used in the pc board I have designed for this project. Pins 3, 4 and 5 plus 10, 11 and 12 of the 14-pin package are all connected together by a copper bar inside the '380 package, on which the chip is mounted. These pins can be soldered to a large area of copper on the pc board to act as a heatsink in relatively low power applications. Where the full power output capability of the '380 may be used, copper shim or tinplate heatsink 'flags' are soldered to these pins to get rid of more heat and keep the temperature of the IC down.

The '380 has a gain of 50 times. That is, it will amplify the input signal level by 50, which is a gain of 34 decibels (34 dB). That is:

Gain	ın	dB =	20	logio	(50)
		-	20	x 1.7	
		-	34	dB.	

The gain of the '380 is fixed by the manufacturer. But what if you want a volume control, as is so often necessary on an audio amplifier? That can be simply arranged by connecting a potentiometer as a voltage divider to the input of the IC. You can see how that's done from the circuit and construction diagrams.

You can use this project to amplify the output of a crystal set or one-transistor receiver to loudspeaker level simply by connecting the output directly to the input of the module.

You can make a 'baby minder' — for keeping an ear on the baby in its cot, from another

Geoff Nicholls

room — as shown later in this article, or you can make a simple intercom — which is also illustrated later. Another article in this issue shows how to use the module in a loudhailer.

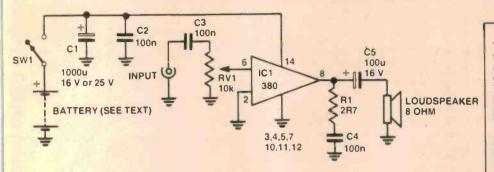
Right, let's get down to the business of building it.

Construction

As you can see from the overlay and wiring diagrams, there's very little to it. Start with the pc board. Whether you've made your own or purchased one, just give it a quick checkover to make sure all the holes are drilled correctly, that there are no small copper 'bridges' between closely spaced tracks (particularly between the IC pins) and no tiny cracks in any tracks. It's unlikely you'll have trouble, but it's always a good idea to check, before you run into trouble!

Note that mounting holes are located at either end of the board. These should be drilled to suit a 4 BA bolt, or whatever size you are using.

You can commence assembling the board by soldering resistor R1 in place, followed by capacitors C2, C3 and C4. All components



mount on the non-copper side of the board. Next identify the positive and negative leads of the two electrolytic capacitors. These are 'polarised' devices and can only go in one way. Solder them in place, putting the positive lead of each in the hole marked with a '+' on the overlay diagram.

The '380 IC can be soldered in place next. Make sure you place it in the board the right way round before soldering the pins. Do not use an IC socket as the board is designed to act parially as a heatsink and pins 3, 4, 5, 10, 11 and 12 must be soldered to the copper area for this purpose.

The heatsink flags can be constructed next if you need to use them. Use heavy gauge copper 'shim' or tinplate sheet (obtainable at hardware and motor spares stores). Two are required and the dimensions and cutting details are shown in the accompanying diagram.

General details for wiring up the speaker,

volume control and an on/off switch are also provided with the overlay diagram. Note that the input lead should be run in shielded cable, in general, especially if this lead needs to be more than 300 mm or so long. This prevents hum pickup from house mains wiring. For short runs, a pair of tightly twisted hookup wires will suffice.

This module will drive any size loudspeaker, from the tiny 50 mm 'transistor radio' types to 400 mm diameter 'monsters'. In fact, the larger a loudspeaker, the more sensitive it's likely to be and the louder it will sound! You don't need more power to drive a larger loudspeaker, despite what you might at first think.

The bigger loudspeakers generally have a more powerful magnet than the smaller types. This makes them more sensitive to the currents flowing in the voice coil. This and the larger cone combine to produce a louder sound.

IC amp module

-HOW IT WORKS --- ETI-464-

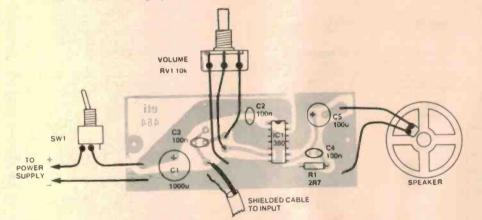
There's not much you can say about this! The whole job is done by the '380 IC audio power amplifier. The input is coupled to the '380 via a capacitor (C3) and the volume control, RV1. The latter is just a voltage divider, applying less or more voltage to the IC's input as the potentiometer is varied, thus varying the volume.

The output of IC1, pin 8, Is biased at half the supply rail (e.g. it will be at 4.5 V if the supply is 9 V). For this reason, the output is capacitively coupled to the loudspeaker via a large value electrolytic capacitor, C5. This presents a low impedance in series with the loudspeaker, which is a relatively low impedance device.

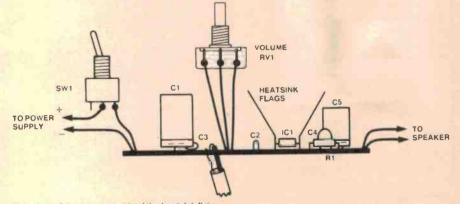
Any tendency to instability of the amplifier is suppressed by the network of R1-C4 connected from the output to common.

The supply rail is bypassed by an electrolytic capacitor, C1, at the low frequencies, and a greencap or ceramic capacitor, C2, at the higher audio frequencies.

Note that provision has been made on the pc board for powering an electret type microphone, simply by adding a resistor adjacent to C3

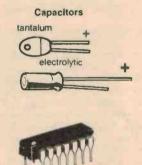


Overlay and wiring diagram. Showing component placement and general wiring details.



Side view. Showing assembly of the heatsink flags.

COMPONENT PINOUTS



DADTO	IST - ETI-464
PARISL	131 - E11-404
Resistors R1 RV1	2R7
	1000u/16 VW or 25 VW RB electro.
C3	100n ceramic bypass 100n greencap 100u/16 VW RB electro.
Semiconductors IC1	LM380
Miscellaneous ETI-464 pc board; S	SPST switch; shielded cable;

Price estimate \$7-\$8

Project 464

Testing it

The easiest way to test it is simply to connect a 9 V battery to the module and turn up the volume control. Then, touch your finger to the 'top' end of the volume control - the right hand lug when looking at the rear of the pot. You should hear hum and noise, or perhaps a loud 'blurting' sound. If not, check that the battery is connected the right way round and that the speaker and volume control wires are all intact and correct. Check that you have the IC correctly orientated.

Connecting a 9 V battery in reverse to the module is unlikely to destroy the IC, but any higher supply voltage connected in reverse sure will, so watch this point.

If the amplifier tends to be unstable, 'squealing' or otherwise 'acting up', try connecting a 4u7/16 V tantalum capacitor between pin 1 of the IC and the adjacent grounded area of the pc board, directly on the underside of the board. The positive lead goes to pin 1. Keep the lead lengths short. This should cure it.

Always keep the amplifier's input leads away from the speaker leads, to avoid feedback which may result in 'howl round'an uncomfortable whistling or howling sound that is affected by moving the leads.

Power supplies

This module can be powered form batteries, a suitable plugpack or transformer and rectifier to suit yourself. The power output depends on the supply voltage and the speaker impedance. As stated earlier, the '380 can drive 4, 8 or 16 ohm speakers. By far the better speaker to use is an 8 ohm impedance type. Fortunately, they're also the most common type

Powered from a nine volt battery, you will get about half a watt (500 mW) output, which is more than adequate for 'personal' listening stations; e.g: providing loudspeaker output from a crystal set or onetransistor radio, etc. The power dissipated by the IC under these circumstances is about three quarters of a watt maximum, so no heatsink flags would be necessary. The module draws only about 5 mA with no signal (called the 'quiescent current').

The absolute maximum supply voltage the IC will tolerate is 22 V. With an 8 ohm speaker, the project will deliver five watts output, which is remarkably loud! Under these circumstances, the power dissipated by the '380 will be a little over three watts maximum and heatsink flags will definitely be necessary. The quiescent current is about 8 mA on a 22 V supply

A plugpack or transformer and rectifier supply suitable for powering this module should provide 12 Vdc at 200 mA or so. Using such a supply, the project will deliver about one to 11/2 watts to an eight ohm speaker, which is quite suitable for an intercom, for example. The heatsink flags are not entirely necessary with this sort of application, especially in an intercom where the amplifier is only used intermittently

(NATTS)

DEVICE DISSIPATION

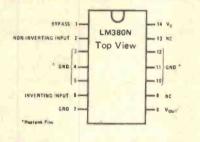
Two nine volt batteries connected in series will supply 18 Vdc or so to the module, which will deliver about three watts to an 8 ohm speaker. Heatsink flags are necessary in this case. Note that, when using a 4 ohm speaker, the supply should not exceed 15 volts.

LM380 audio power amplifier general description

The LM380 is a power audio amplifier for consumer application. In order to hold system cost to a minimum, gain is internally fixed at 34 dB. A unique input stage allows inputs to be ground referenced. The output is automatically self entering to one half the supply voltage.

The output is short circuit proof with internal thermal limiting. The package outline is standard dual-in-line. A copper lead frame is used with the center three pins on either side comprising a heat sink. This makes the device easy to use in standard p-c lavout.

Uses include simple phonograph amplifiers, intercoms, line drivers, teaching machine outputs, alarms, ultrasonic drivers, TV sound systems, AM-FM radio, small servo drivers, power converters, etc.



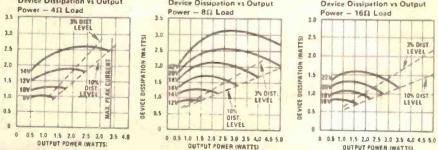
absolute maximum ratings

0	
Supply Voltage	22V
Peak Current	1.3A
Package Dissipation 14-Pin DIP (Notes 6 and 7)	10W
Input Voltage	±0.5V
Storage Temperature -65°	C to +150°C
	°C to +70°C
Junction Temperature	+150°C
Lead Temperature (Soldering, 10 sec)	+300°C
Junction Temperature	+150°C

electrical characteristics

Note 1: Ve = 18V and Te = 25°C unless

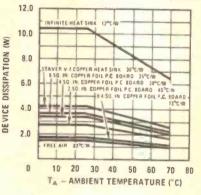
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Power	POUTIAMS	RL = 8Ω, THD = 3%	2.5			w
Gain	Av		40	50	60	V/V
Output Voltage Swing	Vout	$R_{\rm L} = 8\Omega$		14		Voo
Input Resistance	ZIN		-	150k		£
Total Harmonic Distortion	THD			0.2		%
Power Supply Rejection Ratio	PSRR			38		dB
Supply Voltage	Vs		8	00	22	V
Bandwidth	BW	POUT = 2W, RL = 812	-	100k	66	Hz
Quiescent Supply Current	la			7	25	mA
Quiescent Output Voltage	Vouto		8	9.0	10	V
Bias Current	BIAS	Inputs Floating		100	10	A
Short Circuit Current	Isc			1.3		A



features

- Wide supply voltage range
- Voltage gain fixed at 50
- High peak current capability
- Input referenced to GND
- High input impedance
- Low distortion
- Quiescent output voltage is at one-half of the supply voltage
- Standard dual-in-line package

Device Dissipation vs Ambient Temperature



Total Harmonic Distortion

0.5 10 20

Po - OUTPUT POWER (WATTS)

5.0 10

vs Output Power

COVPASS = SHF HEATSINK = TWO

COPPER WINGS

f = 1 kHz V_{CC} = 22V R_L = 851

10 2

9.0 **DISTORTION**

8.0

7.0

6.0

50

4.0

30 TOTAL

2.0

1.0 OH

0 0.1 0.2

HARMONIC

A 'baby minder'

You can 'keep an ear' on a baby asleep (or supposed to be!) in its cot in another room by organising some sort of microphone to pick up sounds from the baby's room to be amplified and heard in another room.

The general wiring diagram for a baby minder is shown in Figure 1. Here, a small 8 ohm loudspeaker is employed as a microphone — and they're remarkably effective. A transformer is needed to 'step up' the tiny voltages produced by the speakermicrophone. A suitable type is generally described as a 'transistor output transformer, 1k centre-tapped to 8 ohm'. Dick Smith Electronics lists a suitable type — cat. no. M-0216. Altronics have a similar one. Tandy lists one also, no. 273-1380.

The '8 ohm' side is connected to the speaker-microphone — this is the side with just two leads. The 1k side of the transformer is connected to the input of the module. Mount the transformer close to the module. The module could be mounted in a suitable cabinet with the speaker, volume controls and on/off switch mounted on the front.

The speaker-microphone could be mounted in a small jiffy box placed in a convenient position in the baby's room, near the cot. This connects to the amplifier via a length of 'twisted pair' cable or light 'figure-8' flex. Try and avoid running this lead adjacent to house mains wiring to avoid possible hum pickup.

I have specified an 8 ohm speaker as a microphone as it is of such a low impedance that the possibility of hum pickup on the cable between the microphone and the amplifier is greatly reduced.

You can either use a battery supply or a 12 Vdc plugpack.

Intercom

The general details for wiring a simple intercom are shown in Figure 2. You'll need two single-pole, double-throw (SPDT) toggle switches with a spring return. Double-pole types are also suitable, just use one side (e.g: C&K type 7208 or similar). You'll also need an M-0216 transformer, or similar, as for the Figure 1 circuit.

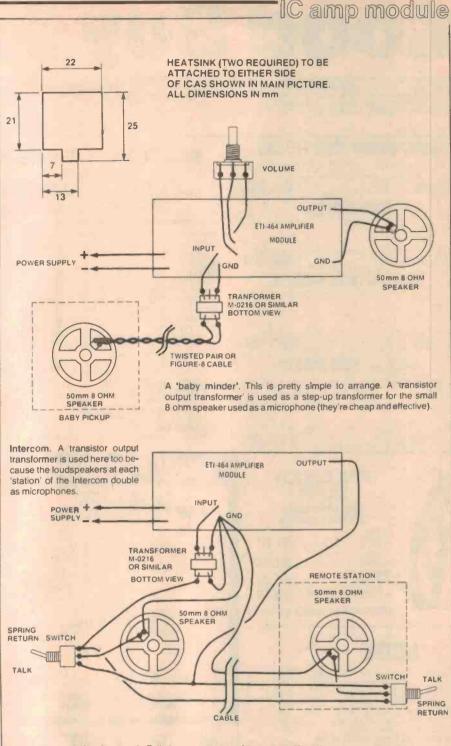
Two small 8 ohm speakers are used as both speaker and microphone at each end of the intercom.

A volume control is necessary on an intercom and a small 'trimpot' can be mounted on the pc board where the volume control connections are made. The hole spacings are suitably placed for soldering a common vertical mounting trimpot in place. Use on of the same value — i.e: 10k. Test out the intercom and set the volume control to suit yourself.

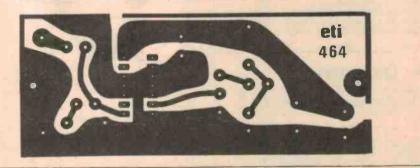
A suitable cable can be made by twisting together three strands of light hookup wire or buying a suitable length of light multicore cable. Note how the various common, or earth, connections are made to the one ground point on the pc board.

Conclusion

Well, I've described how to build yourself a general purpose audio amp module and how to use it in a couple of applications — the rest is up to you. Have fun!



Printed circuit artwork. Full-size reproduction of the pc board layout.





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PROJECTING your voice outdoors is quite a difficult task without some means of 'directing' your voice and amplifying it. Generally, you'll want to address a group of people located some distance away, or a group of people spread out in front of you for some distance. If you can direct your voice over a narrow 'beam', then less of the sound you make is wasted.

The old-fashioned megaphone did that job before 'electronics' entered the picture. Outdoor public address systems came into being with the advent of valves. For many years PA systems were cumbersome, hardly portable beasts until minaturisation came along post World War II. The first 'loudhailer' PA systems portable by one person used miniature valves, a small horn speaker and a set of cumbersome, heavy batteries that didn't last all that long.

When power transistors came along, loudhailers proliferated. They could be held in one hand, used a small number of 'torch' batteries and did the job better than before.

The horn speaker

The horn loudspeaker is by far the best type for outdoor use. Horns can be made weatherproof and have an efficiency of better than 20% compared to a few per cent for ordinary speakers. This allows an amplifier of lower power to be used, with consequent savings in power consumption, physical size and weight.

Horns are intrinsically limited in their frequency response, and their efficiency is inversely proportional to their bandwidth. PA horns are designed to operate over the voice band at maximum efficiency. The horn itself is essentially an impedance transforming device which increases the acoustic loading on the driving diaphragm to allow better 'matching' to the air. The throat area of the horn increases exponentially as you move away from the driver.

The horn may be straight, as shown in Figure 1, or folded, as shown in Figure 2. The folded horn is physically smaller and is the most common type in low cost PA systems. Folding the horn reduces the efficiency slightly but increases the coverage or dispersion, which is usually an advantage.

The straight horn has a long 'throw' and is useful for narrow sound coverage at greater distances, but is more cumbersome, especially for handheld applications!

The project

For our loudhailer, we had to search around for a suitable small folded horn. There is a variety available and prices vary widely. Probably the most common are 130 mm diameter (5") low power folded (or 'reflex') horns generally sold for boat or CB PA use. Rectangular folded horns are also available, having an opening of 200 mm wide by 120 mm or so high.

Efficiencies vary widely and are best judged by the weight! Drivers with larger, heavier magnets are more efficient than those with smaller, lighter magnets. Most have a 'dispersion angle' — the angle over which the majority of sound is dispersed from the horn — of between 60° and 90° . The narrower the dispersion angle, the greater sound level you get at a given distance from the speaker.

Geoff Nicholls



Your shout! Not beautiful, but effective.

The horn we chose for our prototype is imported and marketed by Benelec Pty Ltd, model no. 8-224. It is a 130 mm diameter folded horn, measuring 170 mm long overall. There is a cover on the rear of the horn with plenty of room inside to mount the power amp. module. It has a mounting

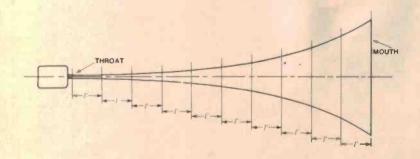
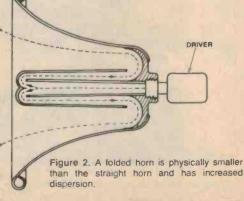


Figure 1. A straight horn has the width of the throat growing exponentially larger with increasing distance from the driver.



Project 465

bracket that allows the horn to be swivelled over a wide range of angles. It is available with driver impedances of 4 ohms or 8 ohms, though the latter is best in this application. The dispersion angle is quoted as 60° , which we saw as desirable, and the output is quoted as being 122 dB (presumably with 1 W drive at one metre). It weighs 1.15 kg, which is not too heavy, yet ensures the sort of driver efficiency desirable for maximum effectiveness.

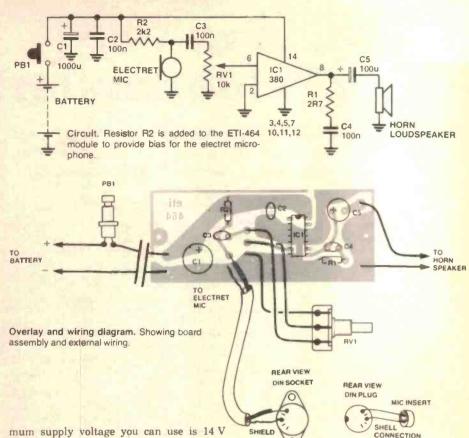
Using this horn, and constructing the loudhailer as described here, you can make yourself heard at 100 metres. If you don't need that much throw, then a lower cost, less efficient horn will suffice, but construction details will have to be worked out to suit yourself. See Shoparound in this issue for horn suppliers.

An electret microphone insert was employed to make a microphone. It proved cheap and effective. To provide a handle and battery case, a short length of 50 mm (i.d.) PVC water pipe was used, along with two end caps. This mounts, via one end cap, on the horn's mounting bracket and the batteries are slipped inside. A momentary action pushbutton switch, mounted on the 'handle', serves as an on/ off switch.

Originally, I tried mounting the mic insert in the centre of the horn's rear cover, but feedback proved a problem and I couldn't utilise the full gain and output of the amplifier module. A little experimentation solved the feedback problem and improved an operator's visibility at the same time.

I mounted the mic insert in a DIN plug which plugs into a socket mounted at the top of the horn's rear cover. I also mounted a gain control pot on the cover. These measures overcame feedback problems and allowed you to see over the top of the horn.

The maximum output a '380 will deliver is five watts into an 8 ohm load using a 20 V supply. With an 18 V supply, the '380 will deliver a maximum of four watts (@ 10% distortion — which is tolerable) into an 8 ohm load. With a 4 ohm load the maxi-



mum supply voltage you can use is 14 V without exceeding the '380's dissipation rating, and you only get three watts' output. In a loudhailer, every watt counts.

Hence, I opted to use an 18 V supply. There are two ways you can arrange this with batteries. Two no. 2362 9 V batteries can be 'snapped' in series. These are 75 mm long with a male snap clip at the positive end and a female snap clip at the negative end. Alternatively, you can use 12 AA cells mounted in three four-cell battery holders. There are two advantages to the latter: the batteries last longer and the whole assembly is considerably cheaper.



Insides out. The amplifier board mounts inside the rear cover of the Benelec 8-224 horn.

PARTS LIST - ETI-465

This requires construction of the ETI-464 amp. module with the addition of the following components:

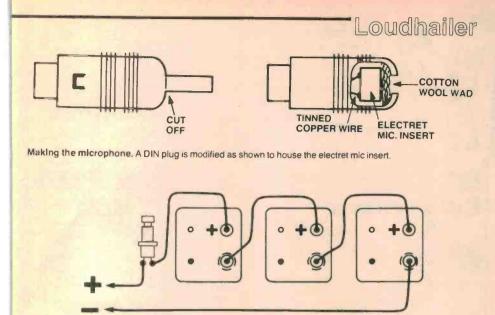
Price estimate \$45-\$50

Construction

Putting the loudhailer together is quite straightforward. The ETI-464 pc board has mounting holes which match the mounting posts on the inside of the 8-224 horn speaker's rear cover. Self-tapping screws are used to mount the board and the mounting holes should be drilled to size before assembling the components to the pc board.

Assemble the power amp module according to the instructions given in the previous article on the ETI-464. Note that R2 has to be added, as shown on the overlay diagram. Don't attach any wires yet until the mechanical assembly has been completed.

Drill the mounting holes for the volume pot and the DIN socket in the speaker's rear



Battery holder wiring. If you use a dozen AA cells, as I did, this is how the battery holders are wired up to provide 18 volts.

cover. The DIN socket goes at the top, the volume pot at the bottom. Also drill a hole in the bottom lip of the cover so that the leads from the battery may be passed through.

Now tackle the handle/battery compartment. Cut a 200 mm length of 50 mm i.d. PVC pipe. File the ends smooth and square and slip the end caps on. Holding them in place with masking tape, drill holes on either side, right through the cap and pipe, so that self-tapping ('PK') screws can be used to secure the end caps in place. Drill these holes to the root diameter of the PK screws.

Remove the end caps and enlarge the holes to the appropriate clearance diameter for the PK screws. Drill holes in one end cap to suit the speaker mounting bracket and to pass the battery leads. Bolt it in place.

Now take the tube and mark a hole position at the 'top' end for the on/off pushbutton switch. It should be located such that it clears the upper end cap, yet is not too far down the tube so that access to the switch connections is restricted.

Now wire up all the battery connectors, the pushbutton switch, the DIN socket, the volume pot, the speaker and the pc board. Check it all carefully when finished, then screw it all together. A little wad of sponge rubber in the upper end of the handle secures the batteries.

Now you can make the microphone. The basic assembly is shown in the accompanying diagram. We found that angling the mic insert *down* (when the unit is plugged in) helped reduce feedback problems and a tendency to 'breathiness'.

The accompanying photographs show the internal and overall assembly, when completed.

Using it

For an initial try-out, set the volume pot about halfway advanced, plug in the microphone, position your mouth about 10 mm or so from the mic, press the button



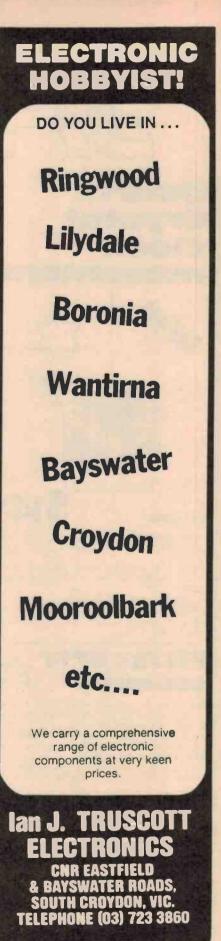
Rear view. The mic plugs into the DIN socket at the top. The volume control is below it. The 'handle' houses the batteries.

and say a few words. No 'howl round' feedback or 'ringing' should be experienced, except perhaps if you're in a small room. Best try the unit outdoors.

Adjust the setting of the volume control for maximum output without feedback or ringing being evident. Always speak very close to the microphone.

If you wish, it may be convenient in some applications to have a 'remote' microphone. An electret insert can be readily installed in a CB-type handheld mic case, with the push-to-talk switch wired in parallel with the loudhailer's on/off pushbutton via the DIN plug and socket.

Happy hailing!





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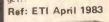
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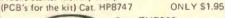
Most imported equipment these days now uses IEC-320 style AC power inlet connectors. Indeed, the electronics mags will soon be specifying these connectors on many of their mains-powered projects to simplify (and therefore make safer) mains wiring. Jaycar now stocks a range of ELECTRICITY AUTH-ORITY APPROVED mains line cords.

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	ENTRY - 2M	\$3.95
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	ENTRY -2M	\$3.95
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This book measuring 230 x 175mm gives full data on popular RAM and EPROMs (2102, 2114, 4116, 4164, 2708, 2732, 2764 etc) as well as many other lesser known types including shift registers etc. Due to industry standardisation, the data in this book will relate to other manufacturers of MOS memory

DATA BY FAIRCHILD





SEE THE OTHER JAYCAR ADS IN THIS MAGAZINE FOR ADDRESS DETAILS

The 'power down' mains appliance timer

Have you ever turned on your electric radiator to warm a room in the morning and forgotten to turn it off on your way out? What about the patio light — ever turned that on and forgotten to turn it off until the next day? This project will not only solve those problems, but has many applications besides.

THIS IS a press-button mains appliance timer. You plug the project into a wall socket, plug your appliance into the project, then press the button to turn the appliance on. Some predetermined time later, the project will automatically 'power down' (or turn off) your appliance.

No doubt you're familiar with those time switches installed in the stair wells of apartment buildings that control the stairway lights. You press the big button and the lights come on just long enough for you to not quite make it up the stairs! These devices employ a mechanical timer, their major problem being that, with use, the period decreases. In addition, you can't preset them to a convenient time over a wide range.

Many 'deluxe' electric heaters these days incorporate a timer of some sort so that the heater can be turned on for a preset period, then turn itself off. 'Super deluxe' models go so far as having a timer with digital display and programming facilities to turn it on and off whenever you like each day for a fortnight! That's probably going a bit far for many applications, though.

Getting 'closer to home', so to speak, take the case of an EPROM eraser. This basically comprises an ultraviolet lamp and a period timer. You bung your EPROMs under the lamp and turn on the timer. The timer then turns the lamp off the required period later (typically, 20 minutes).

In photographic work, film and print expo-

sure times are often handled with a simple period timer. When making multiple exposures for comparisons, etc, binary-related exposure times are frequently used, e.g: print exposures of eight minutes, then four minutes, then two minutes, etc.

This versatile little timer will handle all the above applications and can be configured to suit a huge variety of other applications, probably only limited by your imagination.

This project can be set up for timing periods ranging from one minute up to at least 50 hours. Although I haven't tried it, it is certainly possible to extend the period to around 400 hours (16-17 days). Binary 'segments' of the basic period can be selected by the simple addition of a suitable switch. The project is low in cost and can be built into an appliance if you so wish.

Design

The unit is quite straightforward. Figure 1 shows a block diagram. It consists of an oscillator, a divider, a flip-flop, a relay driver and relay. The relay contacts are used to connect and disconnect the mains power from an attached appliance.

The oscillator output is divided down 8192 times. The divider's output drives the 'reset' input of the 'reset-set' (RS) flip-flop. The flipflop has two outputs — A and B. In the initial, or reset, state, output A is high, reset-

Geoff Nicholls

ting the divider in preparation for counting. Output B of the flip-flop is low and the relay is not operated.

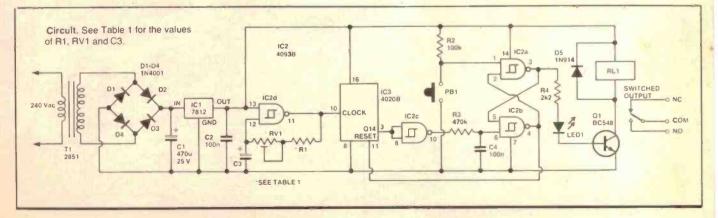
When the start button is pressed, the RS flip-flop is 'set'. Output B goes high, operating the relay and turning on the appliance connected via the relay. Output A goes low, enabling the divider to count down.

At the end of its count, the divider's output goes low, resetting the flip-flop. Output A goes high again, stopping the divider, and output B goes low. The relay then drops out, turning off the appliance connected via its contacts.

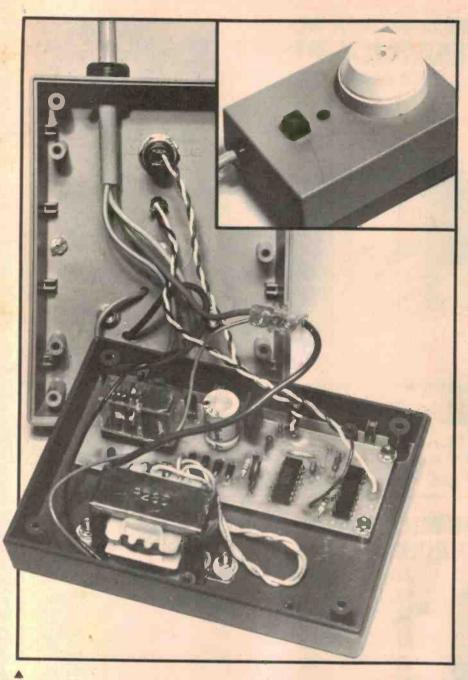
The oscillator's frequency will determine the basic period. Because the division ratio is so high, minor drifts or variations in the oscillator frequency will not materially affect the timing period, hence a simple oscillator can be used.

As the required timing period is determined by R1, RV1 and C3, the values you need for these components should be determined from Table 1 before commencing construction. It would be reasonable to expect a kit supplier *not* to include these in a kit, or to perhaps only include, say, a 1u RBLL or tantalum for C3 and perhaps a 1M trimpot, as these would cover a wide range of timing periods, leaving the choice of R1 to you.

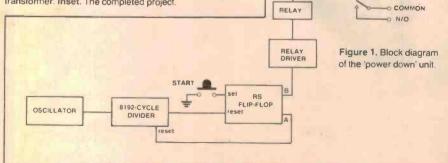
Note that C3 should either be a tantalum type or single-ended 'low leakage' electrolytic for repeatable results.



90 - July 1983 ETI



On the inside. Showing construction of the project and how I mounted it inside the Unibox case. Note that one of the internal pillars is cut down to provide clearance for the transformer. Inset. The completed project.



'power down' timer

HOW IT WORKS - ETI-265

The overall design of the project is discussed in the text, so this shall largely be a blow-byblow description of how it works. Let us start at the 'rear end' first.

Two gates from the 4093, IC2a and b, form the reset-set (RS) flip-flop. This is in the SET state when pin 3 of IC2 is high and pin 4 is low. It is in the RESET state when pin 4 is high and pin 3 low.

When the flip-flop is SET, Q1 is turned on as the high on pin 3 of IC2 biases on its base via R4 and LED1, which lights, indicating the unit is 'active'.

A momentary low on pin 1 of IC2 will SET the flip-flop, while a momentary low on pin 6 will RESET it.

On power-up, C4 will look like a short circuit and keep pin 6 of IC2 low long enough to RE-SET the flip-flop. Thus, pin 4 of IC2 will be high and reset IC3, a 4020 14-stage binary counter with a division ratio of 8192:1.

Pin 3 of IC3 (the stage 14 output) will be low after the IC is reset and IC2 will invert this, pin 10 driving pin 6 high, allowing the flip-flop to be set when PB1 is pushed momentarily.

As soon as PB1 is pushed, the counter begins to count cycles of the Schmitt gate oscillator comprising IC2d, R1, RV1 and C3. The frequency of the oscillator is determined by RV1, R1 and C3 and the threshold voltages of the inputs of IC2d. (See 'Why some CMOS circults don't work as you expect', Lab Notes, in this issue.)

After 8192 cycles of the oscillator, pin 3 of IC3 goes high, thus resetting the flip-flop via IC2c. Pin 3 of IC2 will thus go low and the relay will drop out.

The oscillator works as follows: At powerup, pin 12 of IC2d will be low as C3 appears as a short circuit. Pin 13 will be high and thus pin 11 will be driven high. C3 will charge via R1 and RV1. When the voltage on C3 reaches the gate's upper threshold, pin 11 will go low and C3 begin to discharge via RV1 and R1. When the voltage on C3 fails to the gate's lower threshold, pin 11 will again go high and the cycle will be repeated.

The oscillator frequency is related to the timing interval (or period) as follows:

t.

where f_0 is the oscillator frequency in Hertz, and T is the period in seconds. Expressing this in terms of the period:

$$r = \frac{8192}{f_0}$$

Table 1 allows you to set up the project for a particular timing interval. The range of adjustment provided should compensate for most of the variation in threshold voltages of the different manufacturers' 4093s. The time interval shown in the table refers to the prototype project. If you add the rotary switch to give four ranges, the longest interval will be as per the table, the next will be half that, then quarter that, etc.

N/C

The power supply is quite straightforward. Transformer T1 drops the mains voltage to 12.6 V RMS. This is rectified by a bridge diode comprising D1 to D4. Capacitor C1 is the rectifier reservoir. About 18 volts is developed across C1 and this is regulated to 12 V by a 7812 three-terminal regulator, IC1. Capacitor C2 ensures regulator stability.

Diode D5, across the relay, shorts the reverse-emf generated by the relay coll when Q1 switches off, preventing the high voltage generated from destroying Q1.

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Project 265

Construction

I built one prototype and mounted it in a plastic case, which you see in the photographs here, and another unit which I mounted inside an appliance (ask no questions and ...)

I used an all-plastic Unibox (P/N144) measuring 135 mm long by 100 mm wide by 50 mm deep. (Magraths in Melbourne are a major supplier of these.) It has ample room inside. You could use a UB-1 zippy box (50 x 90 x 150 mm), but an all-plastic case is recommended for safety's sake. Jaycar stock a range of smart, all-plastic (ABS) cases in various colours. The HB6150, 1, 2, 3 series (orange, grey, blue, black) would suit.

First thing to do is determine the mounting positions of the printed circuit board and transformer. Lay the unassembled pc board and the transformer in the base of whatever box you're using and mark the positions of the mounting holes.

Then mark out suitable positions for the mains cable entry, the pushbutton, the LED and the mains socket. For the mains cable entry, I used a Heyco clamp-type grommet to suit the mains cable I used (grommet #1210, SR-6P-4). There are plenty of similar types available. If you don't use a clamp-type grommet, then use an ordinary rubber grommet and a nylon cable clamp.

The mains socket requires holes to be drilled in the box to allow cables to the active, neutral and earth pins to pass through to the socket terminals. I first loosened the grub screws on the socket terminals, then positioned the socket where I wanted it and drilled pilot holes in the case through the terminals.

After drilling all the holes, mount the transformer, output socket, LED and pushbutton, but not the pc board — you've got to assemble that yet! To mount the transformer, I used insulating washers from a TO-3 power transistor mounting kit. An alternative is to use nylon bolts. The latter were used to mount the pc board.

Note that, if you wish, the section of the board on which the relay mounts can be severed from the rest of the board, allowing the relay to be mounted away from the main portion of the electronics.

Assembly of the board is pretty straightforward. Start with the smaller components. Solder all the resistors in place, then the five diodes, C2, C3 and C4. Watch the orientation of C3 and the diodes. Solder a link of tinned copper wire in the position shown, between C4 and R4. Now you can solder IC2 and IC3 in place.

These are CMOS ICs. Use an iron with an earthed tip and only handle the ICs with your thumb and forefinger gripping the ends of the package. Avoid touching the pins. When you have each in place, solder pins 7 and 14 of IC2 first and pins 8 and 16 of IC3 first, before going on to solder the other pins. If you wish, IC sockets may be used without affecting operation of the project.

Next solder Q1 in place. Its orientation can be ascertained from the pinout diagram and the component overlay. Now solder the electrolytic, C1, in place, taking care to orientate it correctly, followed by IC1 (get it the right way round) and RV1. The relay can be soldered to the board last of all.

Wires are run from the pc board to the mains circuitry, the pushbutton and the LED. Only ordinary light hookup wire $(10 \times 0.12 \text{ mm})$ need be used to connect up the LED and the pushbutton. The 2851 transformer primary wires are generally coloured red and black. It has three secondary wires, two the same colour. These are wired to the rectifier diodes D1-D4, as shown in the overlay/wiring diagram.

Wire the transformer primary very carefully. The brown active wire from the mains input cable goes to a terminal connector, where it joins the red wire from the transformer. Take a length of brown mains wire and connect it between this terminal connector and the COMMON relay terminal pad on the pc board. Another length of brown mains wire is run from the normally open (N/O) relay contact pad on the pc board to the active terminal of the mains output socket.

The blue neutral wire runs from the mains input cable direct to the neutral terminal on the mains output socket, along with the black wire from the transformer.

The green and yellow striped earth wire from the mains input cable goes direct to the earth terminal on the mains output socket. This wire should be longer than the other two from the mains input cable for safety reasons. Should the mains cable be accidentally pulled out from the case, the earth wire will be the last to break.

All finished? Check everything thoroughly. That's all there is to it. Next thing to do is test and calibrate it.

Test and calibration

Set RV1 to the middle of its travel, then close the case so that you can't accidentally come in contact with the lethal mains voltage present.

Plug a bedside lamp into the output socket, plug the project into the mains and switch on. Nothing should happen. If all's well, press the pushbutton and the lamp should light. Now time how long it remains on (if it's 50 hours or more, you're going to need a *lot* of patience!).

If you have access to a frequency counter, then setting the timer is much easier. Just attach the frequency counter to pin 11 of IC2. The oscillator frequency is related to the timing period as follows:

Frequency = 8192

where T is the desired period in seconds and the frequency is in Hertz.

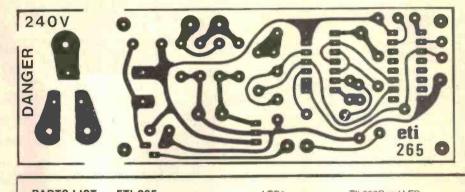
For a timing period of ten minutes, the frequency should be 13.65 Hz. For really long times, you're better off measuring the period of the oscillator output.

All you need do is to adjust RV1 for the correct frequency.

Binary periods

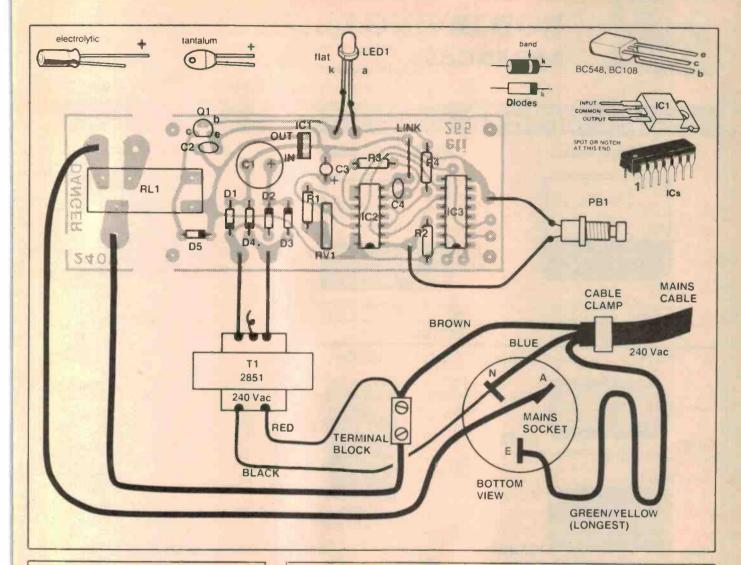
As mentioned earlier, the simple addition of a switch can give you binary (half, quarter, eighth) segments of the basic period. Details are shown in Figure 2.

Cut the track that runs beneath the body of IC3, then wire a single-pole, four-position switch as shown. Position 4 gives you the full period. Position 3 gives the half period, position 2 one-quarter and position 1 gives oneeighth the period. \bullet

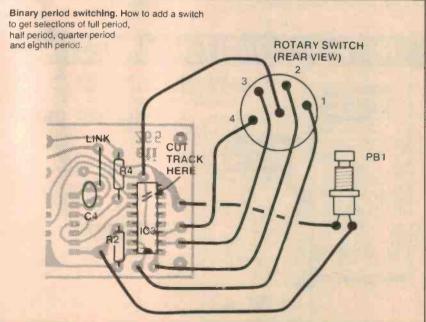


PARTS LIST — ETI-265	LED1TIL220R red LED
Resistors all ¼ W, 5% unless noted R1 47k R2 100k R3 470k R4 2k2	Q1 BC548 Miscellaneous PB1 momentary action pushbutton, large (e.g: D.S.E. no. S-1199 or similar)
Capacitors 470u/25 V single-ended electro. C2,C4 100n greencap C3 1u/16 V tantalum Semiconductors 5 D1-D4 1N4001, 1N4002 or similar D5 1N914, 1N4148 IC1 7812 IC2 4093B IC3 4020B	RL1 pc mount relay, 12 V coil SPDT/5 A contacts (e.g: D.S.E. S-7125, or similar) T1 2851 transformer, 12.6 V CT @ 150 mA ETI-265 pc board; case to suit (e.g: Unibox P/N 144, 100 x 135 x 50 mm, or similar); LED mount; 240 V wall socket; malns cable plug, cable and clamp; wire, etc. Price estimate \$32-\$37

'power down' timer



Timing Interval	C3	R1	RV1
1 minute	100n	47k	100k
5 minutes	100n	270k	500k
10 minutes	100n	470k	1M
20 minutes	10	100k	220k
30 minutes	10	150k	220k
45 minutes	10	270k	500k
1 hour	10	270k	500k
11/2 hours	1u	390k	1M
2 hours	1u	680k	1M
4 hours	10	1M2	2M2
10 hours	100u	27k	50k
20 hours	100u	56k	100k
50 hours	100 u	150k	200k
100 hours	100u	270k	500k
200 hours	100u	560k	1M
400 hours	100 u	1M2	2M
TABLE 1			





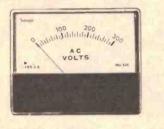
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Q10515	MU45	0-10A	9.75	7.95
Q10520	MU45	0-20V	9.75	7.95
Q10525	MU45	0-30V	9.75	7.95
Q10535	MU45	VU	10.95	9.95
Q10530	MU52E	0-1mA	10.95	8.95
Q10533	MU52E	0-5A	10.95	8.95
Q10538	MU65	0-50uA	13.95	11.95
Q10540	MU65	0-1mA	13.95	11.95
Q10550	MU65	0-100uA	13.95	11.95
Q10560	MU65	0-20V	13.95	11.95

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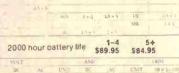
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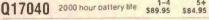
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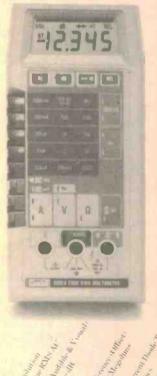
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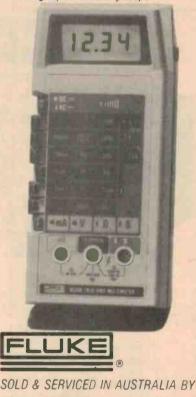
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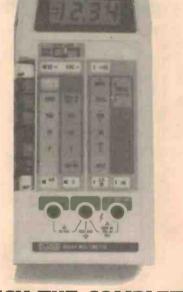


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OP-AMP COOKBOOK

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ETI's practical guide to one of the most versatile electronic building blocks in use today.

Using the 741 - the universal IC

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Contents

LM 3900N - four op-amps in one package

Variable gain op-amp.

The op-amp cookbook — using the 741

Here's how to use the 741 op-amp — one of the most versatile building blocks of all time.

OPERATIONAL AMPLIFIERS (OP-AMPS) CAN be simply described as high-gain direct-coupled voltage amplifier 'blocks' that have a single output terminal but have both inverting and non-inverting input terminals. Op-amps can readily be used as inverting, non-inverting, and differential amplifiers in both a.c. and d.c. applications, and can easily be made to act as oscillators, tone filters, and level switches, etc.

Op-amps are readily available in integrated circuit form, and as such act as one of the most versatile building blocks available in electronics today. One of the most popular op-amps presently available is the device that is universally known as the "741" op-amp In this article we shall describe the basic features of this device, and show a wide variety of practical circuits in which it can be used.

BASIC OP-AMP CHARACTERISTICS AND CIR-CUITS

In its simplest form, an op-amp consists of a differential amplifier followed by offset compensation and output stages, as shown in Fig. 1a. The differential amplifier has inverting and non-inverting input terminals, a high-impedance (constant current) tail to

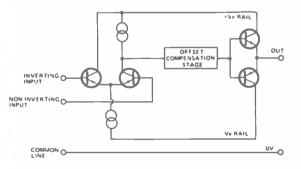
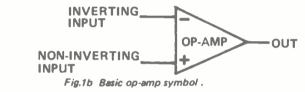


Fig. 1a Simplified op-amp equivalent circuit.

give a high input impedance and a high degree of common mode signal rejection. It also has a high-impedance (constant current) load to give a high degree of signal voltage stage gain.

The output of the differential amplifier is fed to a direct-coupled offset compensation stage, which 100 - July 1983 ETI

effectively reduces the output offset voltage of the differential amplifier to zero volts under quiescent conditions, and the output of the compensation stage is fed to a simple complementary emitter follower output stage, which gives a low output impedance.



LINES OF SUPPLY

Op-amps are normally powered from split power supplies, providing +ve, -ve, and common (zero volt) supply rails, so that the output of the op-amp can swing either side of the zero volts value, and can be set at a true zero volts (when zero differential voltage is applied to the circuits input terminals.)

The input terminals can be used independently (with the unused terminal grounded) or simultaneously, enabling the device to function as an inverting, non-inverting, or differential amplifier. Since the device is direct-coupled throughout, it can be used to amplify both a.c. and d.c. input signals. Typically, they give basic low-frequency voltage gains of about 100 000 between input and output, and have input impedances of 1M or greater at each input terminal.

Fig. 1b shows the symbol that is commonly used to represent an op-amp, and 1c shows the basic supply connections that are used with the device. Note that both input and output signals of the op-amp are referenced to the ground or zero volt line.

SIGNAL BOX

The output signal voltage of the op-amp is proportional to the DIFFERENTIAL signal between its two input terminals, and is given by

 $e_{out} = A_0(e_1 - e_2)$

where $A_0 =$ the open-loop voltage gain of the op-amp (typically 100 000).

 e_1 = signal voltage at the non-inverting input terminal.

e₂=signal voltage at the inverting input terminal.

Thus, if identical signals are simultaneously applied

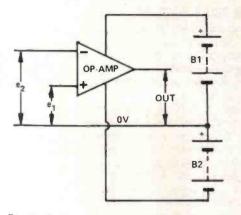


Fig. 1c Basic supply connections of an op-amp.

to both input terminals, the circuit will (ideally) give zero signal output If a signal is applied to the inverting terminal only, the circuit gives an amplified and inverted output. If a signal is applied to the non-inverting terminal only, the circuit gives an amplified but non-inverted output.

By using external negative feedback components, the stage gain of the op-amp circuit can be very precisely controlled.

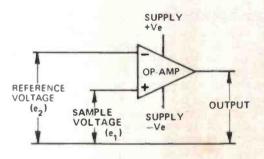


Fig. 2a Simple differential voltage comparator circuit.

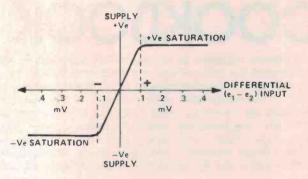


Fig. 2b Transfer characteristics of the differential voltage comparator circuit.

GOING TO GROUND

The op-amp can be made to function as a low-level inverting d.c. amplifier by simply grounding the non-inverting terminal and feeding the input signal to

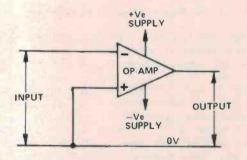


Fig. 3a Simple open-loop inverting d.c. am plifier.

TRANSFER REQUEST

Fig. 2a shows a very simple application of the op-amp. This particular circuit is known as a differential voltage comparator, and has a fixed reference voltage applied to the inverting input terminal, and a variable test or sample voltage applied to the non-inverting terminal. When the sample voltage is more than a few hundred microvolts above the reference voltage the op-amp output is driven to saturation in a positive direction, and when the sample is more than a few hundred microvolts below the reference voltage the output is driven to saturation in the negative direction.

Fig. 2b shows the voltage transfer characteristics of the above circuit. Note that it is the magnitude of the differential input voltage that dictates the magnitude of the output voltage, and that the absolute values of input voltage are of little importance. Thus, if a 1V reference is used and a differential voltage of only 200uV is needed to switch the output from a negative to a positive saturation level, this change can be caused by a shift of only 0.02% on a 1V signal applied to the sample input. The circuit thus functions as a precision voltage comparator or balance detector. the inverting terminal, as shown in Fig. 3a. The op-amp is used 'open-loop' (without feedback) in this configuration, and thus gives a voltage gain of about 100 000 and has an input impedance of about 1M. The disadvantage of this circuit is that its parameters are dictated by the actual op-amp, and are subject to considerable variation between individual devices.

CLOSING LOOPS

A far more useful way of employing the op-amp is to use it in the closed-loop mode, i.e., with negative feedback. Fig. 3b shows the method of applying negative feedback to make a fixed-gain inverting d.c. amplifier. Here, the parameters of the circuit are controlled by feedback resistors R_1 and R_2 . The gain, A of the circuit is dictated by the ratios of R_1 and R_2 , and equals R_2/R_1 .

The gain is virtually independent of the op-amp characteristics, provided that the open-loop gain (A_o) is large relative to the closed-loop gain (A). The input impedance of the circuit is equal to R_1 , and again is virtually independent of the op-amp characteristics.

OP-AMP COOKBOOK

It should be noted at this point that although R₁ and R₂ control the gain of the complete circuit, they have no effect on the parameters of the actual op-amp, and the full open-loop gain of the op-amp is still available between its inverting input terminal and the output. Similarly, the inverting terminal continues to have a very high input impedance, and negligible signal current flows into the inverting terminal. Consquently, virtually all of the R₁ signal current also flows in R₂, and

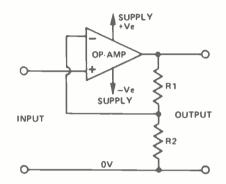


Fig. 4a Basic non-inverting d.c. amplifier

signal currents i_1 and i_2 can be regarded as being equal, as indicated in the diagram.

Since the signal voltage appearing at the output terminal end of R_2 is A times greater than that appearing at the inverting terminal end, the current flowing in R_2 is A times greater than that caused by the inverting terminal signal only. Consequently, R_2 has an apparent value of R_2/A when looked at from its inverting terminal end, and the R_1-R_2 junction thus appears as a low-impedance VIRTUAL GROUND point.

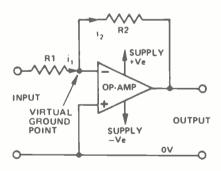


Fig. 3b Basic closed-loop inverting d.c. amplifier.

INVERT OR NOT TO INVERT ...

It can be seen from the above description that the Fig. 3b circuit is very versatile. Its gain and input impedance can be very precisely controlled by suitable choice of R_1 and R_2 , and are unaffected by variations in the op-amp characteristics. A similar thing is true of the non-inverting d.c. amplifier circuit shown in Fig. 4a. In this case the voltage gain is equal to $(R_1 + R_2)/R_2$ and the input impedance is approximately equal to (A_0/A) Zin where Zin is the open-loop input impedance of the op-amp. A great advantage of this circuit is that it has a very high input impedance.

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FOLLOW THAT VOLTAGE

The op-amp can be made to function as a precision voltage follower by connecting it as a unity-gain non-inverting d.c. amplifier, as shown in Fig. 4b. In this case the input and output voltages of the circuit are identical, but the input impedance is very high and is roughly equal to $A_0 \times Z_{in}$.

The basic op-amp circuits of Figs. 2a to 4b are shown as d.c. amplifiers, but can readily be adapted for a.c. use. Op-amps also have many applications other than as simple amplifiers. They can easily be made to function as precision phase splitters, as adders or subtractors, as active filters or selective amplifiers, as precision half-wave or full-wave rectifiers, and as oscillators or multivibrators, etc.

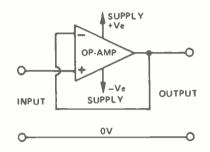


Fig. 4b Basic unity-gain d.c. voltage follower

OP-AMP PARAMETERS

An ideal op-amp would have an infinite input impedance, zero output impedance, infinite gain and infinite bandwidth, and would give perfect tracking between input and output. Practical op-amps fall far short of this ideal, and have finite gain, bandwidth, etc., and give tracking errors between the input and output signals. Consequently, various performance parameters are detailed on op-amp data sheets, and indicate the measure of "goodness" of the particular device. The most important of these parameters are detailed below.

OPEN-LOOP VOLTAGE GAIN, A. This is the low-frequency voltage gain occuring directly between the input and output terminals of the op-amp, and may be expressed in direct terms or in terms of dB. Typically, d.c. gain figures of modern op-amps are 100 000, or 100dB.

INPUT IMPEDANCE, Z_{in}. This is the impedance looking directly into the input terminals of the op-amp when it is used open-loop, and is usually expressed in terms of resistance only. Values of 1M are typical of modern op-amps with bi-polar input stages, while F.E.T. input types have impedances of a million meg or greater.

OUTPUT IMPEDANCE, Z_o. This is the output impedance of the basic op-amp when it is used open-loop, and is usually expressed in terms of resistance only. Values of a few hundred ohms are typical of modern op-amps.

INPUT BIAS CURRENT, I_b. Many op-amps use bipolar transistor input stages, and draw a small bias current from the input terminals. The magnitude of this current is denoted by I_{b} , and is typically only a fraction of a microamp.

SUPPLY VOLTAGE RANGE, V, Op-amps are usually operated from two sets of supply rails, and these supplies must be within maximum and minimum limits. If the supply voltages are too high the op-amp may be damaged, and if the supply voltages are too low the op-amp will not function correctly. Typical supply limits are $\pm 3V$ to $\pm 15V$.

INPUT VOLTAGE RANGE, V_{i(max)}. The input voltage to the op-amp must never be allowed to exceed the supply line voltages, or the op-amp may be damaged. V_{i(max)} is usually specified as being one or two volts less than v_e.

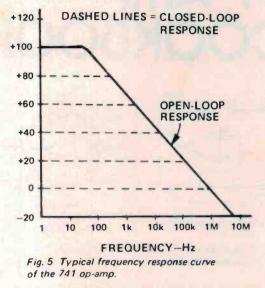
OUTPUT VOLTAGE RANGE, V_{o(max)}. If the opamp is over driven its output will saturate and be limited by the available supply voltages, so V_{o(max)} is usually specified as being one or two volts less than V_s.

DIFFERENTIAL INPUT OFFSET VOLTAGE, V_{io} . In the ideal op-amp perfect tracking would exist between the input and output terminals of the device, and the output would register zero when both inputs were grounded. Actual op-amps are not perfect devices, however, and in practice slight imbalances exist within their input circuitry and effectively cause a small offset or bias potential to be applied to the input terminals of the op-amp. Typically, this DIFFERENTIAL INPUT OFFSET VOLTAGE has a value of only a few millivolts, but when this voltage is amplified by the gain of the circuit in which the op-amp is used it may be sufficient to drive the op-amp output to saturation. Because of this, most op-amps have some facility for externally nulling out the offset voltage.

COMMON MODE REJECTION RATION, c.m.r.r. The ideal op-amp produces an output that is proportional to the difference between the two signals applied to its input terminals, and produces zero output when identical signals are applied to both inputs simultaneously, i.e., in common mode. In practical op-amps, common mode signals do not entirely cancel out, and produce a small signal at the op-amps output terminal. The ability of the op-amp to reject common mode signals is usually expressed in terms of common mode rejection ratio, which is the ratio of the op-amps gain with differential signals. C.m.r.r. values of 90dB are typical of modern op-amps.

TRANSITION FREQUENCY, f_{T} , An op-amp typically gives a low-frequency voltage gain of about 100dB, and in the interest of stability its open-loop frequency response is tailored so that the gain falls off as the frequency rises, and falls to unity at a transition frequency denoted f_{T} . Usually, the response falls off at a rate of 6dB per octave or 20dB per decade. Fig. 5 shows the typical response curve of the type 741 op-amp, which has an f_{T} of 1MHz and a low frequency gain of 100dB.

Note that, when the op-amp is used in a closed-loop amplifier circuit, the bandwidth of the circuit depends on the closed-loop gain. If the amplifier is used to give a gain of 60dB its bandwidth is only 1kHz, and if it is used to give a gain of 20dB its bandwidth is 100kHz. The f_{τ} figure can thus be used to represent a gain-bandwidth product.



	PARAMETER	741 VALUE
Ao	OPEN-LOOP VOLTAGE GAIN	100dB
ZIN	INPUT IMPEDANCE	1M
Zo	OUTPUT IMPEDANCE	150R
10	INPUT BIAS CURRENT	200nA
VSIMAXI	MAXIMUM SUPPLY VOLTAGE	18V
V. (MAK)	MAXIMUM INPUT VOLTAGE	:13V
Vo IMAX)	MAXIMUM OUTPUT VOLTAGE	:14V
Vio	DIFFERENTIAL INPUT OFFSET VOLTAGE	2mV
c.m.m.r.	COMMON MODE REJECTION RATIO	90dB
FT	TRANSITION FREDUENCY	1MHZ
S	SLEW RATE	1V/uS

Table 1 Typical characteristics of the 741 op-amp.

SLEW RATE. As well as being subject to normal bandwidth limitations, op-amps are also subject to a phenomenon known as slew rate limiting, which has the effect of limiting the maximum rate of change of voltage at the output of the device. Slew rate is normally specified in terms of volts per microsecond, and values in the range 1V/us to 10V/us are common with most popular types of op-amp. One effect of slew rate limiting is to make a greater bandwidth available to small output signals than is available to large output signals.

THE 741 OP-AMP.

Early types of i.c. op-amp, such as the well known 709 type, suffered from a number of design weaknesses. In particular, they were prone to a phenomenon known as INPUT LATCH-UP, in which the input circuitry tended to switch into a locked state if special precautions were not taken when connecting the input signals to the input terminals, and tended to self-destruct if a short circuit were inadvertently placed across the op-amp output terminals. In addition, the op-amps were prone to bursting into unwanted oscillations when used in the linear amplifier mode, and required the use of external frequency compensation components for stability control.

These weaknesses have been eliminated in the type 741 op-amp. This device is immune to input latch-up problems, has built-in output short circuit protection, and does not require the use of external frequency compensation components. The typical performance

OP-AMP COOKBOOK

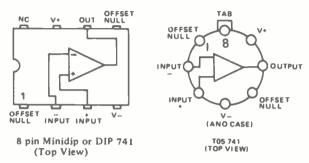


Fig. 6 Outlines and pin connections of the two most popular 741 packages.

characteristics of the device are listed in Table 1.

The type 741 op-amp is marketed by most i.c. manufacturers, and is very readily available. Fig. 6 shows the two most commonly used forms of packaging of the device Throughout this chapter, all practical circuits are based on the standard 8-pin dual-in-line (D.I.L. or DIP) version of the 741 op-amp.

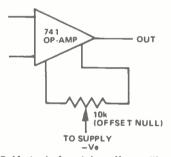


Fig. 7 Method of applying offset nulling to the 741 op-amp.

The 741 op-amp can be provided with external offset nulling by wiring a 10k pot between its two null terminals and taking the pot slider to the negative supply rail, as shown in Fig. 7.

Having cleared up these basic points, let's now go on and look at a range of practical applications of the 741 op-amp.

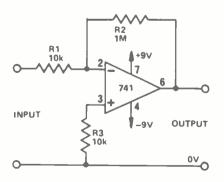


Fig. 8a x100 inverting d.c. amplifier.

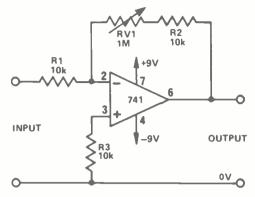


Fig. 8b Variable gain (x1 to x100) inverting d.c. amplifier.

BASIC LINEAR AMPLIFIER PROJECTS. (Figs. 8 to 11).

Figs. 8 to 11 show a variety of ways of using the 741 in basic linear amplifier applications.

The 741 can be made to function as an inverting amplifier by grounding the non-inverting input terminal and feeding the input signal to the inverting terminal. The voltage gain of the circuit can be precisely controlled by selecting suitable values of external feedback resistance. Fig. 8a shows the practical connections of an inverting d.c. amplifier with a pre-set gain of x100. The voltage gain is determined by the ratios of R₁ and R₂, as shown in the diagram.

The gain can be readily altered by using alternative R_1 and/or R_2 values. If required, the gain can be made variable by using a series combination of a fixed and a variable resistor in place of R_2 , as shown in the circuit of Fig. 8b, in which the gain can be varied over the range x1 to x100 via R_2 .

VARIATIONS

A variation of the basic inverting d.c. amplifier is shown in Fig. 9a. Here, the feedback connection to R_2 is taken from the output of the R_3-R_4 output potential divider, rather than directly from the output of the op-amp, and the voltage gain is determined by the ratios of this divider as well as by the values of R_1 and

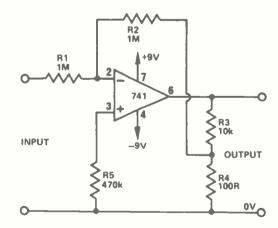


Fig. 9a High impedance x100 inverting d.c. amplifier.

 R_2 . The important feature of this circuit is that it enables R_1 , which determines the input impedance of the circuit, to be given a high value if required, while at the same time enabling high voltage gain to be achieved.

The basic inverting d.c. amplifier can be adapted for

a.c. use by simply wiring blocking capacitors in series with its input and output terminals, as shown in the x100 inverting a.c. amplifier circuit of Fig. 9b.

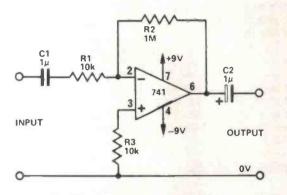


Fig. 9b x100 inverting a.c. amplifier.

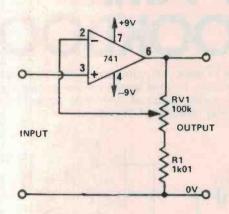


Fig. 10b Non-inverting variable gain (x1 to x100) d.c. amplifier.

BLOCKING OUT

The basic non-inverting d.c. circuit of Fig. 10 can be modified to operate as a.c. amplifiers in a variety of ways. The most obvious approach here is to simply wire blocking capacitors in series with the inputs and outputs, but in such cases the input terminal must be d.c. grounded via a suitable resistor, as shown by R_3 in the non-inverting x100 a.c. amplifier of Fig. 11a. If this resistor is not used the op-amp will have no d.c. stability, and its output will rapidly drift into saturation. Clearly, the input resistance of the Fig. 11a circuit is equal to R_3 , and R_3 must have a relatively low value in the interest of d.c. stability This circuit thus loses the non-inverting amplifier's basic advantage of high input resistance.

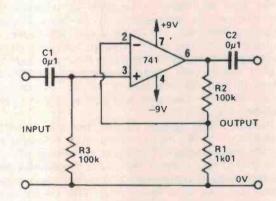


Fig. 11a Non-inverting, high input-impedance, x100 a.c. amplifier,

DRIFTING INTO STABILITY

A useful development of the Fig. 11a circuit is shown in Fig. 11b. Here, the values of R_1 and R_2 are increased and a blocking capacitor is interposed between them At practical operating frequencies this capacitor has a negligible impedance, so the voltage gain is still determined by the ratios of the two resistors. Because of the inclusion of the blocking capacitor, however, the inverting terminal of the op-amp is subjected to virtually 100% d.c. negative feedback from the output terminal of the op-amp, and the circuit thus has excellent d.c. stability. The low end of R_3 is

NON-INVERTING . . .

The amp can be made to function as a non-inverting amplifier by feeding the input signal to its non-inverting terminal and applying negative feedback to the inverting terminal via a resistive potential divider that is connected across the op-amp output. Fig. 10a shows the connections for making a fixed gain (x100) d.c. amplifier.

The voltage gain of the Fig. 10a circuit is determined by the ratios of R_1 and R_2 If R_2 is given a value of zero the gain falls to unity, and if R_1 is given a value of zero the gain rises towards infinity (but in practice is limited to the open-loop gain of the op-amp). If required, the gain can be made variable by replacing R_2 with a

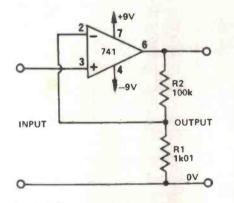


Fig. 10a Non-inverting x100 d.c. amplifier.

potentiometer and connecting the pot slider to the inverting terminal of the op-amp, as shown in the circuit of Fig. 10b The gain of this circuit can be varied over the range x1 to x100 via R₁.

... AND RESISTANCE TO INPUTS

A major advantage of the non-inverting d.c. amplifier is that it has a very high input resistance. In theory, the input resistance is equal to the open-loop input resistance (typically 1M) multiplied by the open-loop voltage gain (typically 100 000) divided by the actual circuit voltage gain. In practice, input resistance values of hundreds of megohms can readily be obtained.

OP-AMP COOKBOOK

connected to the C_3-R_1 junction, rather than directly to the ground line, and the signal voltage appearing at this point is virtually identical with that appearing at the non-inverting terminal of the op-amp

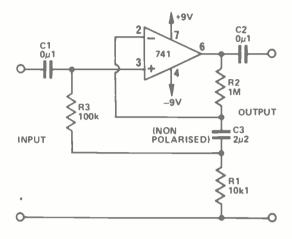


Fig. 11b Non-inverting x100 a.c. amplifier.

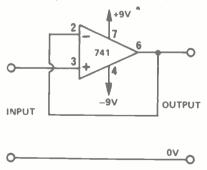
Consequently, identical signal voltages appear at both ends of R_3 , and the apparent impedance of this resistor is increased close to infinity by bootstrap action.

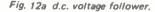
This circuit thus has good d.c. stability and a very high input impedance. In practice, this circuit gives a typical input impedance of about 50M.

VOLTAGE FOLLOWER PROJECTS (Figs. 12 to 13).

A 741 can be made to function as a precision voltage follower by connecting it as a unity-gain non-inverting amplifier. Fig. 12a shows the practical connections for making a d.c. voltage follower. Here, the input signal is applied directly to the non-inverting terminal of the op-amp, and the inverting terminal is connected directly to the output, so the circuit has 100% d.c. negative feedback and acts as a unity-gain non-inverting d.c. amplifier.

The output signal voltage of the circuit is virtually identical to that of the input, so the output is said to 'follow' the input voltage. The great advantage of this circuit is that it has a very high input impedance (as high as hundreds of megohms) and a very low output impedance (as low as a few ohms). The circuit acts effectively as an impedance transformer.





PRACTICE, AND ITS LIMITS

In practice the output of the basic Fig. 12a circuit will follow the input to within a couple of millivolts up to magnitudes within a volt or so of the supply line potentials. If required, the circuit can be made to follow to within a few microvolts by adding the offset null facility to the op-amp.

The d.c. voltage follower can be adapted for a.c. use by wiring blocking capacitors in series with its input and output terminals and by d.c.-coupling the non-inverting terminal of the op-amp to the zero volts line via a suitable resistor, as shown by R_1 in Fig. 12b. R_1 should have a value less than a couple of megohms, and restricts the available input impedance of the voltage follower.

LACED UP OHMS

If a very high input-impedance a.c. voltage follower is needed, the circuit of Fig. 12c can be used. Here, R1 is boostrapped from the output of the op-amp, and its apparent impedance is greatly increased. This circuit has a typical impedance of hundreds of megohms.

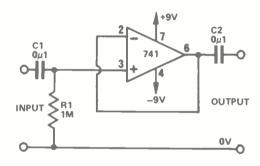


Fig. 12b a.c. voltage follower.

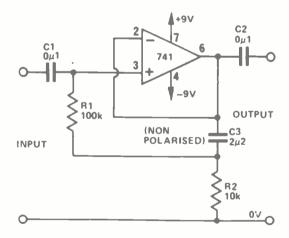


Fig. 12c Very high input-impedance a.c. voltage follower.

DRIVING CIRCUITS AMP-LY

The 741 op-amp is capable of providing output currents up to about 5mA, and this is consequently the current-driving limit of the three voltage follower circuits that we have looked at so far. The current-driving capabilities of the circuits can readily be increased by wiring simple or complementary emitter follower booster stages between the op-amp output terminals and the outputs of the actual circuits, as shown in Figs. 13a and 13b respectively.

Note in each case that the base-emitter junction(s) of

the output transistor(s) are included in the negative feedback loop of the circuit: Consequently, the 600mV knee voltage of each junction is effectively reduced by a factor equal to the open-loop gain of the op-amp, so the junctions do not adversely effect the voltage-following characteristics of either circuit.

The Fig. 13a circuit is able to source current only, and can be regarded as a unidirectional, positive-going, d.c. voltage follower. The Fig. 13b circuit can both source and sink output currents, and thus gives bidirectional follower action. Each circuit has a current-driving capacity of about 50mA. This figure is dictated by the limited power rating of the specified output transistors. The drive capability can be increased by using alternative transistors.

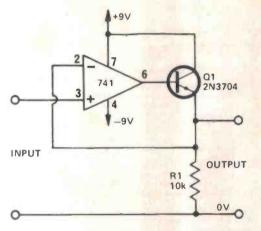


Fig. 13a Unidirectional d.c. voltage follower with boosted output (variable from 0V to +8V at 50mA.)

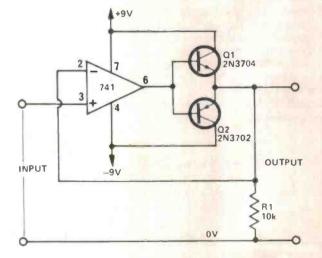


Fig. 13b Bidirectional d.c. voltage follower with boosted output (variable from 0V to $\pm \pm 8V$ at 50mA).

MISC AMP PROJECTS (Figs. 14 to 22)

Figs. 14 to 22 show a miscellaneous assortment of 741 amplifier projects, ranging from d.c. adding circuits to frequency-selective amplifiers.

Fig. 14 shows the circuit of a unity-gain inverting d.c. adder, which gives an output voltage that is equal to the sum of the three input voltages. Here, input resistors R, to R_a and feedback resistor R_a each have the same value, and the circuit thus acts as a unity-gain inverting d.c. amplifier between each input terminal and the output. Since the current flowing in each input resistor also flows in feedback resistor R4, the total current flowing in R, is equal to the sum of the input currents, and the output voltage is equal to the negative sum of the input voltages. The circuit is shown with only three input connections, but in fact can be provided with any number of input terminals. The circuit can be made to function as a so-called 'audio mixer' by wiring blocking capacitors in series with each input terminal and with the output terminal.

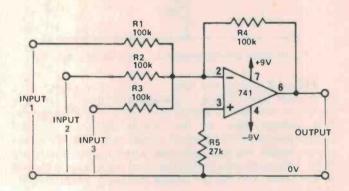


Fig. 14 Unity-gain inverting d.c. adder, or 'audio mixer'.

FIG. 15 shows how two unity-gain inverting d.c. amplifiers can be wired in series to make a precision unity-gain balanced phase-splitter. The output of the first amplifier is an inverted version of the input signal, and the output of the second amplifier is a non-inverted version.

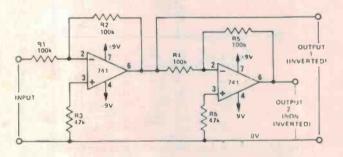


Fig. 15 Unity-gain balanced d.c. phase-splitter.

FIG. 16 shows how a 741 can be used as a unity-gain differential d.c. amplifier. The output of this circuit is equal to the difference between the two input signals or voltages, or to $e_1 \cdot e_2$. Thus, the circuit can also be used as a subtractor. In this type of circuit the component values are chosen such that $R_1/R_2 = R_4/R_3$, in which case the voltage gain $A_v = R_2/R_1$. The circuit can thus be made to give voltage gain if required.

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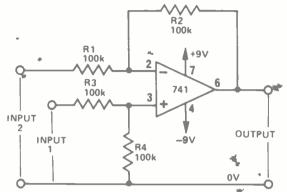


Fig. 16 Unity-gain differential d.c. amplifier, or subtractor.

FIG. 17 shows the amp can be made to act as a non-linear (semi-log) a.c. voltage amplifier by using a couple of ordinary silicon diodes as feedback elements. The voltage gain of the circuit depends on the magnitude of applied input signal, and is high when input signals are low, and low when input signals are high. The measured performance of the circuit is shown in the table, and can be varied by using alternative R₁ values.

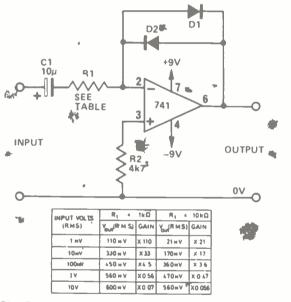


Fig. 17 Circuit and performance table of non-linear (semi-log) a.c. voltage amplifier.

FIG. 18 shows how the 7Å1 can be used together with a junction-type field-effect transistor (JFET) to make a so-called constant-volume amplifier. The action of this type of circuit is such that its peak output voltage is held sensibly constant, without distortion, over a wide range of input signal levels, and this particular circuit gives a sensibly constant output over a 30dB range of input signal levels.

The measured performance of the circuit is shown in the table. C_1 determines the response time of the

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amplifier, and may be altered to satisfy individual needs.

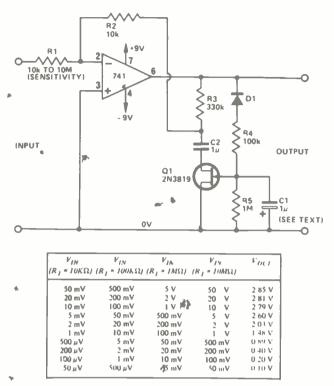


Fig. 18 Circuit and performance details of constant-volume amplifier.

ACTION TAKEN

The action of the Fig. 18 circuit relies on the fact that the JFET can act as a voltage-controlled resistance which appears as a low value when zero bias is applied to its gate and as a high resistance when its gate is negatively biased. The JFET and R_3 act as a gain-determining a.c. voltage divider (via C_2), and the bias to the JFET gate is derived from the circuits output via the D_1 - C_1 network. When the circuit output is low the JFET appears as a

low resistance, and the op-amp gives high voltage gain. When the circuit output is high the JFET appears as a high resistance, and the op-amp gives low voltage gain. The output level of the circuit is thus held sensibly constant by negative feedback.

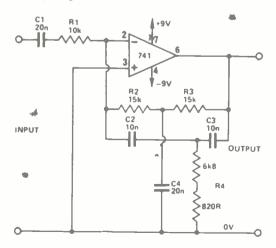


Fig. 19 1kHz tuned (acceptor) amplifier (twin-T).

CHOOSE YOUR FREQUENCY

The 741 op-amp can be made to function as a frequency-selective amplifier by connecting frequency-sensitive networks into its feedback loops. Fig. 19 shows how a twin-T network can be connected to the op-amp so that it acts as a tuned (acceptor) amplifier, and Fig. 20 shows how the same twin-T network can be connected so that the op-amp acts as a notch (rejector) filter. The values of the twin-T network are chosen such that $R_2=R_3=2 \times R_4$, and $C_2=C_4/2$, in which case its centre (tuned) frequency = 1/6.28 R_2 . C₂. With the component values shown, both circuits are tuned to approximately 1kHz.

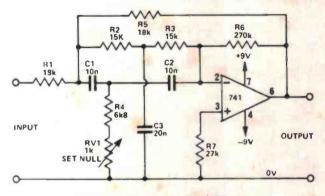
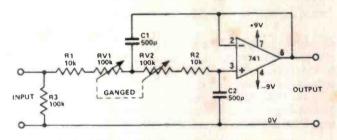
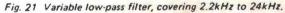


Fig. 20 1kHz notch (reject) filter.





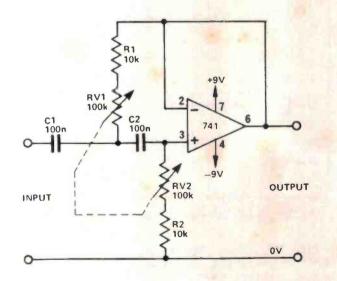
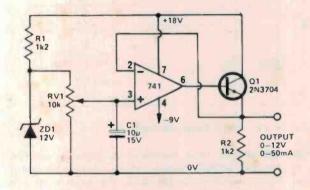


Fig. 22 Variable high-pass filter, covering 235Hz to 2.8kHz.

Finally, to complete this section, Figs. 21 and 22 show the circuits of a couple of variable-frequency audio filters. The Fig. 21 circuit is that of a low-pass filter which covers the range 2.2kHz to 24kHz, and the Fig. 22 circuit is that of a high-pass filter which covers the range 235Hz to 2.8kHz. In each case, the circuit gives unity gain to signals beyond its cut-off frequency, and gives a 2nd order response (a change of 12dB per octave) to signals within its range.

INSTRUMENTATION PROJECTS (Figs. 23 to 31)

Figs. 23 to 31 show a variety of instrumentation projects in which the 741 can be used. The circuits range from a simple voltage regulator to a linear-scale ohmmeter.



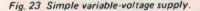


FIG. 23 shows the circuit of a simple variable-voltage power supply, which gives a stable output that is fully adjustable from OV to 12V at currents up to a maximum of about 50mA. The operation of the circuit is quite simple. ZD_1 is a zener diode, and is energised from the positive supply line via R_1 . A constant reference potential of 12V is developed across the zener diode, and is fed to variable potential divider RV_1 .

The output of this divider is fully variable from OV to 12V, and is fed to the non-inverting input of the op-amp. The op-amp is wired as a unity-gain voltage follower, with Q_1 connected as an emitter follower current-booster stage in series with its output.

Thus, the output voltage of the circuit follows the voltage set at the op-amp input via RV₁, and is fully variable from OV to 12V. Note that the circuit uses an 18V positive supply and a 9V negative supply.

Also note that the voltage range of the above circuit can be increased by using higher zener and unregulated supply voltages, and that its current capacity can be increased by using one or more power transistors in place of Q_1 .

FIG. 24 shows how a 741 op-amp can be used as the basis of a stabilised power supply unit (P.S.U.) that covers the range 3V to 30V at currents up to 1A. Here, the voltage supply to the op-amp is stabilized at 33V via ZD₁, and a highly temperature-stable reference of 3V is fed to the input of the op-amp via ZD₂.

The op-amp and output transistors Q_1-Q_2 are wired as a variable-gain non-inverting d.c. amplifier, with gain variable from unity to x10 via RV₁, and the output voltage is thus fully variable from 3V to 30V via RV₁. The output voltage is fully stabilized by negative feedback.

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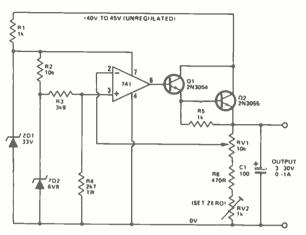


Fig. 24 3V - 30V, 0-1 amp stabilised p.s.u.

FIG. 25 shows how overload protection can be applied to the above circuit. Here, current-sensing resistor R, is wired in series with the output of the regulator, and cut-out transistor Q_3 is driven from this resistor and is wired so that its base-collector junction is able to short the base-emitter junction of the Q_1 - Q_2 output transistor stage.

Normally, Q_3 is inoperative, and has no effect on the circuit, but when P.S.U. output currents exceed 1A a potential in excess of 600mV is developed across R, and biases Q_3 on, thus causing Q_3 to shunt the base-emitter junction of the Q_1 - Q_2 output stage and hence reducing the output current. Heavy negative feedback takes place in this action, and the output current is automatically limited to 1A, even under short-circuit conditions.

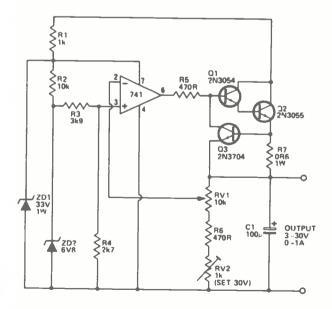


Fig. 25 3V - 30V stabilised p.s.u. with overload protection. 110 - July 1983 ETI

FIG. 26a shows how a 741 can be used in conjunction with a couple of silicon diodes as a precision half-wave rectifier. Conventional diodes act as imperfect rectifiers of low-level a.c. signals, because they do not begin to conduct significantly until the applied signal voltage exceeds a 'knee' value of about 600mV.

When diodes are wired into the negative feedback loop of the circuit as shown the 'knee' voltage is effectively reduced by a factor equal to the open-loop gain of the op-amp, and the circuit thus acts like a near-perfect rectifier

The overall voltage gain of the Fig. 26a circuit is dictated by the ratios of R_1 and R_2 to R_3 , as in the case of a conventional inverting amplifier, and this circuit thus gives a gain of unity. The circuit can be made to

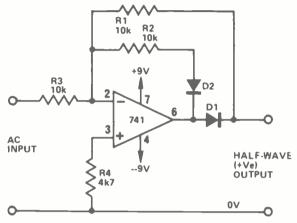


Fig. 26a Precision unity-gain half-wave rectifier.

act as a precision half-wave a.c./d.c. converter by designing it to give a voltage gain of 2.22 to give form-factor correction, and by integrating its rectifier output, as shown in Fig. 26b.

Note that each of the Fig. 26 circuits has a high output impedance, and the outputs must both be fed into loads having impedances less than about 1M.

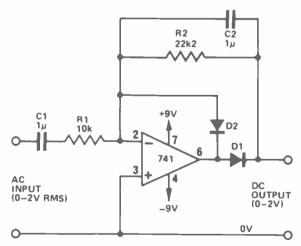


Fig. 26b Precision half-wave a.c./d.c. converter.

FIG. 27 shows how op-amp can be used as a high-performance d.c. voltmeter converter, which can be used to convert any 1V f.s.d. meter with a sensitivity better than 1k/V into a voltmeter that can read any

value in the range 1mV to 10V f.s.d. at a sensitivity of 1M/V. The voltage range is determined by the R₁ value, and the table shows some suitable values for common voltage ranges.

FIG. 28 shows a simple circuit that can be used to convert a 1mA f.s.d. meter into a d.c. voltmeter with any f.s.d. value in the range 100mV to 1000V, or into a d.c. current meter with any f.s.d. value in the range 1uA to 1A. Suitable component values for different ranges are shown in the tables.

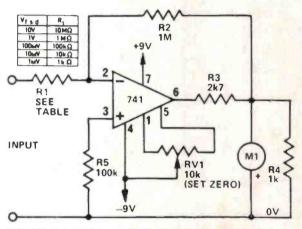
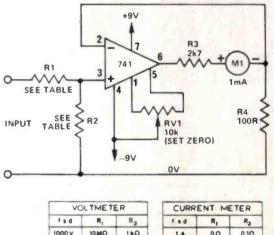


Fig. 27 High-performance d.c. voltmeter converter.



1000 V	10 M Q	110	1.4	00	01Ω
100V	10MQ	10 k 🖸	100 mA	0Ω	10
IOV	10MΩ	100kQ	10MA	00	100
11	900k Q	100k D	Teat	00	1000
100 m V	0Ω	100kQ	100 # A	οQ	140
			IOMA	00	iOkΩ
			1µA	QQ	100M2

Fig. 28 Simple d.c. voltage or current meter.

FIG. 29 shows the circuit of a precision d.c. millivoltmeter, which uses a 1mA f.s.d. meter to read f.s.d. voltages from 1mV to 1000mV in seven switch-selected ranges.

FIG. 30 shows the basic circuit of a precision a.c. volt or millivolt meter. This circuit can be used with any moving-coil meter with a full scale current value in the range 100uA to 5mA, and can be made to give any full scale a.c. voltage reading in the range 1mV to 1000mV. The tables show the alternative values of R_1 and R_2 that must be used to satisfy different basic meter sensitivities, and the values of R_3 and R_4 that must be used for different f.s.d. voltage sensitivities.

HOME OHM

Finally, to conclude, Fig. 31 shows how the 741 op-amp can be used in conjuncton with a 1mA f.s.d. meter to make a linear-scale ohmmeter that has five decade ranges from 1k to 10M.

The circuit is divided into two parts, and consists of a voltage generator that is used to generate a standard test

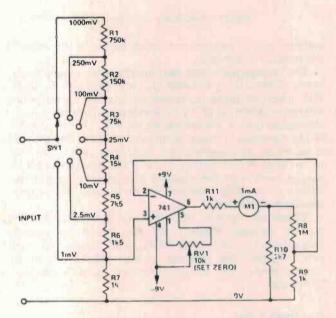


Fig. 29 Precision d.c. millivoltmeter.

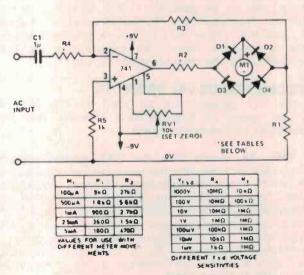
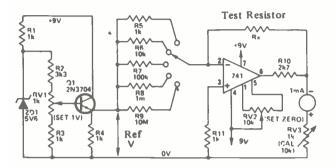


Fig. 30 Precision a.c. volt/millivolt meter.

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voltage, and a readout unit which indicates the value of the resistor under test.

The voltage generator section of the circuit comprises zener diode ZD₁, transistor Q₁, and resistors R₁ to R₄. The action of these components is such that a stable reference potential of 1V is developed across R₄, but is adjustable over a limited range via RV₁. This voltage is fed to the input of the op-amp readout unit. The op-amp is wired as an inverting d.c. amplifier, with the 1mA meter and RV₃ forming a 1V f.s.d. meter across its output, and with the op-amp gain determined by the values of ranging resistors R₅ to R₉ and by negative feedback resistor R_x.

Since the input to the amplifier is fixed at 1V, the output voltage reading of the meter is directly proportional to the value of R_x , and equals full scale when R_x and the ranging resistor values are equal. Consequently, the circuit functions as a linear-scale ohmmeter.

CALIBRATION

3

The procedure for initially calibrating the Fig. 31 circuit is as follows: First, switch the unit to 10k range and fix an accurate $10k\Omega$ resistor in the R_x position. Now adjust RV_1 to give an accurate 1V across R_4 , and then adjust RV_2 to give a precise full scale reading on the meter. All adjustments are then complete, and the circuit is ready for use.

MISCELLANEOUS 741 PROJECTS

The 741 op-amp can be used as the basis of a vast range of miscellaneous projects, including oscillators and sensing circuits.

FIG. 32 shows how the 741 op-amp can be connected as a variable-frequency wien-bridge oscillator, which covers the basic range 150Hz to 1.5kHz, and uses a low-current lamp for amplitude stabilisation. The output amplitude of the oscillator is variable via RV₄ and has a typical maximum value of 2.5V r.m.s. and a t.h.d. value of 0.1%. The frequency range of the circuit is inversely proportional to the C₁-C₂ values. The circuit can give a useful performance up to a maximum frequency of about 25kHz.

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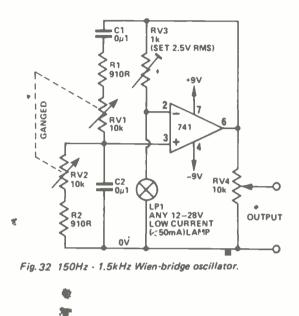


Fig. 33 shows how either a 741 or a 709 op-amp can be connected as a simple variable-frequency square-wave generator that covers the range 500Hz to 5kHz via a single variable resistor. (The circuit produces a good symmetrical waveform.)

The frequency of oscillation is inversely proportional to the C₁ value, and can be reduced by increasing the C₁ value, or vice-versa. The amplitude of the square wave output signal can be made variable, if required, by wiring a $10k\Omega$ variable potential divider across the output terminals of the circuit and taking the output from between the pot slider and the zero volts line.

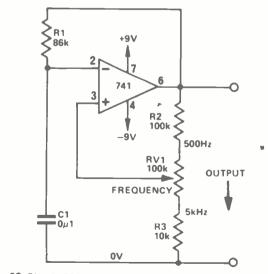


Fig. 33 Simple 500Hz - 5kHz square wave generator.

The op-amp cookbook — using the LM 3900N

Four op-amps in a single 14-pin package — at less than 25 cents per amplifier.

THESE days it's nothing unusual to find four op-amps in a single integrated circuit package, but when the LM3900 was released by National Semiconductor in the mid-seventies very few linear devices were so closely-packed. The LM3900 contains four independent, internally compensated amplifiers in a single 14-pin dual-in-line encapsulation.

All four amplifiers are fabricated on a single silicon chip. Each amplifier contains seven transistors, a diode and a capacitor, whilst other internal components are used in the bias and power supplies.

The LM3900 has maintained its popularity over the years, partly because of its low cost (less than a dollar in 1982), but mainly because of its versatility and reliability.

CONNECTIONS

The connections of the four separate amplifiers are shown in Fig.1. Each amplifier has a non-inverting input (marked +), an inverting input (marked -) and an output connection.

In addition, there is a single common positive supply connection and a common ground connection (negative supply line) for the whole device.

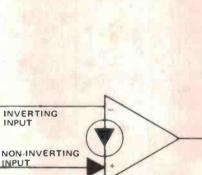
INTERNAL CIRCUIT

Conventional high gain amplifiers employ a differential input stage to provide inverting and non-inverting inputs, but a rather different approach is employed in the LM3900N. A 'current mirror' is employed in the non-inverting input circuit, the current 'reflected' in this mirror being subtracted from that which enters the inverting input.

This type of amplifier therefore acts as a differential stage by amplifying the difference between two *currents* rather than the difference between two voltages (as in a conventional amplifier).

The type of amplifier used in the LM3900N may be referred to as a 'Norton' amplifier, since Norton is the name of the person who developed a theorem relating the *current* flowing in a circuit to the equivalent current generator and shunt impedance.

Fig.1. The connections of the LM3900N.



SYMBOL

The symbol recommended for each of the four Norton amplifier stages in the device is shown in Fig.2. This symbol distinguishes this type of amplifier from the standard operational amplifier symbol and avoids confusion in circuits.

The symbol of Fig. 2 contains an indication that there is a current source between the inverting and non-inverting inputs and implies that the amplifier uses a current mode of operation. In addition, the circuit symbol indicates that current is removed from the inverting input, whilst the arrow on the non-inverting input shows that this functions as a current input.

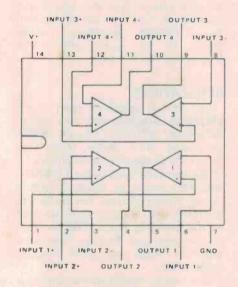


Fig.2. The symbol for one of the Norton amplifiers of the LM3900N.

PERFORMANCE

OUTPUT

The LM3900N has the advantage that it can operate from a single supply voltage over the range of four volts to 36 volts. Most conventional operational amplifiers require supplies symmetrical with respect to ground (typically \pm 15 V); the LM3900N can be used with such supply lines if desired.

The maximum peak to peak output amplitude of an LM3900N amplifier is only 1 V less than the supply voltage employed. The current consumed from the power supply is typically 6.2 mA (maximum 10 mA).

The typical voltage gain of each amplifier is 2800 or nearly 70 dB. The

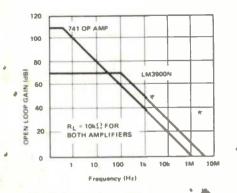


Fig. 3. Comparison of the gain of the LM3900N with that of a 741 amplifier, at various frequencies.

minimum gain of any amplifier is 1200. The variation of this gain with frequency is compared with that of the well known type 741 operational amplifier in Fig. 3. It can be seen that the LM3900N amplifiers provide about 10 dB more gain at all frequencies above 1 kHz.

APPLICATIONS

The Norton amplifiers used in the LM3900N device entail the use of somewhat different circuit design techniques than those used with conventional operational amplifiers.

The inverting input of the LM3900N amplifiers must be supplied with a steady biasing current. The current to the non-inverting input modulates that to the inverting input. The fact that current can pass between the input terminals leads to some unusual applications.

Both inputs of each of the amplifiers in the LM3900N are clamped by diodes so as to keep their potentials almost constant at one diode voltage drop (about 0.5 V) above the ground potential of pin 7. External input voltages must therefore be converted to input currents by placing series resistors in each input circuit.

USE AS AN AC AMPLIFIER

The LM3900N forms a useful ac amplifier, since its output can be baised to any desired steady voltage within the range of the output voltage * •

swing. The ac gain is independent of the biasing level and the single power supply required greatly simplifies circuit design.

A simple ac amplifier circuit is shown in Fig.4. The gain is approximately equal to R2/R1 or 10 with the circuit values shown. The mean potential at the output is half the supply voltage. The value of R3 should be twice that of R2 since the current passing through each of these two resistors is then the same. The positive supply and ground connections are not shown in Fig.4

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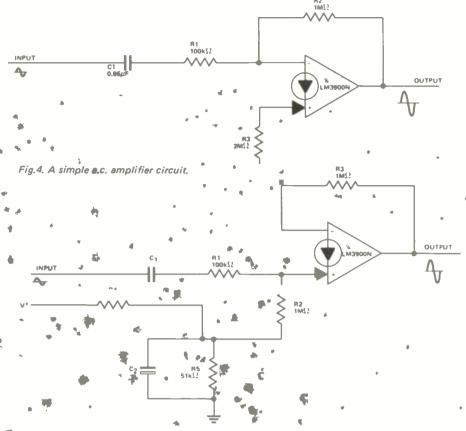


Fig.5. A simple non-inverting a.c. amplifier.

Fig.6. An amplifier which has a high gain and a high input impedance.

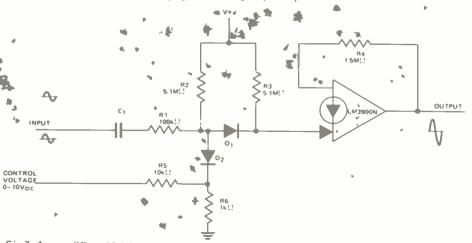
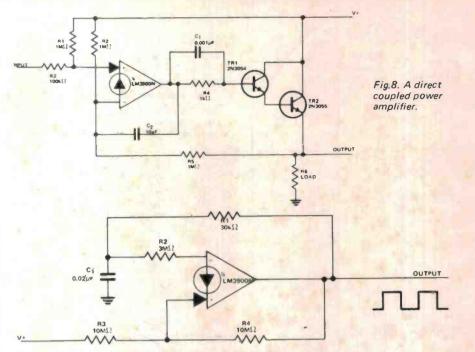


Fig.7. An amplifier which has a gain controlled by an input voltage.





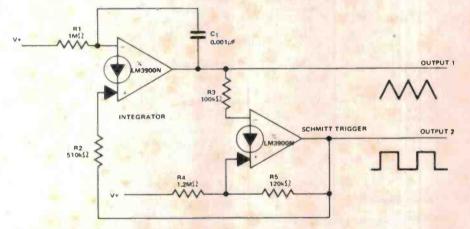


Fig. 10. A clrcuit for generating triangular and square-wave.

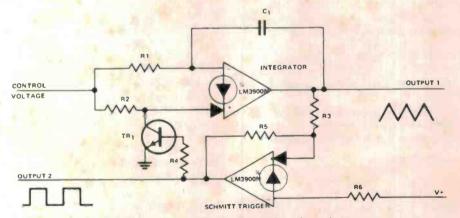


Fig. 11. A voltage controlled oscillator which produces triangular and square-waves.

for simplicity, but R_3 should be returned to the same positive supply line as that used to feed pin 14.

The circuit of Fig. 4 provides a phase inverted output. Any ripple on the power supply line will appear on the output at half amplitude.

NON-INVERTING AC AMPLIFIER

The circuit of Fig. 5 shows an amplifier which provides an output in phase with the input. The gain is equal to $R_3/(R_1 + r_d)$ where r_d is the small signal impedance of the input diode. The value of r_d is equal to 0.026

divided by the current passing through R₂ to the non-inverting input.

The capacitor values should be chosen so that the impedance of these components is considerably less than the circuit impedance at the points concerned.

HIGH IMPEDANCE AND HIGH GAIN

The circuits of Figs. 4 and 5 have an input resistance, R_1 or 100 k ohm. If this resistor is increased to provide a higher input impedance, the gain of the circuit will fall. However, the circuit of Fig. 6 has been designed so that it provides both a high input impedance and a high gain using a simple amplifier. With the component values shown, the input impedance is one megohm and the gain 100.

The voltage applied to R_2 is made equal to the output voltage (which is half the supply voltage). The value of R_2 is equal to the sum of R_3 and R_4 ; these resistors set the dc bias. If desired, R_2 may be made four megohms and its lower end connected to the V+ supply.

Resistors R_4 and R_5 form a potential divider so that only 1/100 of the alternating output voltage is developed across the $C_2 - R_5$ circuit. This fraction of the output voltage is fed back to the inverting input via R_3 . As R_3 and R_1 are equal, the gain is $R_4 R_5$. As R_5 is decreased, the gain approaches the open loop gain of the amplifier.

VOLTAGE CONTROLLED GAIN

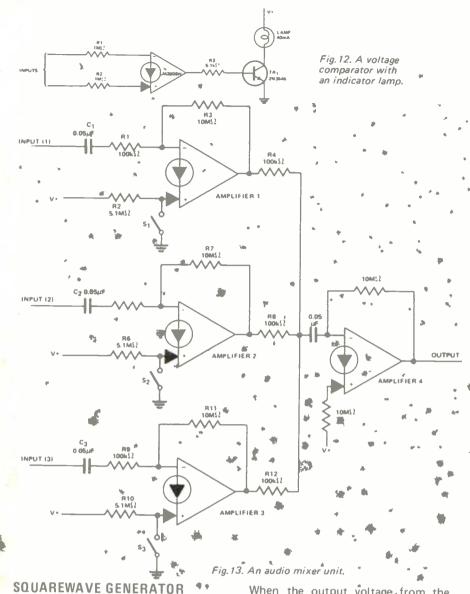
An amplifier with a gain which can be controlled by the value of a steady applied voltage is shown in Fig. 7.

A current flows from the positive supply through R_3 to provide a bias which prevents the output of the amplifier from being driven to saturation as the control voltage is varied. When D_2 is non-conducting, the currents passing through both R_2 and R_3 enter the non-inverting input and the gain is of maximum. This occurs when the control voltage approaches 10 V.

The gain is a minimum when the control voltage is zero. In this case D_2 is conducting and only the current passing through R_3 enters the non-inverting input of the amplifier.

DIRECT COUPLED POWER AMPLIFIER

In the circuit of Fig. 8, the output from an LM3900N amplifier is fed to a Darlington pair of power transistors. This circuit can deliver over three amps into a suitable load when the transistors are correctly mounted on heat sinks.



The multiple amplifiers in the LM3900N device are very suitable for current flowing through R2 use in waveform generators at input voltage) can also be designed using the device.

A simple square wave generator is • output 1. shown in Fig. 9. The capacitor C1 alternately charges and discharges between voltage limits which are set by R_2 R_3 and R_4 . The circuit is basically of the Schmitt trigger type, the voltages at which triggering occurs being approximately V+/3 and 2V+/3.

TRIANGULAR WAVEFORM GENERATOR

A triangular waveform generator can be made by using one amplifier of a LM3900N device as an integrator and " another amplifier as a Schmitt trigger circuit. A suitable circuit is shown in Fig. 10; it has the unusual advantage that only the one power supply is required.

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When the output voltage from the Schmitt* trigger circuit is low, the is integrated, by C1 to produce the ij, frequencies of up to about 10 kHz. Degative slope of the triangular wave at Voltage controlled oscillators (the Wooutput 1. When the output 2 voltage frequency of which is dependent on an . from the Schmitt trigger is high, current flows through R₂ to produce the rising part of the waveform at 1 11

The output waveform will have good symmetry if $R_1 = 2R_2$. The output frequency is given by the equation:

$$f = \frac{V^+ - V_{BE}}{2R_1C_1 V}$$

where R1 · 2R2, VBE is the steady voltage at the inverting input (0.5 V) and V is the difference between the tripping points of the Schmitt trigger.

VOLTAGE CONTROLLED OSCILLATOR

A simple voltage controlled oscillator circuit which produces both triangular and square wave outputs is shown in Fig. 11. As in Fig.10, one amplifier is employed as an integrator.

When the output of the Schmitt trigger is high, the clamp transistor TR₁ is conducting and the input current passing through Ro is shunted to ground. The current passing through R1 causes a falling ramp to be formed.

When the Schmitt circuit changes state, its output switches TR₁ to the • non-conducting state. The current flowing through R₂ can be made twice that flowing through R_1 ($R_2 = R_1/2$) so that the rising part of the ramp has a similar slope to the negative part.

The greater the value of the control voltage in Fig.11, the greater the frequency of oscillation. However, the voltage must exceed the constant input voltage (VBE) or the circuit will fail to oscillate.

VOLTAGE COMPARATOR

The circuit of Fig. 12 shows how an LM3900N amplifier may be employed to compare two input voltages and to indicate the result by means of a small lamp. If the input voltage connected to the non-inverting input is appreciably more positive, than the , other winput, the output of the amplifier will provide a positive voltage which renders TR₂ conducting. The lamp will then be illuminated.

One of the inputs may be a reference voltage so that one can then compare a single input voltages against this constant reference.

AUDIO MIXER

The amplifiers of a LM3900N device can be conveniently used to make a mixer unit for audio purposes; the unit enables three separate audio signals to be mixed together to produce a composite output. The circuit shown w in Fig. 13 provides this facility using only a single LM3900N device and also enables any one channel to be selected by switches. The currents passing through the resistors R₄ R₈ and R₁₂ are summed in the input circuit of the fourth amplifier.

If S_1 is open, amplifier 1 will be driven to saturation by the current. passing through R2. It will therefore be inactive.

CONCLUSION

This short article has attempted to show a few of the numerous applications of this economical integrated circuit. Many more applications (such as phase locked loops, temperature sensing circuits, differentiators, tachometers, staircase generators, active filters, etc) are given in a report AN-72 produced by National Semiconductor.

The op-amp cookbook — using the 3080

A 'control' current varies the gain of this op-amp, which makes it very useful in some special applications.

THE CA3080 IS KNOWN as an operational transconductance amplifier (OTA). This is a type of op-amp, the gain of which can be varied by means of a control current, (I_{ABC}). The device has a differential input, a control input known as the 'amplifier bias input' and a current output. It differs in many respects from conventional op-amps and it is these differences that can be used to realise many useful circuit blocks.

Voltage controlled amplifier

The CA3080 can be used as a gain controlling device. A useful circuit is shown in Figure 1. The input signal is attenuated by R1, R2 such that a 20 mV peakto-peak signal is applied to the input terminals. If this voltage is much larger, then significant distortion will occur at the output. In fact, this distortion is put to good use in the triangle-tosinewave converter. (Figure 3, but we're jumping the gun).

The gain of the circuit is controlled by the magnitude of the current I_{ABC} . This current flows into the CA3080 at pin 5, which is held at one diode voltage drop above the – Vcc rail. If you connect pin 5 to 0 V, then this diode will get zapped (and so will the IC!). The maximum value of I_{ABC} permitted is 1 mA and the device is 'linear' over four decades of this current. That is, the gain of the CA 3080 is 'linearly' proportional to the magnitude of the I_{ABC} current over a range of 0.1 uA to 1 mA. Thus, by controlling I_{ABC} , we can control the signal level at the output.

The output is a current output which has to be 'dumped' into a resistive load (R5) to produce a voltage output. The output impedance seen at IC1 pin 6 is 10k (R5), but this is 'unloaded' by the voltage follower (IC2) to produce a low output impedance.

The circuit involving IC3 is a precision voltage-to-current converter and this can be used to generate I_{ABC} . When Vin (control) is positive, it linearly controls the gain of the circuit. When it is negative, I_{ABC} is zero and so the gain is zero. This type of circuit is known by several names. It is a voltage controlled amplifier; (VCA), or an amplitude modulator, or a two quadrant multiplier.

One problem that occurs with the CA3080 is that of the 'input offset voltage'. This is a small voltage diffe-

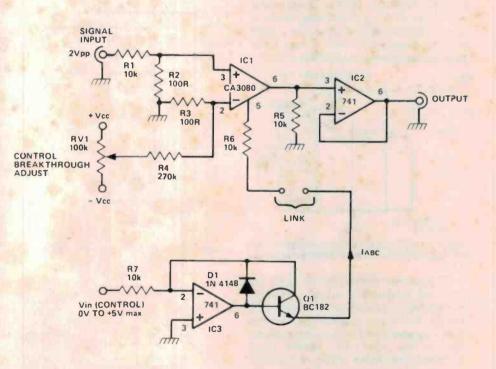


Figure 1. A voltage controlled amplifier. Gain is varied by varying RV1. You can modulate a signal passing through the amplifier by joining the 'link' and applying a modulating signal to the input of IC3 (at R7). This sort of circuit is also known as a 'two quadrant multiplier'.

OP-AMP COOKBOOK

SIGNAL AMAAAAAAAA OV



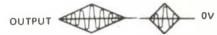


Figure 2. Illustrating the operation of the voltage controlled amplifier shown in Figure 1.

rence, or 'offset', between its input terminals. When there is no signal input and the control input is varied, a voltage similar to the control input will appear at the output. By adjusting RV1 it is possible to null out most of this control breakthrough.

The effect of modulating Vin (control) is illustrated in Figure 2.

Triangle to sinewave converterth

By overloading the input of a CA3080 it is possible to produce a 'sinusoidal' transfer function. That is, if a triangle waveform of the correct magnitude is applied to the CA3080 input, the output will be distorted in such a way as to produce a sinewave approximation.

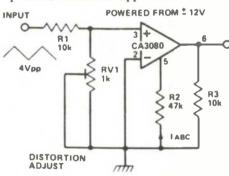


Figure 3. This circuit will convert a triangle wave to a sinewave with a resultant distortion of around 1.8%.

In the circuit shown (Figure 3), RV1 is adjusted so that the output waveform resembles a sinewave. I tested this circuit using an automatic distortion analyser and found the sinewave distortion to be only 1.8%, mostly third harmonic distortion which, for such a simple arrangement, seems very reasonable indeed. This could be used to produce a sinewave output from a triangle/square wave oscillator.

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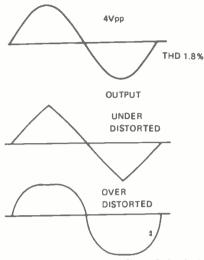
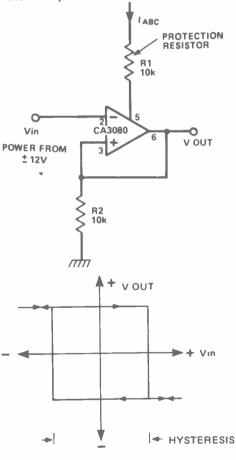


Figure 4. The output of the Figure 3 circuit should be adjusted (by RV1) to produce the waveform shown at top.



LABC CONTROLS HYSTERESIS



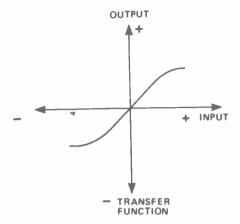


Figure 5. Transfer function of the Figure 3 circuit.

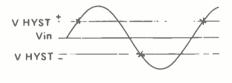




Figure 7. How the Schmitt trigger of Figure 6 works.

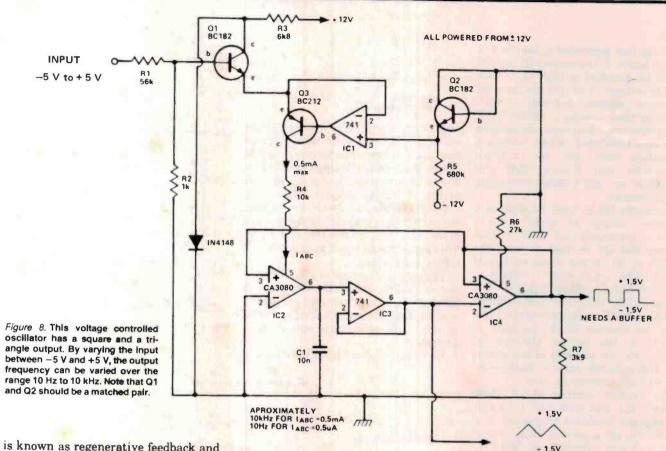
The result of varying RV1 is illustrated in Figure 4 and the transfer function of the circuit is shown in Figure 5.

Schmitt trigger

Most Schmitt trigger circuits prove to be very complicated when it comes to calculating the hysteresis levels. However, by using the CA3080 these calculations are rendered trivial, plus there is the added bonus of fast operation. The hysteresis levels are calculated from the simple equation,

 $V_{HYST} = +/-(I_{ABC} \times R2)$ The output squarewave level is in fact equal in magnitude to the hysteresis levels. The circuit operation is as follows (referring to Figure 7):

Imagine the output voltage is high. The output voltage will then be equal to $(R2 \times I_{ABC})$ which we will call + VHYST. If VIN becomes more positive than + VHYST, the output will start to move in a negative direction, which will increase the voltage between the input terminals which will further accelerate the speed of the output movement. This



is known as regenerative feedback and is responsible for the Schmitt trigger action. The output snaps into a negative state at a voltage equal to $-(R2 \times I_{ABC})$ which is designated as $-V_{HYST}$. Only when V_{IN} becomes more negative than $-V_{HYST}$ will the output change back to the +V_{HYST} state.

The Schmitt trigger is a very useful building block for detecting two discrete voltage levels and finds many uses in circuit designs.

Voltage controlled oscillator

By using two CA3080s and some 741 op-amps it is possible to make an oscillator, the frequency of which is voltage controllable. This unit finds many applications in the fields of electronic music production and test equipment.

The circuit (Figure 8) has been given a logarithmic control law, that is, the frequency of operation doubles for every volt increase in the control voltage. This makes it ideal for musical applications where linear control voltages need to be converted into musical intervals (which are logarithmically spaced) and also for audio testing where frequencies are generally measured as logarithmic functions. One CA3080, IC2, is an integrator. The I_{ABC} current that drives this IC is used to either charge or discharge C1. This produces triangular waveforms which are buffered by IC3, which then drives the Schmitt trigger IC4. The hysteresis levels for this device are fixed at +/-1.5 V, being determined by R6 and R7.

The output of the Schmitt trigger is fed back in such a way as to control the direction of motion of the integrator's output. If the Schmitt output is high, then the integrator will ramp upwards and vice versa.

Imagine that the integrator is ramping upwards. When the integrator's output reaches the upper hysteresis level, the Schmitt will flip into its low state, and the integrator will start to ramp downwards. When it reaches the low hysteresis level the Schmitt will flip back into its high state. Thus the integrator ramps up and down in between the two hysteresis levels.

The speed at which it does this, and hence the oscillating frequency, is determined by the value of I_{ABC} for IC2. The larger the current, the faster the capacitor is charged and discharged.

Two outputs are produced, a triangle wave (buffered) from IC3 and a squarewave (unbuffered) from IC4. If the squarewave output is loaded, then the oscillation frequency will change so a buffer is advisable.

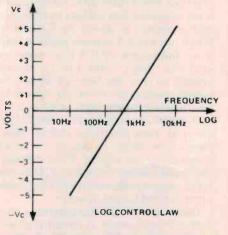


Figure 9. Voltage versus frequency characteristic of the Figure 9 circult.

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The log. law generator is composed of Q1, 2, 3 and IC1. Transistors Q1 and Q2 should be matched so that their base emitter voltages (Vbe) are the same for the same emitter current, (50uA). Matching these devices to within 5 mV is satisfactory, although unmatched pairs could be used. When matching transistors, take care not to touch them with your fingers. This will heat them up and produce erroneous measurements.

Transistor Q2 is used to produce a reference voltage of about -0.6 V, which is connected to IC1 pin 3. This op-amp and Q3 is used to keep the emitter of Q1 at the same voltage of -0.6 V. The input control voltage is attenuated by R1, R2 such that a + 1 Vincrease at the input produces a change of only +18 mV at the base of Q1. However, the emitter of Q1 is fixed at -0.6 V, so the current through Q1 doubles. (It is a property of transistors that the collector current doubles for every 18 mV increase in Vbe).

The emitter current of Q1 flows through Q3 and into IC2, thus controlling the oscillator frequency. It is possible to get a control range of over 1000 to 1 using this circuit. With the values shown, operation from 10 Hz to 10 kHz is achieved. Reducing C1 to 1n will increase the maximum frequency to 100 kHz, although the waveform quality may be somewhat degraded.

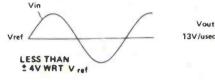
Changing C1 to 1uF. (non-polarised) will give a minimum frequency of 0.1 Hz.

Fast comparator

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The high slew rate of the CA3080 makes it an excellent fast voltage comparator and a circuit is shown in Figure 10. When pin 2 of IC1 is more positive than Vref, the output of IC1 goes negative and vice versa. Vref can be moved around so that the point at which the output changes can be varied. As long as the input sinewave level is quite large (1 V say) then the output can be made to move at very fast rates indeed. However, care must be taken to avoid overloading the inputs. If the differential input voltage exceeds 5 V, then the input stage breaks down and may cause an undesired output to occur.

One use of a fast comparator is in a tone burst generator. A circuit is shown in Figure 11. This device produces bursts of sinewaves, the burst starting



and finishing on axis crossings of the sinusoid. The CA3080 is configured here as a voltage comparator, used to detect these axis crossings and to produce a square wave output which then drives a binary divider (IC3). The divider produces a 'divide by sixteen' output which is high for eight sinewave cycles and then low for the next eight. This signal is then used to gate ON and OFF the sinewave.

The gate mechanism is a pair of transistors which short the sinewave to ground when the divider output is high and let it pass when the divider output is low. The resulting output is a toneburst.

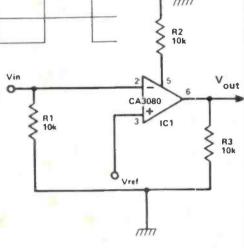
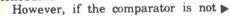
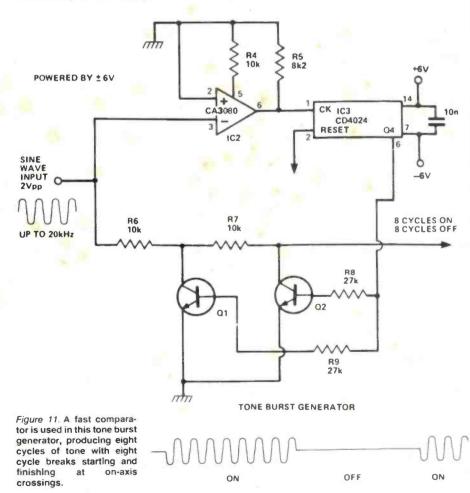
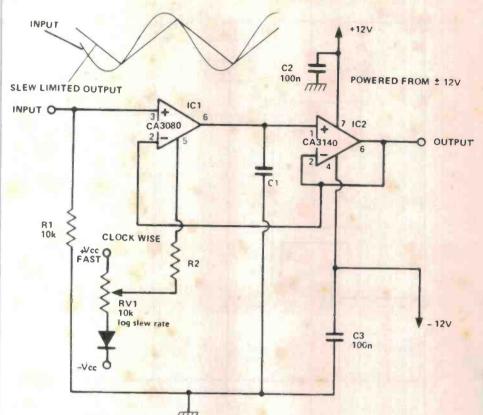
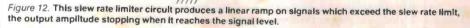


Figure 10. Example of a fast comparator.









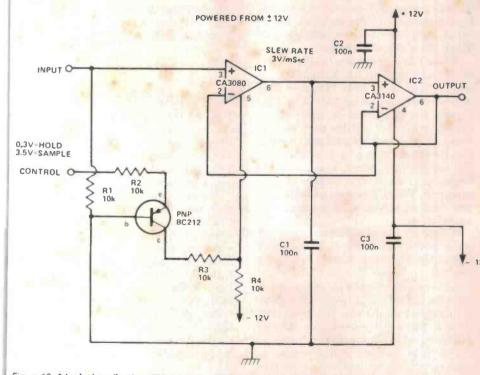


Figure 13. A typical application of the slew rate limiter is this sample and hold circuit.

very fast then there will be a delay in generating the gate and so the tone burst will not start or finish on axis crossings.

Using the circuit shown, operation up to 20 kHz is obtainable.

Slew limiter

The current output of a CA3080 can be used to produce a controlled slew limiter. By connecting the output current to a capacitor, the output voltage-cannot move faster than a rate given by

Slew Rate = I_{ABC} Volts per sec.

Note that I_{ABC} determines the slew rate and as I_{ABC} is a variable then so is the slew rate.

A suitable circuit is shown in Figure 12. The output voltage is buffered by a voltage follower, IC2. This is a MOSFET op-amp which has a very high input impedance, which is necessary to minimise the loading on C1.

When an input signal is applied to IC1 the output tries to move towards this voltage but its speed is limited by the slew rate. Thus, the output produces a linear ramp which stops when it reaches the input signal level.

Sample and hold

A typical application of the slew limiter circuit is in a sample and hold circuit. The circuit in Figure 13 could be termed an analogue memory. When the control voltage is high, the circuit will 'remember' or 'hold' the input voltage level present at the time. The result is shown in Figure 14.

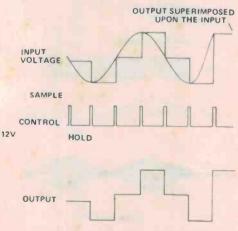


Figure 14. Illustrating the operation of the sample and hold circuit of Figure 13.

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In this circuit, I_{ABC} is either hard ON (sample) or completely OFF (hold). In the sample mode, the output voltage quickly adjusts itself so that it equals the input voltage. This enables a short sample period to be used.

In the HOLD mode, I_{ABC} is zero and so the voltage on C1 should remain fixed.

Such circuits are used in music synthesizers (to remember the pitch), in analogue-to-digital converters and many other applications.

A multiplier/modulator

The CA3080 is basically a twoquadrant multiplier, that is, it has two inputs, one of which can accept bipolar signals (positive and negative going) the inverting or the non-inverting input — the other can only accept a unipolar signal — the control input, pin 5.

Whilst a two-quadrant multiplier is very useful in a wide variety of applications, a four-quadrant multiplier has extra advantages. For example, apart from amplitude modulation, it can perform frequency doubling and ring modulation. See Figure 16. Now, a fourquadrant multiplier has two inputs, both of which can accept bipolar signals. An example of a four-quadrant multiplier is a frequency converter in a radio receiver. The familiar diode ring mixer is another example of a four-quadrant multiplier.

The circuit in Figure 15 is fairly similar to that of the two-quadrant multiplier shown in Figure 1. This circuit has several important differences.

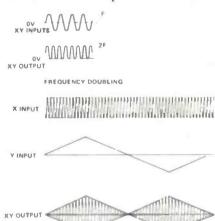






Figure 16. Illustrating the various operations of the four quadrant multiplier of Figure 15.

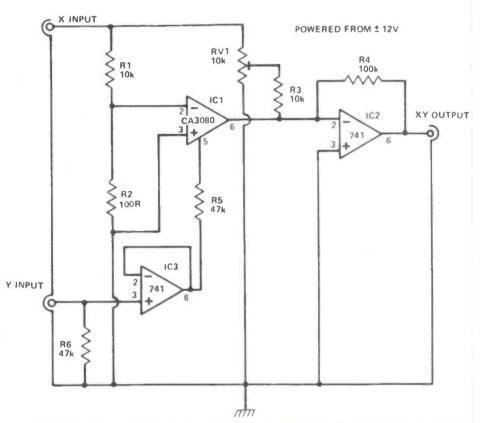


Figure 15. This multiplier/modulator can be used to produce a 'Dalek' voice when working as a ring modulator. It can also be used as a frequency doubler.

A 741 op-amp, IC3, is used to generate I_{ABC} in such a way that its input, the 'Y' input, can go both positive and negative. Thus, the Y input is bipolar.

When Y is at zero volts (no input) and there is a signal on the X input the desired output (X x Y) should be zero. This is achieved by adjusting RV1 so that the signal via IC1 (this is inverted) is exactly cancelled out by that via R3. Now, when Y is increased positively, a non-inverted value of X is produced at the output and, when Y is increased negatively, an inverted value of X is produced. When Y is zero, so is the output. This is known sometimes as ring modulation.

If a speech signal is connected to the X input and an audio oscillator to the Y input, the resulting sound is that of a 'Dalek'.

Also, if a sinewave is connected to both the X and Y inputs, the XY product is a sinewave of twice the frequency. This is known as a frequency doubler, but it will only work with sinewaves.

For more theoretical information on four-quadrant multipliers, especially the variable transconductance type, see "Operational Amplifiers" (second edition), by G.B. Clayton, published by Newnes-Butterworths.

Single pole filter/wah wah

The guitar 'wah wah' effects unit employs a filter which can be manually 'swept' across the middle of the audio frequency range, generally from around 500 Hz to 5 kHz or so, producing the peculiar 'wah wah' sound.

A single pole, voltage-controlled, low pass filter can be constructed using a CA3080 as a current-controlled resistor. The circuit is showh in Figure 17.

A simple, low pass RC filter configuration is employed, the controllable 'R' is the CA3080 and the 'C' is C1. Varying I_{ABC} varies the amount of current drive to C1. This circuit configuration would normally be a slew limiter, except that the signal level to the input of the CA3080 is kept deliberately low (R1 and R2 form a 100:1 attenuator) so that the IC operates in its linear mode. This enables it to look like a variable resistor.

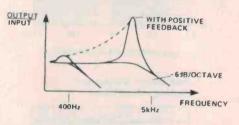
LINK 0+ Vcc + 12V RV1 10k 0 R3 lin MANUAL POWERED FROM ± 12V LABC D1 IN4148 6 3 + CA3080 2 ____ - 12V OUTPUT Vec R1 10k 741 0 C1 470p R2 100R Ş R4 100R R5 10k R6 10k POSITIVE FEEDBACK 0 n C2 10n LINK

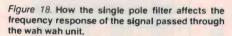
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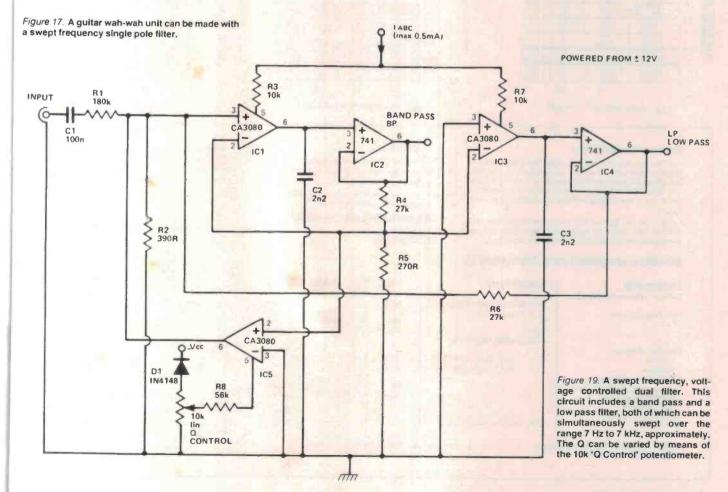
When this resistor is varied, the break frequency of the filter also varies.

By applying some positive feedback around the filter (R6, C2) it is possible to produce a peaky filter response. The peak actually increases with frequency, producing the wah wah effect.

The circuit as shown can be swept from about 400 Hz at the lower extreme to about 4 kHz at the upper extreme. See Figure 18.

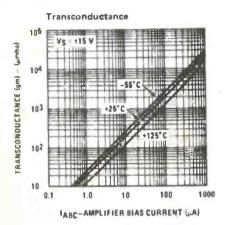




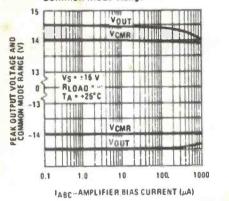


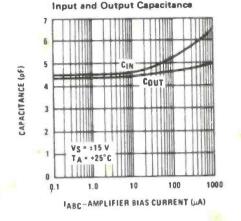
OP-AMP COOKBOOK

SELECTED DATA ON THE 3080

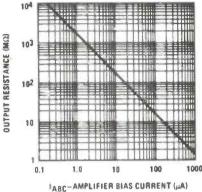


Peak Output Voltage and Common Mode Range

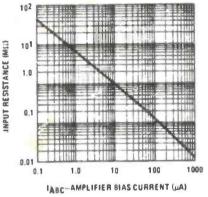








Input Resistance



· Slew rate (unity gain compensated): 50 V/us

Flexible supply voltage range: +/-2 V to +/-18 V

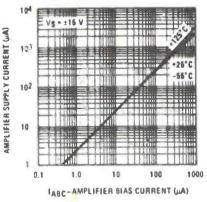
+/-18 V

. Fully adjustable gain: 0 to gm RL limit

Absolute maximum ratings

Adjustable power consumption





General description

The 3080 is a programmable transconductance block intended to fulfil a wide variety of variable gain applications. The 3080 has differential inputs and high impedance push-pull outputs. The device has high input impedance and its transconductance (gm) is directly proportional to the amplifier blas current (I_ABC).

High slew rate together with programmable gain make the 3080 an ideal choice for variable gain applications such as sample and hold, multiplexing, filtering, and multiplying.

Electrical characteristics, 3080 (Note 1).

Conditions				
		9600	13000	umho
Over Specified Temp. Range	5400			umho
$R_{I} = 0$, $I_{ABC} = 5uA$		5		uA
	350	500	650	uA
	300			uA
Over Specified Temp.Range				
RL = ,5uA IABC 500uA	+12	+14.2		V
	-12	-14.4		V
E ABO		1,1		mA
1	80	110		dB
	+/-12	+/-14		V
				k
				MHz
Linity Onin Componented		_		V/us
	Over Specified Temp. Range $R_L = 0$, $I_{ABC} = 5uA$ $R_L = 0$ $R_L = 0$ Over Specified Temp.Range $R_L = , 5uA$ I_{ABC} 500uA	$ \begin{array}{c c} \text{Over Specified Temp, Range} \\ \text{R}_L = 0, \text{I}_{ABC} = 5\text{uA} \\ \text{R}_L = 0 \\ \text{R}_L = 0 \\ \text{Over Specified Temp, Range} \\ \text{R}_L = , 5\text{uA} \text{I}_{ABC} \text{S00uA} \\ \text{R}_L = , 10 \\ \text{R}_L =$	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	$ \begin{array}{c} \text{Over Specified Temp, Range} \\ \text{R}_L = 0, \text{I}_{ABC} = 5\text{UA} \\ \text{R}_L = 0 \\ \text{R}_L = 0 \\ \text{Over Specified Temp, Range} \\ \text{R}_L = 0, \text{I}_{ABC} = 5\text{OUA} \\ \text{I}_{ABC} = 5O$

_

Features

· Extended gm linearity

Supply Voltage 3080

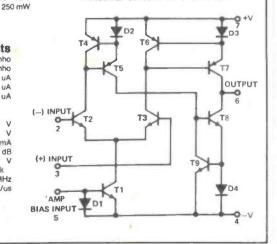
Power Dissipation

Differential Input Voltage Amplifier Bias Current (I_{ABC}) DC Input Voltage

Output Short Circuit Duration

+/-5V 2mA +/V_S to -V_S Indefinite

Internal circuit of the 3080



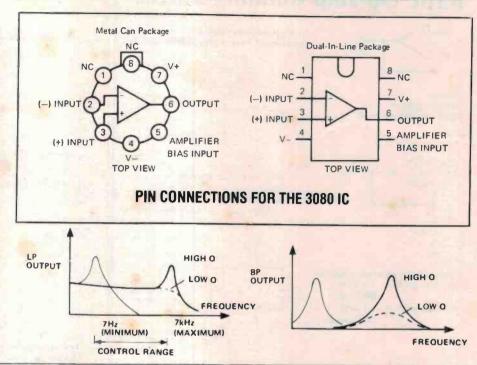
Note 1: These specifications apply for $V_S = +7 - 15$ V and $T_A = 25$ °C, amplifier bias current (I_{ABC}) = 500uA, unless otherwise specified.

Voltage controlled filter

A standard dual integrator filter can be constructed using a few CA3080s. By varying I_{ABC} , the resonant frequency can be swept over a 1000 to 1 range. IC1 and IC3 are two current-controlled integrators. IC2 and IC4 are voltage followers which serve to buffer the high impedance outputs of the integrators. A third CA3080 (C5) is used to control the Q factor of the filter. Q factors as high as 50 can be obtained. The resonant frequency of the filter is linearly proportional to I_{ABC} and hence this unit is very useful in electronic music production.

There are two outputs, a low pass and a band pass response. Minimum frequency is around 7 Hz to 10 Hz, upper frequency is around 7 kHz or so. Changing C2 and C3 will alter the upper and lower frequency limits.

Figure 20. Illustrating the operation of the filters in the Figure 19 circuit.



NOT 20 PROJECTS NOT 25 PROJECTS NOT EVEN 35 PROJECTS ... BUT ... 50 PROJECTS! ... WOW!

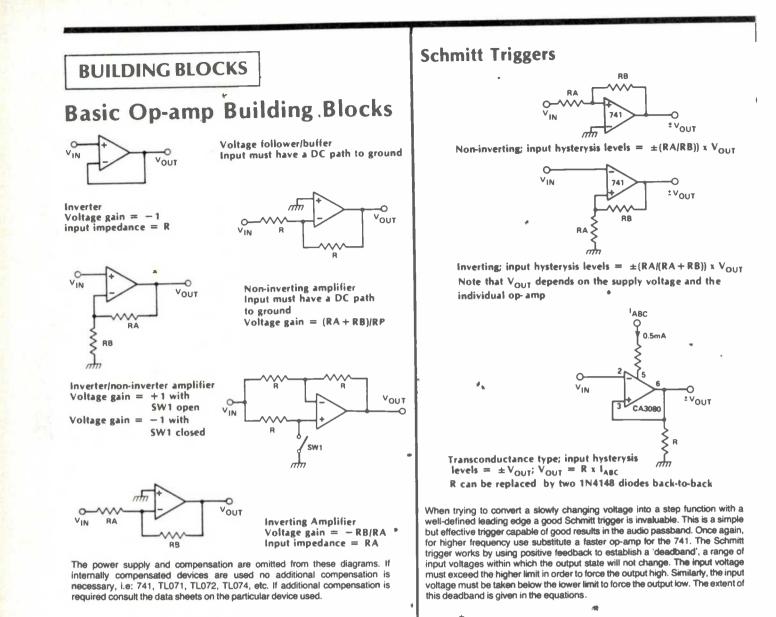


The great, grand, gl-normous HOBBY ELECTRONICS PROJECT BOOK

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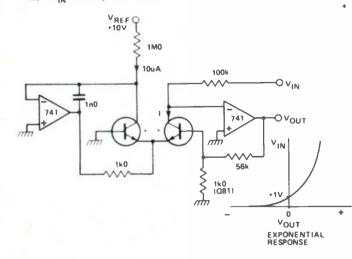
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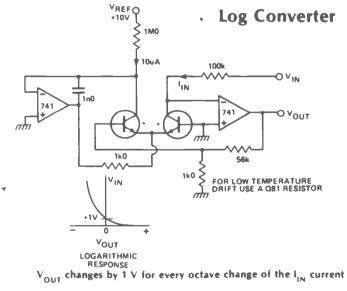


Antilog (Exponential) Converter

 $V_{OUT} = 1 \times 100 k$

The current I doubles for every 1 V increase of V_{1N} When $V_{1N} = 0$ V, I = 10 uA





*The matched transistors can be two BC212L in thermal contact, or a dual transistor (LM394), or pat of an array (CA3046)

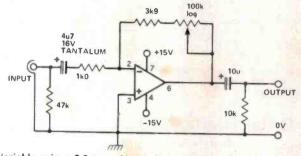
AUDIO

OP-AMP COOKBOOK

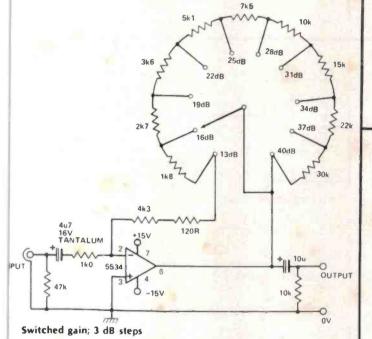
Low Impedance Source Preamp

Very low input noise Input noise = $4 \text{ nV} \sqrt{\text{Hz}}$

Equivalent input noise voltage = 0.56 uV_{RMS} (20 kHz bandwidth) Input impedance = 1k0 (suitable for microphone)

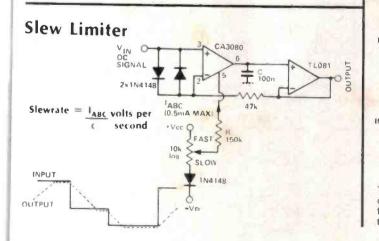


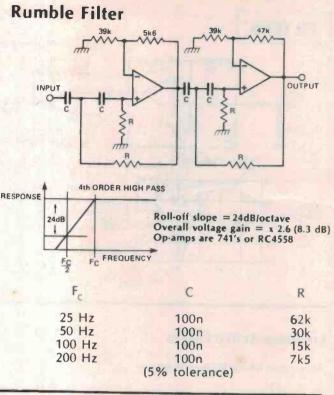
Variable gain; x 3.9 to x 100 (12 dB to 40 dB)



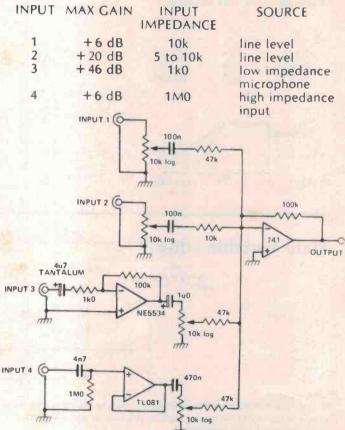
The NE5534N is a very low-noise op-amp specifically intended for audio appli-cations. The device boasts one of the lowest noise figures of all op-amps combined with good slew rate and large signal bandwidth figures. The lowest-noise devices have the designation NE5534AN. Suitable supply

decoupling is essential if best results are to be obtained.

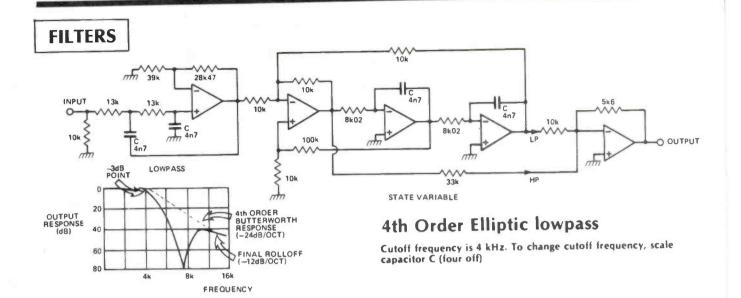




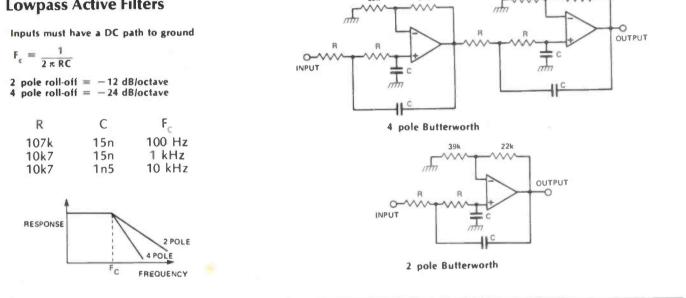
Simple Mixer



This simple mixer has been provided with four different types of input circuit. Any combination of these could however be used. Once again, the 741 limits the high frequency response and slew rate capabilities. To improve performance substitute the 741 for a faster device such as an NE5534N or TL071, etc.





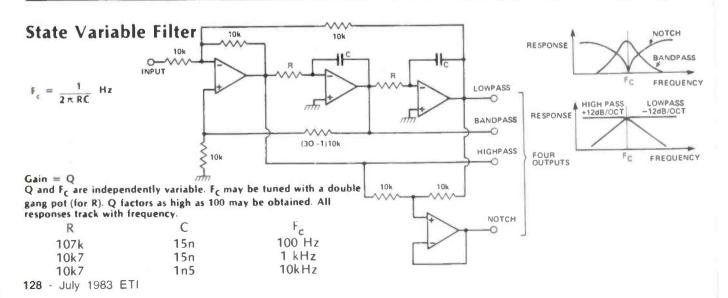


47k

39k

5k6

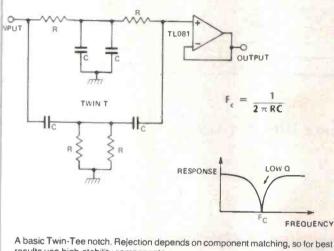
39



OP-AMP COOKBOOK

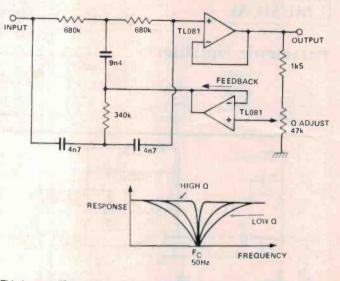
Active Notch Filter

The two R's in parallel represent R/2 The two C's in parallel represent 2C For 50 Hz, R = 680k, C = 4n7 (a hum remover)



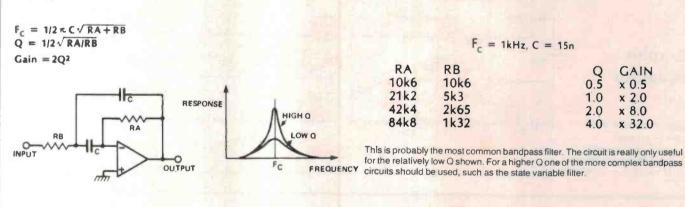
results use high-stability components.

50 Hz Notch, Variable Q



This is a modified version of the basic Twin-Tee notch filter. The Q can be adjusted by controlling the amount of feedback with the 47k potentiometer. The rejection offered by the circult is determined by the matching of the passive components, but even with ordinary components a figure of 30 dB to 40 dB should be obtained.

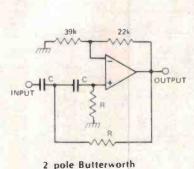
Bandpass Active Filter

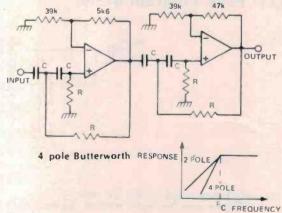


Highpass Active Filters

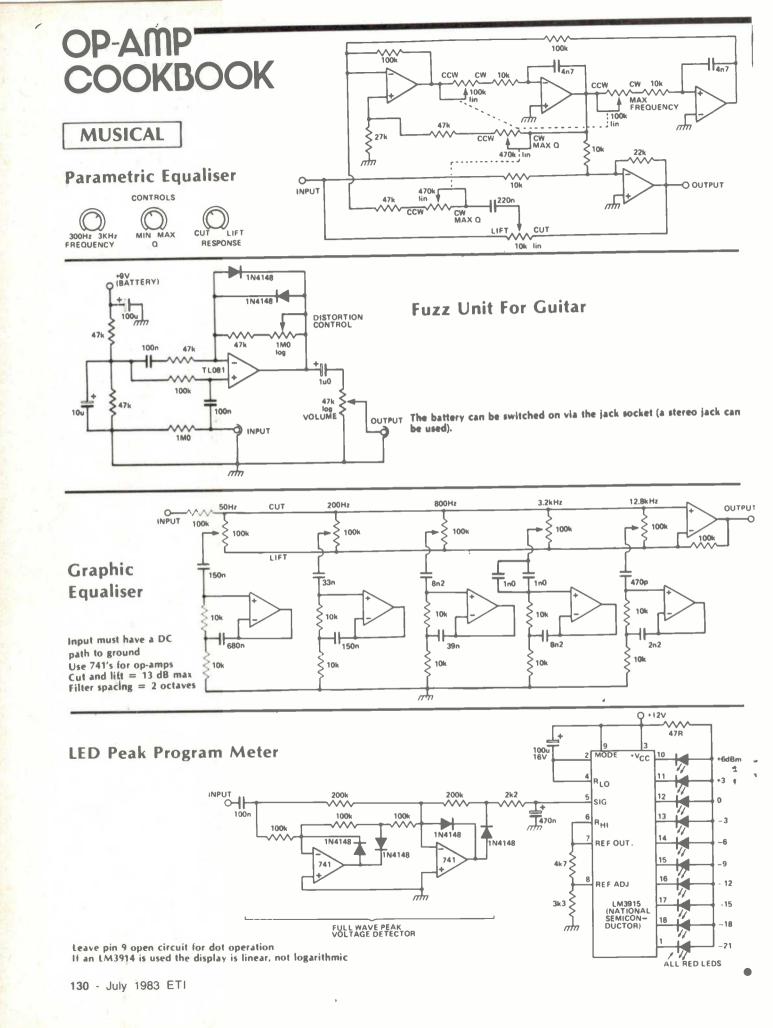
$F_c = \frac{1}{2\pi RC} Hz$	
2 pole roll-off =	+ 12 dB/octave
4 pole roll-off =	+ 24 dB/octave

107k 15 10k7 15 10k7 1r	in <mark>1 kHz</mark>





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- function. Auto-Fix and Hold-Off
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- function. •Triggering waveform on CH 3.
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Why some CMOS circuits don't work as you expect

Stephen Dolding

A 4093 is a 4093, right? Well . . . yes, and no. There are quite a few pitfalls in the CMOS 'jungle' and it's handy to know about them before venturing forth.

CMOS '4000' SERIES integrated circuits are manufactured by at least six major manufacturers and the 74C series by at least two major manufacturers, but it must not be assumed that a 4XXX from one manufacturer is interchangeable with a 4XXX device from another manufacturer. This article explains some of these differences.

Schmitt gate oscillator

What could be simpler than the oscillator circuits shown in Figures 1(a) and 1(b)? There are so few components that you would expect these circuits to work first time.

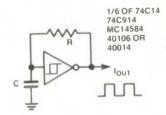


Figure 1a. Different manufacturers ICs will produce different results

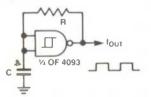


Figure 1b. Such a simple circuit but the value of Iout can vary from 52.9 Hz to 249 Hz, depending on the brand of 4093 used

You have selected stable, close tolerance components and calculated the frequency according to the formula:

$$f_{out} = \frac{1}{CRlog_{e} \left[\left(\frac{V_{cc} - V_{t-}}{V_{cc} - V_{t+}} \right) \left(\frac{V_{t+}}{V_{t-}} \right) \right]}$$

where C = capacitor value in uF

- R = resistor value in kilohms
- $f_{out} = frequency out in kHz$
- $V_{cc} = supply voltage$
- $V_{t+} = upper trigger level of Schmitt$ trigger
- $V_{t_{2}} =$ lower trigger level of Schmitt trigger

Now, the question is, "what are the values of V_{t+} and V_{t-} ?" We need to refer to the manufacturer's data sheet for an answer. But which manufacturer? There are at least six different manufacturers to choose from and each one gives a different range of possible values for V₁₊ and V₁.

Considering the 4093 IC (Figure 1(b)); if all the databooks are consulted it is found that for $V_{ee} = +5 V$ the *highest* typical value of $V_{t+} = 3.6 \text{ V}$ (extreme = 4.3 V).

The *lowest* typical value of $V_{t+} = 2.7 \text{ V}$ (extreme = 1.7 V).

The *highest* typical value of $V_{t_{r}} = 2.2$ V

(extreme = 3.3 V).The *lowest* typical value of $V_{t_{2}} = 1.4$ V (ex

treme = 0.7 V. Now if V_{t} = 3.6 V. V_{t} = 1.4 V. V_{cc} = +5 V. C = 100 nF and R = 10k. then from the formula above $t_{\rm out} = 52.9$ Hz.

With the same type from another manufacturer, V_{t+} could be 2.7 V and V_{t-} could be 2.2 V, in which case recalculation gives a frequency of 249 Hz!

These figures are based on typical values of trigger level. Extremes of high and low trigger levels could give frequencies ranging from 27.5 Hz to 961 Hz with the same values of C = 100 nF and R = 10k. This gives a frequency range of almost 35:1 if the whole spectrum of possibilities is considered.

Now it will be clear why the circuit may oscillate at a frequency which is considerably different from what was expected or intended by the designer who only consulted one manufacturer's databook!

The monostable

Now let us look at another CMOS circuit often used by the hobbyist - the 4528 dual monostable, shown in Figure 2.

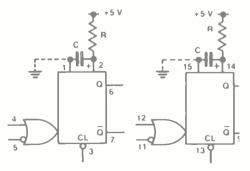


Figure 2. The 4528 dual monostable can have variations in its performance of up to ±50%

Lab Notes

Here again, different manufacturers give different formulae for the monostable time constant. A typical formula (for +5 V supply) is

t = 0.37CR

However, it could vary from t=0.32CR to t=0.42CR typical, with variations up to $\pm 50\%$.

The pulse width depends very much on the supply voltage. Some manufacturers' 4528s give increasing pulse width with increasing supply voltage, others give a reverse effect.

The formula given above depends on the value of C being greater than 10n. For smaller values of capacitance the manufacturers data sheets need to be consulted.

It should also be noted that some manufacturers require pins 1 and 15 to be grounded externally for correct operation. To overcome the variations in timing formula, a CMOS 4538 integrated circuit can be used in place of the 4528. The 4538 is pin compatible with the 4528 IC and the formula is:

$$\mathbf{t} = \mathbf{C}\mathbf{R}$$

In all timing circuits using CMOS ICs it is wise to make provision for trimming the value of the timing resistor to allow for adjustment.

The counter/divider

Next we come to a well-known decade counter/divider IC — the 4017.

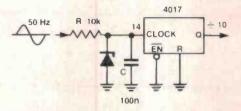


Figure 3. This circuit will only work correctly with particular brands of 4017 decade counter/dividers. The wrong choice can result in false counting.

Will the circuit in Figure 3 always work correctly? No, only with a Motorola 14017 or an RCA 4017, because these have an internal Schmitt trigger on the clock input. Other manufacturers' 4017s do not, so false counting may result.

BCD decoder

Another fairly common integrated circuit likely to give problems is the 4028 BCD-to-10-line decoder. We now come to see that actual logical differences can occur between one manufacturer's 4XXX and another manufacturer's 4XXX.

With the 4028, some manufacturers (Motorola and RCA) do not decode the six 'illegal' binary codes 1010 to 1111 (i.e. 10-15), while other manufacturers (including National, Fairchild and Philips) decode these outputs as if the input was 8 (1000) or 9 (1001). The problem of logical differences between one manufacturer's device and another manufacturer's device (with supposedly the same type number) applies also to the 4585 fourbit comparator and even to the ubiquitous 555 and 556 timers. There may be other examples too. The problem fortunately does not occur with the range of quad gates.

The moral

The above-mentioned examples were all encountered during the design of one piece of industrial equipment which made use of these common CMOS parts.

You may well ask "If design engineers, who have ready access to all the data books, can run into such problems, what about the unsuspecting hobbyist, who has no data?"

The moral of this article is that "forewarned is forearmed". It is hoped that this article may at least prevent some construction projects from being abandoned because they do not appear to work correctly at first sight. Designers who publish projects should check that there are at least two manufacturers' ICs which will work in the circuit as intended and, if necessary, spell out the names of suitable manufacturers in the parts list. Best of all, only design circuits that will work with all manufacturers devices of the same basic type number (though this may not always be possible).

If problems occur, all that may be required is to try an IC from a different manufacturer.

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IDEAS FOR EXPERIMENTERS

These pages are intended primarily as a source of ideas. As far as reasonably possible all material has been checked for feasibility, component availability etc, but the circuits have not necessarily been built and tested in our laboratory. Because of the nature of the information in this section we cannot enter into any correspondence about any of the circuits, nor can we produce constructional details.

Monster flasher

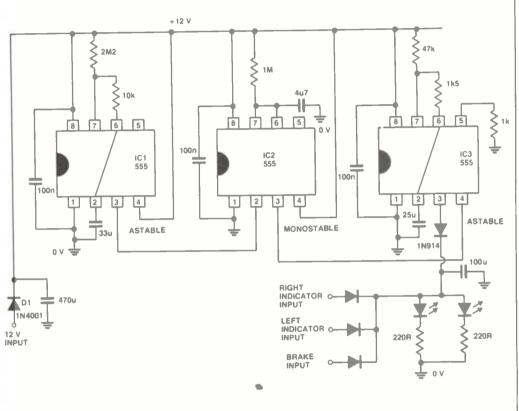
Andrew Stewart of Gumdale Queensland has a little monster which sits on the dashboard of his car and flashes its eyes (LEDs) with the indicator and brake lights. Now many of his friends want a 'flashing eyed' monster. He sent us a copy of his most advanced unit.

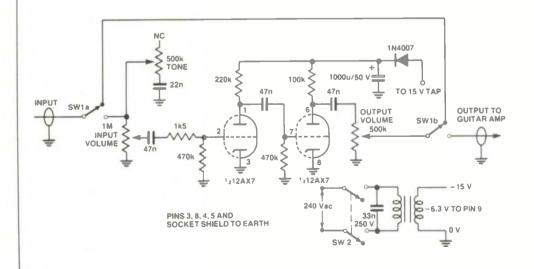
As the ignition system in a car is often noisy D1 is needed to isolate the 470u capacitor from the 12 V supply.

IC1 555 gives a time delay of over a minute and when it finishes counting it sends a pulse to IC2 555. The output of IC2 then goes high for approximately 4-5 seconds. This, in turn, causes IC3 to oscillate and the monster's eyes blink.

The 1k resistor brings the duty cycle close to 50%. The 100 uF capacitor on the output of 1C3 makes the eyes look as if they are fading in and out.

The monster I used for mounting the LEDs in was a cheap plastic lizard. The ICs can be mounted on Vero board and placed inside a matchbox up under the dashboard.





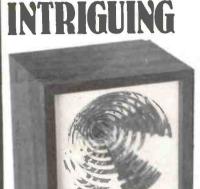
Valve distortion unit

If you really want a valve distortion unit, Jeff Parkes of Collingwood Victoria has designed one.

I built this effect pedal in an aluminium box with dimensions of 130 mm x 75 mm x 55 mm. It is almost hum and noise free and sounds great with a gutsy guitar such as the 'Les Paul' type.

I have left it running for several days and the performance has remained constant. The transformer is a 2155 and SW1 is an audio quality DPDT footswitch.





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And what do you pay for this LCD display? ONLY \$2.95 each or \$2.50 each 10 up. Staggering value. Cat ZM9015 Pack of 50 Molex pins (only 42 required) Cat PI6540 Only \$1.00 Cat A2005

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2 amp - Secondary 2: 15VAC 2 amp - Secondary 3: 8VAC 8 amp.

A typical DC supply could be ±15V DC @ 1.5A & 5V DC @ 8 A or ±12V DC @ 2A & 5VDC @ & 5V DC @ 8 A or ±12V DC @ 2A & 5V CC 8A This transformer would normally sell for 84

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codemaster

95

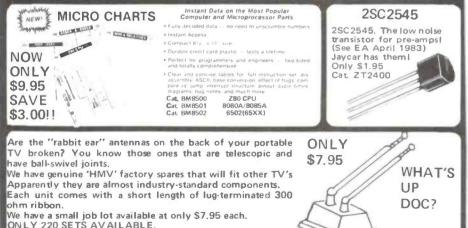
WAS \$12.50

Many of you know the clever parlour game that uses coloured tokens to stretch the brain to work out a hidden code in a minimum number of moves. The people that came up with the game

used a descriptive name which no-one else can use. It is a popular game and is well known under this name. Our game is similar to this game but - naturally is similar to this game but - haturally-its electronicil And, what's more, you can play against the machine - alone. Each XM7015 Codemaster measures 140(1)x85(w)x25(d) looks similar to a pocket calculator and runs off a standard 9V cell. Provision is made for a mains adaptor as well.

The Codemaster once sold for \$29,50 but Jaycar has made a huge scor purchase. You save a fortunel Grab one now for only \$12,501! \$9.95 has made a huge scoop

(For a further clue to the origin of this game read this page carefully)



\$7.95

IDEA OF THE MONTH

A IDEA OF THE MONTH A

12 V

BUZZER

TO + VE VIA DASHBOARD LIGHT DIMMER CONTROL TO CHASSIS VIA

Parking light reminder

Stephen Mann, Forrestfield WA

There are many different circuits around for devices to remind you to switch your parking lights off. However, this idea must surely be one of the simplest and the cheapest

The buzzer requires only two connections to the vehicle; one to a door courtesy light switch and the other to the positive battery terminal via the dashboard light mer control. dimmer control.

It's very simple to operate. The buzzer will sound when the parking lights are on and you open the car door. This warns you to turn the parking lights off.

An added feature of this circuit is that if you need to have the parking lights on and the door open, you can silence the buzzer by operating the dashboard dim-

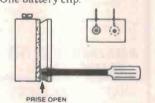
261 transistor battery clip

Darryl Green of Elanora Qld. is another enterprising experimenter

I was breadboarding a small circuit requiring an op amp and ±9 V rails and found that I only had one 216 9 V battery clip.

But I did have an old 216 battery and discovered that by prising the metal case open I could remove the top plate with the clips mounted on it.

I then soldered wires on to the back of the plate, taking care not to melt the plastic. Hey presto! One battery clip.



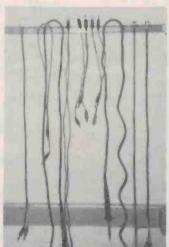
Simple holder for leads. probes etc

If you want to have a tide electronic workshop this low cost lead holder should help.

H.F. Nissink of Launceston Tasmania sent in the idea which makes it easy to store test leads. BNC to BNC connectors and test probes etc.

Slotted cable is readily available through electrical suppliers at a fairly cheap rate. This cable ducting, cut to the required length and bolted to the wall. works well as a lead storage rack.

The slot size is ideal for electronic type leads. As well, the cable ducting may be cut along its length through the centre of the slots with the base of the Ushape bolted onto the wall as shown.



Matchbox cabinet for small components

Chris Nixon of Bentley WA has together to form a miniature found a use for empty match- cabinet. On the opening end of boxes.

of storage for smaller components components contained in the e.g: resistors, capacitors, transistors etc. is to glue matchboxes

each matchbox write the values A simple but effective method and abbreviated names of the 'drawers'

PRIZE WORTH S90!

'IDEA OF THE MONTH' CONTEST

COUPON

Cut out and send to: Scope/ETI 'Idea of the Month' Contest, ETI Magazine, 140 Joynton Ave, Waterloo NSW 2017.

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Scope Laboratories, who manufacture and distribute soldering irons and accessory tools, have offered to sponsor a contest with a prize to be given away every month for the best item submitted for publication in the 'Ideas for Experimenters' column - one of the most consistently popular features in ETI. Each month we will be giving away a Scope Panavise Multi-purpose Work Centre, Model 376/300/312, comprising a self-centering head (376), standard base (300) and tray base mount (312), all worth about \$901 Selections will be made at the sole discretion of the editorial staff of ETI Magazine. Apart from the prize, each winner will be paid \$10 for the item published. You must submit original ideas of circuits which have not previously been published. You may send as many entries as you wish

RULES

This contest is open to all persons normally resident i Australia with the exception of members of the staff of Scope Laboratories, Federal Publishing Company Proprietry Limited, ESN, The Litho Centre and/or associated companies. Closing date for each issue is the last day of the month

Entries received within seven days of that date will be accepted If postmarked prior to and including the date of the last day of the month.

The winning entry will be judged by the Editor of ETI, whose decision will be final. No correspondence can be entered into regarding the decision.

Winner will be advised by telegram the same day the result is declared. The name of the winner, together with the winning idea, will be published in the next possible issue of ETI.

Contestants must enter their names and address where indicated on each entry form. Photostats or clearly written copies will be accepted but if sending copies you must cut out and include with each entry the month and page number from the bottom of the page of the contest. In other words you can send in multiple entries but you will need extra copies of the magazine so that you send an original page number with each entry.

This contest is invalid in states where local laws prohibit entries

Entrants must sign the declaration on the coupon that they have read the above rules and agree to abide by their conditions





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open for trading during the normal trading hours for their particular area (either 9-5.30 or 8.30-5). Many stores are also open for late night trading. Please ring the store concerned for their particular hours.



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Dear Customers

Quite often, the products we advertise are so popular they run out within a few days. Or unforseen circumstances might hold up shipments so that advertised lines are not in the stores by the time the advert appears. And very occasionally, an error might slip through our checks and appear in the advert after all, we're human tool). Please don't blame the store manager of staff; they cannot solve a dock strike on the other side of the world, or fix an error that's appeared in print. If you're about to drive across town to pick up an advertised line, why not play it safe and give the store a call first just in case Thanks.

Dick Smith and Staff

Communications NEWS

Marconi Radio Society re-launches historic callsign

The callsign used to introduce Britain's first scheduled radio entertainment broadcast, 2MT, will be heard on the amateur bands later this year after a 60-year break in transmission.

British Home Office approval has been granted for the use of the callsign G2MT by the Marconi Radio Society, a group formed recently by amateur radio enthusiasts employed at the Stanmore (UK) headquarters of Marconi Space & Defence Systems Ltd and at other company sites in the locality.

The callsign will be used at Stanmore for the first time at 1200 hours BST on Saturday 2nd July 1983 using equipment owned and operated by members.

The frequencies used will depend on the prevailing propagation conditions but it is hoped to organise contacts with amateur radio clubs affiliated to the BBC and with similar groups within the GEC-Marconi organisation.

The founding of the Marconi Radio Society and its authorisation to use the historic callsign G2MT during 'World Communi-

The 11-channel memory can

The JST-100 is fully equipped with capabilities such as three

easily store and recall not only

VFO frequencies but also work-

PLL circuits phase-locked with a highly stable 10 MHz standard

crystal oscillator; two digital

VFOs permitting independent

selection of bands and modes.

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thesiser; digital two-VFO system; 11-channel memory;

memory finder capability; easy

tuning with main dial and up-

down buttons; easy-to-operate

panel design; receiver input

the working VFO being held. Features available include:

ing frequency bands and modes.

access.

Synthesised HF transceiver from JRC

The compact JST-100 HF transceiver employs an 8-bit

microprocessor to control all the complicated functions

such as band and mode selection and memory channel

cations Year' has been given wholehearted support by Marconi Space & Defence Systems Limited.

As well as enjoying the use of the company's facilities at Stanmore, the Society's Patron is General Sir Harry Tuzo and its President Dr. W. Bardo, Chairman and Technical Director respectively of MSDS.

Marconi's Wireless Telegraph Company Limited was first granted an experimental licence in the summer of 1920 to use the callsign to introduce news bulletins.

This permission did not apparently extend to music, even in the accompanying role, and the licence was swiftly revoked following the broadcast of 'dramma per musica' by a Danish tenor.

Following representations to the then Postmaster General by the Wireless Society of London

(now the Radio Society of Great

Britain), the company was later authorised to recommence transmissions.

And so, the first scheduled entertainment broadcast in the UK was transmitted on 700 metres from Writtle, near Chelmsford, on 14th February 1922 under the callsign 2MT ('Two-Emma-Tock'). The 'G' (for England) has now been added to accord with current practice.

The licence restricted broadcasting to half an hour each Tuesday evening and the station was required to cease transmitting for three minutes in every ten. These frequent intervals were spent in checking to hear whether any complaints had been received.

The broadcasts provided amateurs with invaluable checking references and their content set the pattern for later public broadcasting programmes; for example, the first radio play was produced (Cyrano de Bergerac) and a rudimentary 'Children's Hour' was evolved before transmissions ceased in January 1923.



Radio amateur canonised

The first licensed radio amateur to become a saint is Father Maximilian Kolbe, SP3RN, who died at the infamous Auschwitz death camp on 14 August 1941.

Fr Kolbe spent several years as a missionary before starting his amateur radio activities in 1938. He was arrested by the Gestapo in 1939 and sent to Auschwitz, near his home parish of Krakow.

In 1941, when one of his fellow prisoners, who was the head of a large family, was selected for death, Fr Kolbe volunteered to go in his place. He was executed on 14 August that year.

Fr Kolbe was posthumously awarded the highest Polish military medal, the Virtuti Militarta Golden Cross. Pope John Paul canonised Fr Kolbe as a saint on 10 October 1982, thus making him the first licensed radio amateur to become a saint. (Inx Radio Comm.)

Delsound get A.E.A.

Melbourne-based Antenna Engineering Australia Pty Ltd has announced the appointment of Delsound Pty Ltd as distributor of A.E.A. products in Queensland.

Delsound will hold stocks of A.E.A. antennas and clamps, etc. Contact at **Delsound** is Bud Pouncett. Phone Brisbane **229-6155**, or call at 1 Wickham Terrace, Brisbane.



circuit of narrow BPF type; large multi-function display; remote control of frequencies and modes; overmodulation indication; unique ALC input; frequency data output.

Options include a power supply, antenna tuner, speaker, desk microphone, hand microphone and key.

Further details about this exciting rig are available from Vicom, 57 City Rd, South Melbourne Vic. 3205. Brances in Sydney, NSW and Lower Hutt. N.Z.

Latest Bearcat slashes the price of scanning

The Bearcat 200FB offers a great deal to the fledgling scanning enthusiast for \$349. Here's a guick review of this recently released rig.



Available through Dick Smith stores, this latest Bearcat features a 16-channel memory (with battery backup), auto or manual search over three bands (66-88 MHz, 138-174 MHz, 407-512 MHz), direct channel access, automatic channel lockout and priority functions plus patented selective scan delay so you don't miss the reply on two-way conversations.

The rig can be operated direct from the vehicle battery, or from the mains using a commonly available plugpack. It comes complete with its own telescopic antenna, which suits desk-top operation of the unit, and an external antenna socket so you can really 'pull in' those signals from an outside antenna.

We jumped at the opportunity to review one of these rigs and here's what we found.

The Bearcat 200FB is a compact plastic-cased rig, measuring just 235 mm wide by 220 mm deep by 70 mm high. A 'porch' at the front contains the controls and a sloping panel from the top to the porch contains the 9-digit fluorescent display.

The volume and squelch knobs at the left are well-designed, just right for thumb-and-forefinger operation. The programming keyboard occupies most of the right hand side of the porch. It employs sensor-touch operation and the loudspeaker 'bips' when you press the keys.

Internally, it appears a wellconstructed unit, all the electronics being on a single printed circuit board.

On the air, the unit proved to be very sensitive, easily pulling in signals from base stations 50 km away on just its own telescopic whip. Programming it is a breeze.

The handbook that comes with it is clearly written, well set out and easy to use.

The display is a little hard to see with light directly falling on it, but if you locate the unit away from direct light, there's no problem. In any case, the same goes for many other scanners we've seen.

The handbook warns of 'birdies' — spurious signals, generated within the receiver, and conveniently lists them, thus enabling you to avoid them. We had a look at them and looked up our frequency table and think they should not be troublesome.

Basically, the Bearcat 200FB is a delight to use, offers many useful features and is worthy of serious consideration if you're looking for a scanner and don't want to pay big bucks'.

JIL SX-200 A BETTER SCANNING MONITOR RECEIVER. COVERS 26-88 MHz & 108-180 MHz & 380-514 MHz





GFS Electronic Imports 15 McKeon Road, Mitcham, 3132, Vic. Telex 38053 GFS Phone: (03) 873 3939

Monitors over 33,000 frequencies from 26 to 88 MHz, 108 to 180 MHz and 380 to 514 MHz. Bands included within this range are HF and UHF CB, 27 and 155 MHz MARINE, Australian LOW BAND, AIRCRAFT band, VHF SATELLITE band, 10 Mx, 6 Mx, 2 Mx and 70CMx AMATEUR BANDS, VHF High BAND as well as UHF two-way band.

Mechanically rugged the SX-200 uses high quality double-side Epoxy-Glass printed circuit boards throughout. Some of its other outstanding features include 3 MODE SQUELCH circuitry which allows the lockout of spurious and carrier only signals, extremely low spurious count, AM and FM detection on all bands, FINE TUNING control for off channel stations, 240 VAC or 12 Volt DC operation, Accurate QUARTZ CLOCK, Squelch operated OUTPUT for switching a tape recorder etc, 16 Memory channels, MEMORY BACKUP, which lasts up to two years, high SENSITIVITY and SIGNAL-TO-NOISE ratio on all bands, CRYSTAL FILTER for excellent SELECTIVITY and easy servicability due to component layout as well as a 90 day warranty.

Its high quality and performance is testified by the fact that it is in use by a large number of State government and Federal bodies including most state and federal police departments. Contact GFS, the Australian Distributors, or our interstate outlets for full technical specifications. We also market a range of pocket scanning receivers and transceivers. Contact us for full details.

PRICE \$599 INC. S.T. + \$12 P&P; SERVICE MANUAL \$12 + \$1.50 P&P; SCAN-X BASE ANTENNA \$62 + \$10 P&P. EXP-32-32 CHANNEL MEMORY EXPANDER KIT \$53 + \$5 P&P. A4-AM AUTO AM KIT FOR AIRBAND \$32 + \$5 P&P. INTERSTATE DEALERS: NSW: (02) 211-0531; QLD: (07) 397-0808; SA: (08) 269-4744.

SCANNERS' WORLD

Dream Machine

This rig must surely rate as every scanning enthusiast's 'dream machine'. It's just about got everything that opens and shuts.

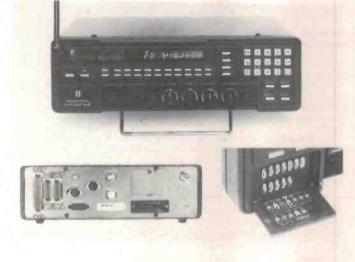
The J.I.L. SX-400 scanning receiver offers continuous coverage from 26 MHz to 520 MHz in 12 bands. AM and FM detection, two scan rates (4 & 8, ch/sec), 20 memory channels and an IEEE standard interface buss.

An optional-extra converter tronic Impo can extend this coverage to 800 MHz in 480 channels and an (03)873-3939.

. .

extra 12 spot frequencies can be received up to 3.7 GHz! A demodulator for multiplex transmission is also available.

Definitely for the scanner who must hear everything! Further details available from the Australian agents, GFS Electronic Imports, 15 McKeon Rd, Mitcham Vic. 3132. (03)873.3030

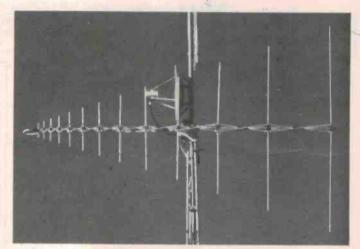


INTERESTING MARITIME CHANNELS

Here is a short list of interesting maritime channels for scanning enthusiasts in Sydney and Melbourne. The shore stations are located near the harbourside so you need to be in close range to get good reception or at a high location if you're some distance away from the harbour in either city.

Sydney		
156.375 MHz	FM	Navigation warning
156.525	FM	Maritime search & rescue
		State
156.650	FM	Maritime Services Board
156.800	FM	Weather
454.000	FM	State Maritime Services Board
454.275	FM	State Maritime Services Board
Melbourne		
156.400 MHz	FM	Harbour control
156.550	FM	Harbour control
156.600	FM	Harbour control
156.700	FM	Harbour control
498.650	FM	Sea pilots
493.475	FM	Seapilots

We would welcome more contributions to this column, particularly from the other states. Send your contributions to: The Editor, 'Scanners' World', ETI Magazine, PO Box 21, Waterloo NSW 2017.



New high gain directional antennas

GFS Electronic Imports of Mitcham, Victoria, recently announced the release of two high gain broadband directional antennas designed for use in a wide variety of VHF/UHF applications.

Particularly suited for scanning receiver use, both models provide excellent performance in fringe areas when compared with a standard discone (such as the Scan-X).

Known as the LOG-S and the LOG-SP, the new antennas are of the log periodic type. The Model LOG-S has nine elements with a claimed gain of 9½ (9 dBi) and a bandwidth from 100 to 520 MHz. Boom length is 1.02 metres.

The LOG-SP has a bandwidth from 65 to 520 MHz, comprises 13 elements and features a claimed gain of 10¹/₂ (11.5 dBi).

The scanner's manual

Dick Smith's Australian Radio Frequency Handbook, 2nd Edition, must surely be the 'text' every scanner should have.

Written by scanner 'expert', Jack McDonald, this 170-pageplus book contains a wealth of information on scanning, including a detailed table of frequencies and services you'll find on the VHF and UHF bands.

There are 15 short chapters at the start of the book that explain, in simple language, all the basics you need to know to get into scanning.

The book shows you how to build yourself a simple antenna, how to install a scanner in a car and how to interpret the various codes you hear used on the air.

The book costs \$12.95 and is listed as catalogue No. B-9600.

Its boom length is just over 3 m.

Both antennas are also quite suited to transmission applications over their designed bandwidth. Maximum input power handling is given as 200 watts.

The LOG-S and LOG-SP are both available exclusively through GFS Electronic Imports or any of their agents. Price of the LOG-S is \$89 plus \$10 freight, and the LOG-SP is \$125 plus \$10 freight.

For further information, contact GFS Electronic Imports, 15 McKeon Rd, Mitcham Vic. 3132. (03)873-3939.



Enquire at any Dick Smith store or Dick Smith Electronics, P.O. Box 321, North Ryde NSW 2113.

SHOPAROUND

This page is to assist readers in the continual search for components, kits, printed circuit boards and other parts for ETI projects and circuits. If you are looking for a particular item or project and it is not mentioned here, check with our advertisers.

ETI-464 general purpose IC audio amplifier

This project will doubtless be widely stocked as a kit. At the time of going to press, the following firms indicated they'd be stocking kits: Altronics in Perth, Dick Smith Electronics stores everywhere, Electronic Agencies in Sydney and Rod Irving Electronics in Melbourne. You might also try All Electronic Components in Melbourne.

Almost any electronics component supplier will stock the uA380 or LM380 IC and pc boards will be stocked by the suppliers listed at the end of this column.

If you want a negative or positive transparency to make your own pc board, one can be obtained for \$1.50 post paid from: ETI-464 Artwork, ETI Magazine, P.O. Box 21, Waterloo NSW 2017. Make cheques payable to 'ETI Artwork Sales' and ensure you ask for a positive or negative, according to what you need.

ETI-465 loudhailer

As this uses the ETI-464 module, the above item applies. If you want to use the sensitive and efficient Benelec 8-224 horn recommended in the article, we know that at least Electronic Agencies in Sydney will be stocking it. In Melbourne, ask for it at Rod Irving Electronics or All Electronic Components.

Kits in one form or another will be stocked by Dick Smith stores, Electronic Agencies in Sydney, Rod Irving Electronics in Melbourne and possibly Altronics in Perth and All Electronic Components in Melbourne.

Batteries and battery holders to suit this project are widely stocked in electronics stores. The ABS plastic water pipe and caps can be obtained from most hardware stores.

ETI-1520 wideband amp

This useful little project will become the output stage of our upcoming lab. standard function and pulse generator. However, it, has many other uses, as suggested in the article.

144 - July 1983 ETI

At this stage, we have no indication of who'll be stocking kits, but Rod Irving Electronics is probably a good place to start.

Parts shouldn't be too difficult to obtain. The CA3100 wideband op-amp, which forms the heart of the unit, is distributed by AWA Microelectronics in Sydney (02)638-9022. Radio Despatch and Jaycar in Sydney stock the device and possibly Rod Irving Electronics in Melbourne.

The 2N3866 and 2N2905A complementary high speed output devices are widely stocked. The Thermalloy slip-on heatsinks we used on these transistors in the prototype, part no. 2228B, are distributed by Soanar Electronics.

Printed circuit boards will be available from the suppliers listed at the end of this column.

For those of you making your own pc board, a same-size positive or negative transparency can be obtained for \$2.85 post paid from: ETI-1520 Artwork, ETI Magazine, P.O. Box 21, Waterloo NSW 2017. Make cheques payable to ETI Artwork Sales' and ensure you ask for a positive or negative, as you want.

ETI-265 'power down' timer

There's nothing in the way of 'special' components used in this project, so you should have little difficulty getting the parts together. We understand kits will be stocked by Jaycar in Sydney. Altronics in Perth and Rod Irving Electronics in Melbourne. All Electronic Components in Melbourne may also stock kits.

The Unibox case we used for our prototype (P/N 144) is stocked by **Magraths** in Melbourne but a wide variety of cases may be used — even the ubiquitous jiffy box! **Jaycar** stock a new range of smart allplastic ABS cases, of which the HB6150, 1, 2, 3 series (orange, grey, blue, black) would suit.

Printed circuit boards will be available from the suppliers listed at the end of this column. If you're making your own board and want a positive or negative same-size transparency, one can be obtained for \$1.85 post paid from ETI-265 Artwork, ETF Magazine, P.O. Box 21, Waterloo NSW 2017. Make cheques or



The mighty monitor! The Micron 300 mm (12") green screen monitor must offer just about the best value for money you can get in a computer monitor. Whilst it does feature 10 Hz to 20 MHz bandwidth and high resolution, like many others you see around, it has the unique feature of 'reverse picture' capability. By simply turning up the brightness and turning back the contrast — presto, instant black-on-white!

The display area will show 24 lines of 80 characters and the resolution is claimed to be 800 lines at the centre. It takes composite video input (neg. sync.), 70 ohms terminated, 10k unterminated. It has a full complement of trace geometry controls and an auxiliary unfiltered rectified output that can deliver 12 Vdc. (That could be handy!).

The review model was delivered by Jaycar and we couldn't wait to fire it up on the lab MicroBee. Needless to say, it performed very well indeed. In fact, we could hardly believe the 'Bee's VDU output was so good!

Enquiries for the Micron green screen monitor should be directed to Jaycar in Sydney or Altronics in Perth.

money orders payable to 'ETI Artwork Sales' and ensure you ask for a positive or negative, as you want.

Printed circuit board and panel suppliers

Almost every pc board ever published by ETI may be obtained from the following suppliers:

> All Electronic Components 118 Lonsdale St Melbourne Vic, 3000

RCS Radio 651 Forest Rd Bexley NSW 2207

Panels, meter scales and dial faces for almost every ETI project published may also be obtained from the above two firms.

For pc boards produced over the last three to five years, the following suppliers generally keep stocks on hand:

Electronic Agencies 115-117 Parramatta Rd Concord NSW 2137 and 117 York St

Sydney NSW 2000

Radio Despatch Service 869 George St Sydney NSW 2000

Rod Irving Electronics 425 High St Northcote Vic. 3070

James Phototronics 522 Grange Rd Fulham Gardens SA 5024

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computing for business

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It should be of interest to business people contemplating, implementing or already using computer data processing or to any non-technical person curious to know why and how computers are used in Australian businesses and organisations.

SMALL BUSINESS COMPUTERS: A GUIDE TO EVALUATION AND SELECTION 81134P

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Ideal for the Inexperienced user, this text emphasises management considerations in determining the feasibility, economics, evaluation, selection, contracts and practicality of installing a computer.

SMALL BUSINESS COMPUTER SYSTEMS

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DON'T (OR HOW TO CARE FOR YOUR COMPUTER)

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INVENTORY MANAGEMENT FOR SMALL COMPUTERS 39848A S24.95

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BASIC FOR BUSINESS FOR THE TRS-80 90352P

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This book covers the application of BASIC language to business uses on the TRS-80 models II and III.

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\$6.56

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RADIO STATIONS GUIDE

\$5.92 **BP55** This is an aid for all those who have a radio receiver. Shows the station site, country, frequency and/or wavelength, as well as Effective Radiation Power of the transmitter and in some cases, the station's call sign as well.

AN INTRODUCTION TO RADIO OXING **BP91**

BP91 30.30 One section is devoted to amateur band reception and the other section covers broadcast band reception, with advice on suitable equipment and the techniques employed when using that equipment. The construction of a number of useful accessories is described.

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general

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EXPERIMENTS IN ARTIFICIAL INTELLIGENCE FOR SMALL COMPUTERS

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Besides including both simple and more sophisticated burglar alarm circuits using light, infrared and ultra-sonics, this book also gives circuits for gas and smoke detectors. flood alarms, fire alarms, doorphones, baby alarms, etc.

ELECTRONIC CALCULATOR USERS' HANOBOOK S5.12 **BP33**

Presents formulae, data, methods of calculation, con Presents formulae, data, methods of calculation, con-version factors, etc. for use with the simplest or most sophisticated calculators. Includes the way to calculate using only a simple four-function calculator: fugonometric functions; hyperbolic functions; logarithms, square roots and powers

YOUR ELECTRONIC CALCULATOR AND YOUR MONEY \$4 64 **BP54**

Starts with a basic revision of percentages and decimals. then deals with montpages, cars, insurance, fuel, shopping, tax etc. There's a section on investment and the last section deals with the calculator in a small business.

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FOR SALE: STAX SRA-12S full preamplifier includes high voltage amp, drives electrostatic headphones directly. No longer made but critical spares included. Mint condition. Retailed \$1100, sell \$550 onc. (07)229-3631.

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WANTED: ELCASSETTES and player In any condition. Also four track tape deck. Lachlan MacDonald, P.O. Box 218E, Bailarat, East Vic. 3350. (053)41-3658.

AUDIORAMA RECORDING CLUB is the club for you. We seek the enthusiast and the beginner. Many services available. For details contact G. Harris, 7 Coleman Ave, Homebush NSW 2140.

FOR SALE: LED PEAK record/output level display monitor. Calibration/reference level sets, professional features (invaluable Revox A77 owners) Audio Technology. Model 510B, sell \$135. (02)620-1203

FOR SALE: HANIMEX 4400 SL synchrocorder. Cassette recorder with inbuilt slide projector controller. As new condition, with accessories, Instructions, \$190. Terry Day (02)73-0433 ext. 440.

COMMUNICATIONS

WANTED: SINGLESIDEBAND receiver, 12 V, crystal-locked, loudspeaker, covering channels used on small ships, broadcasts by Brisbane radlo. I require a circult to be designed and a set to be made. Contact H. Fraser, PO Box 158, BunderIm Qid 4556.

FOR SALE: ARC 3-channel/3-servo remote control system, never been used, \$100. Amphenol 10-channel AM transceiver, \$50. Scott (068)47-2532.

FOR SALE: SX200 scanner, IC290A, All mode 2 metres, Daiwa VHF-UHF SWR. All as new. FM321 80-channel UHF, 16-element LPY. (047)54-2214, leave message.

MISCELLANEOUS

FOR SALE: DISCO STROBE, variable flash rate, working 100%. Hardly used, great for partles, \$20. (02)451-1620.

SELL: ADDO tape punch, reader, spare punch mechanism, 10 rolls tape, \$80. Model 15 TTY, \$35. Paul Fuller, 24 Penman St, New Lambton NSW 2305. (049)57-2409 weekends.

FOR SALE: UNUSED parts, 74181N, DM74126N, DEC7476, FND500, NM74C175, DM74153N, MC14012B, DM74150N, DM7489N, DM7476N, F7442APC, MC14001B, SN 7483 AJ, DM7474N, LM325N, 7622, HEF4081P, DM7486N, MC14078B, 7705, DM7448N. (08)295-3181.

WANTED: TELONIC sweep generators, 1006-1011-1030, state price. Will consider others with same coverage. Hickman, PO Box 74, Hamilton NSW 2303. (049)48-5553.

COMPUTERS

FOR SALE: ETI-660 3K colour, modulator. Fibreglass case, hex keypad, power supply, loudspeaker, step-repeat function plus many programs on cassette, \$165. (02)542-1365

OSI SUPERBOARD II: Complete with 8K RAM, power supply, monitor, instructions and two BASIC games books. Lot \$350. Dane Howe (03)350-1646.

VOICE SYNTHESISER: Built and tested. Based on EA Compuvoice. Phoneme programmable, includes speaker, data, plugpack. Suits parallel Centronics port: System-80, Super-80 etc. \$250. Greg (02)644-3491 ah.

FOR SALE: HEWLETT-PACKARD HP-85 desktop computer, 32K. Compact, built-in VDU, randomaccess mInicartridge and thermal printer. BASIC, VIsicalc, games, tapes, paper. As new, \$2500. (03)80-5635 ah.

MODEM CLUB for System-80 and TRS-80 computers. (03)397-6972 for verbal enrolment.

FOR SALE: DBASE II including ZIP and TUTOR. Relational database management system and report generator for Osborne 1. Total value over \$900. Will sell \$400. Vic (02)747-4218.

FOR SALE: DISK DRIVES for Apple. Two lowprofile 250K drives. Brand new, never used, \$380 each. (03)339-5604 bh or (03)379-4438. WANTED TO BUY: Standard mini-floppy drive, Epson printer. B.J. Wight (08)356-0817.

FOR SALE: DIGITAL CASSETTE drives (3), 720K/ cassette, 24 Kbaud. Fast fwd/bwd file search, load a 15K file in 5 s. Full documentation including proven interface. \$120 each. (03)339-5604 bh or (03)379-4438 ah.

MICROBEE OWNERS: The Sydney MicroBee Users' Group (formerly Northside) is your most concentrated source of Information, with monthly meetings and newsletters. Phone Tony Williams (02)909-3951 or Colin Tringham (02)92-6408 ah.

FOR SALE: MICROBEE 'Guess the Number' program. Three levels of play. Tape S5. T. Knowler, 37 Bingley Crescent, Fraser ACT 2615.

ACT VIC-20 BIMONTHLY NEWSLETTER: Many interesting articles and programs. June Issue \$1.50. Bimonthly \$8 per year. Write to Chris Groenhout, 25 Kerferd St, Watson ACT 2602.

SELL: ZX 80/81, 4/8K ROM, 16K RAM. Complete with printer, tape recorder, mags, etc. Perfect running order. Half price \$399. (08)384-4208

SELL: TI RS232 Interface, \$180. Tandy Mk. VII printer, \$350. SD Z80 starter kit, \$250. B.J. Wight (08)356-0817.

FOR SALE: ZX81, 1K, leads, manual, 1.2 A power supply, numerous programs, constructional articles (add-on memory, etc). \$170 ono. F. Los, 13 Kroombit St, Dulwich HIII NSW 2203.

SELL: MICROACE (ZX80 equivalent). 2K adaptor, leads, manual, one tape, three books, \$170 ono. G. Clarke, 33 May St, Inverell NSW 2360.

SELL: CENTRONICS 737 printer. 30 cps upper/ lower case, three font styles, friction/tractor feed. With manual, carry case and leads, \$800. J. Thomas (02)546-4321.

FOR SALE: SYSTEM-80 computer, 16K RAM, joystick Interface, over \$400 worth of software, manuals, books, etc. Good condition. Software recorded on 50 tapes. \$250. Tasmania (002)72-6412.

FOR SALE: ETI-660 microcomputer, ETI-760 modulator, 3K RAM, colour option. Constructional, programming and software articles, also most of published software on tape. S140. Peter (02)708-2014.

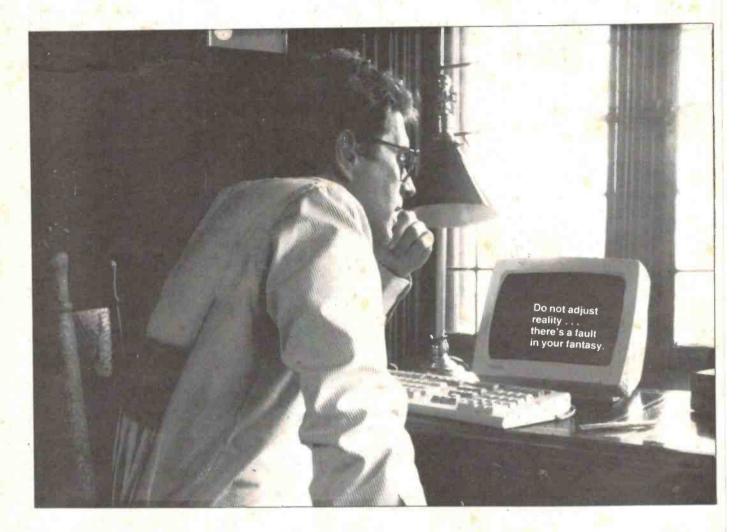
SELL: TARBELL SD/DD floppy disk controller board, \$300. Godbout RAM 16 64K, 8/16 bit memory board, \$650. A. Cummins, 17 TorrIngton Rd, Maroubra NSW 2035. (02)349-3684.

SIEMENS 100 TELEPRINTER: A limited number of operational units are available. The models come with or without a keyboard and/or tape punch or reader. They can be picked up or delivered for vehicle running costs, meet you halfway, literally. From \$65. Frank Rees, 27 King St, Boort Vic 3537.

SPARE PARTS: Most parts except motors available, to be sold as subassemblies, e.g. any part for type basket (character font) can be obtained by buying whole assembly. I have some arms (badly bent) with characters intact for those capable of changing ends. All parts from \$2 to \$15. Parts available include tape punch and reader assemblies complete, each \$5. Despatch post or rail. Frank Rees, 27 King St, Boort Vic 3537.

SUPPORT AVAILABLE: As a keen hobbyist I'll correspond with anyone needing assistance with their hobby. Frank Rees, 27 King St, Boort Vic 3537.





FOR THOSE FANS of the computer and the ubiquitous computer game who have braved the queues and the crowds to see the movie (and I use that term loosely) 'Tron', ponder a moment what went on on 'the inside'. There was the hero, whizzing around inside the circuits, through gates, over flop-flops, down the buss etc. in his own little 'carrier'. Pursuing our micro hero were numerous 'baddies', but along the way were some friendly 'components'. It was all a deadly serious fantasy in the best cowboys-versuscops-and-robbers, Indians, good-versus-evil traditions.

If you haven't seen Tron, dig it out in some suburban cinema or from a videotape library and *see it*. You'll never view a computer or computer game in

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the same benign, familiar-with-that fashion ever again.

Sidestepping the kill-or-be-killed basis of the Tron fantasy and drifting off in a direction more reminiscent of Woody Allen or Mel Brooks, just let your thoughts wander along a more whimsical or satirical path.

Imagine the sort of conversations you might hear if you could plug in to the circuits inside a computer or computer game and have all those 'instructions' decoded into speech. Your digitodoodling Dregs writer did this exercise recently while doing a dump to the office output port and found these 'conversations' drifting through his memory banks:

"I haven't had a good interruption in

microseconds," said the microprocessor to the buss driver.

"Would you like a POKE?" said the data buffer to the screen RAM.

"Seal it with a handshake," said the I/O port to the printer.

"No PEEKing," said the protected RAM to the address buss.

"PEEK before you POKE, or you'll risk corrupting your data," said the OS to the input.

Perhaps some of you Dregs fans might come up with some 'conversations' in the same vein? We'll publish the 'best' of them as they come along — no prizes, just a chance to join in the fun. Write to: "Conversations on the buss", ETI Magazine, P.O. Box 21, Waterloo NSW 2017.



The Sony CDP101 The magic of digital audio becomes a magnificent reality.

Digital Audio is a revolution. The greatest advance in home music reproduction since the



gramophone record. As you'd expect, Sony is the leader of this revolution with its magnificent CDP-101 player that offers you original studio master quality at home.

For the technically minded, the specifications read more convincingly than any superlatives • flat frequency

response over the entire audible range • dynamic range and signal to noise ratio over 90dB • perfect channel separation • immeasurable wow and flutter • negligible distortion.

Sony's CDP-101 uses an optical laser pick-up (incorporating three micro processors), it is easier to use than a conventional turntable and connects easily to your existing system.

Other features include • fully automatic linear skate front disc loading • automatic music sensor • dual function digital readout of playtime • audible fast forward and reverse • 10 function wireless remote control. Compact Discs Last Forever

Just 12 cms in diameter, the Compact Disc plays up to 60 minutes of music. It's protected from scratches, dust and finger prints by a plastic coating; and because the pick-up is a laser beam, deterioration is non-existent. Reproduction remains perfect virtually forever.

Hundreds of titles will be available with many more to follow from major companies such as CBS.

CDP-101 Specifications

Frequency Range	$5Hz-20kHz \pm 0.5dB$
Dynamic Range	more than 90dB
S/N	more than 90dB
Channel Separation	more than 90dB (at 1kHz)
Harmonic Distortion	less than 0.004% (at 1kHz)
Wow and Flutter	immeasurable

Contact Sony for the name of your nearest dealer. Sydney (02) 266 0655, Adelaide and N.T. (08) 212 2877, Brisbane (07) 44 6554, Perth (09) 3238686, Melbourne (03) 4193133, Launceston (003) 44 3078, Wollongong (042) 71 5777.

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Every car, like a new born baby, starts off with a beautiful body. But the trick is keeping it that way, so the good news is Wash-n-Wax and Glo-Wash from The Kitten System.

Now you can buy a beautiful body off the shelf, and keep your car in showroom condition. Glo-Wash is made especially for cleaning your

nce v valuable investment. for a Whereas ordinary detergent and water may only clean the surface, Glo-Wash removes stubborn road grime that would otherwise become ingrained and cause your paintwork to rapidly deteriorate. And if you want to give your car extra special protection, The Kitten System brings you Wash-n-Wax. Wash-n-Wax is

the blend of specially developed high quality cleaning agents and waxes that protects as it waxes your car's body. But the beauty of both Wash-n-Wax and Glo-Wash is their wash-on/hose-off simplicity; there's no need to chamois anymore. And a good looking body won't cost you an arm and a leg: a bottle of Wash-n-Wax or Glo-Wash will wash the

average car at least 17 times. So for a good looking body, use Glo-Wash or Wash-n-Wax

from The Kitten System at least once a week.



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